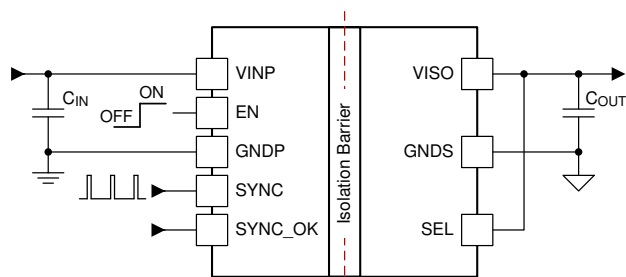


# UCC12040 High-Density, Low-EMI, 3-kV<sub>RMS</sub> Basic Isolation DC/DC Module

## 1 Features

- High-density DC/DC module with optimized integrated transformer technology
- Input voltage range: 4.5 V to 5.5 V
- Output voltages (selectable): 5.4 V, 5.0 V, 3.7 V, 3.3 V
- Output power: 500 mW
- Peak efficiency: 60%
- Line regulation (typical): 1%
- Load regulation (typical): 1.5%
- Meets CISPR32 Class B EMI limits without ferrite beads on a 2-layer PCB
- Spread spectrum modulation (SSM)
- Robust isolation barrier:
  - Isolation rating: 3 kV<sub>RMS</sub>
  - Surge capability: 6.5 kV<sub>PK</sub>
  - Working voltage: 850 V<sub>RMS</sub>
  - CMTI (typical): ±100 V/ns
- Short circuit recovery
- Thermal shutdown
- 16-pin wide-body SOIC package with > 8-mm creepage and clearance
- Extended temperature range: –40°C to 125°C
- **Safety-related certifications:**
  - 5657-V<sub>PK</sub> basic isolation per DIN V VDE V 0884-11:2017-01
  - 3000-V<sub>RMS</sub> isolation for 1 minute per UL 1577
  - UL certification per IEC 60950-1, IEC 62368-1 and IEC 60601-1 end equipment standards
  - CQC approval per GB4943.1-2011



Simplified Application

## 2 Applications

- PLC analog input & output modules
- Grid protection relays & smart breakers
- Medical patient monitors
- Isolated bias for digital isolators
- Isolated bias for RS-485/RS-422/CAN transceivers
- Isolated bias for voltage and current sensors

## 3 Description

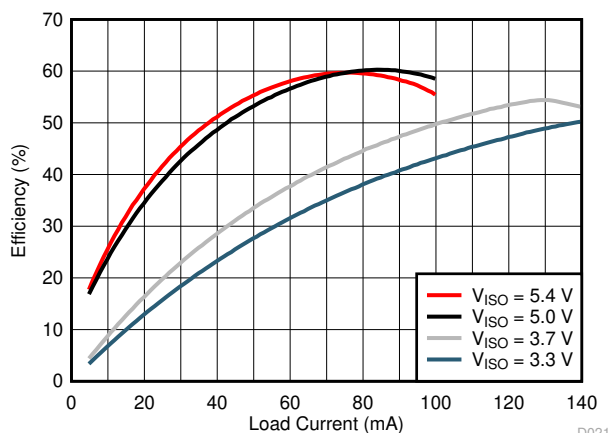
UCC12040 is a DC/DC power module with 3-kV<sub>RMS</sub> basic isolation rating designed to provide efficient, isolated power to isolated circuits that require a bias supply with a well-regulated output voltage. The device integrates a transformer and DC/DC controller with a proprietary architecture to provide 500 mW (typical) of isolated power with low EMI.

The UCC12040 integrates protection features for increased system robustness. The device also has an enable pin, synchronization capability, and regulated 5-V or 3.3-V output options with headroom. The UCC12040 is a low-profile, miniaturized solution offered in a wide-body SOIC package with 2.65-mm height.

### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)
UCC12040	DVE SOIC (16)	10.30 mm × 7.50 mm

- (1) For all available packages, see the orderable addendum at the end of the data sheet.



V<sub>INP</sub> = 5.0 V

T<sub>A</sub> = 25°C

Typical Efficiency vs. Load

D021



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## 4 Revision History

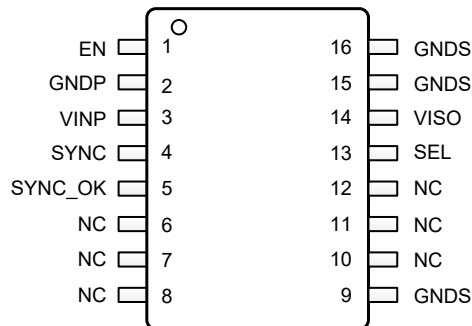
NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision A (April 2020) to Revision B (May 2021)	Page
• Added Safety Certification Information. ....	1
• Updated ESD values. ....	4
• Updated voltage isolation specs per DIN V VDE V 0884-11:2017-01.....	5
• Updated Switching Characteristics (non-sync).....	8

Changes from Revision * (December 2019) to Revision A (April 2020)	Page
• Updated $I_{VINO}$ Typical Values in <i>Electrical Characteristics</i> Table .....	7

## 5 Pin Configuration and Functions

### DVE Package 16-Pin SOIC Top View



**Table 5-1. Pin Functions**

PIN		TYPE <sup>(1)</sup>	DESCRIPTION
NAME	NO.		
EN	1	I	Enable pin. Forcing EN low disables the device. Pull high to enable normal device functionality.
GNDP	2	P	Power ground return connection for VINP.
GNDP	9	P	Connect to GNDP plane on printed circuit board. Do not use as only ground connection for VISO. Ensure pin 15 is connected to circuit ground.
	16		
GNDP	15	P	Secondary side ground return connection for VISO. Connect bypass capacitor from VISO to this pin.
NC	6	—	Pins internally connected together. No other electrical connection. Pins belong to primary-side voltage domain. Connect to GNDP on printed circuit board.
	7		
	8		
	10	—	No internal connection. Pin belongs to isolated voltage domain. Connect to GNDP on printed circuit board.
	11		
	12		
SYNC	4	I	Synchronous clock input pin. Provide a clock signal to synchronize multiple devices or connect to GNDP for standalone operation using the internal oscillator. If the SYNC pin is left open make sure to separate it from any switching noise to avoid false clock coupling.
SYNC_OK	5	O	Active-low, open-drain diagnostic output. Pin is asserted LOW if there is no external SYNC clock or one that is outside of the operating range of the UCC12040 is detected. In this state, the external clock is ignored and the DC-DC converter is clocked by the internal oscillator. The pin is in high-impedance if a clock is applied on SYNC.
SEL	13	I	V <sub>ISO</sub> selection pin. V <sub>ISO</sub> setpoint is 5.0 V when SEL is shorted to V <sub>ISO</sub> , 5.4 V when SEL is connected to V <sub>ISO</sub> through a 100-kΩ resistor, 3.3 V when SEL is shorted to GNDP, and 3.7 V when SEL is connected to GNDP through a 100-kΩ resistor. For more information see the <a href="#">Section 7.4</a> section.
VINP	3	P	Primary side input supply voltage pin. A 10-μF ceramic capacitor to GNDP on pin 2, placed close to the device pins, is required.
VISO	14	P	Isolated supply voltage pin. A 10-μF ceramic capacitor to GNDP on pin 15, placed close to the device pins, is required. See <a href="#">Section 8.2.2.1</a> section.

(1) P = Power, G = Ground, I = Input, O = Output

## 6 Specifications

### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

	MIN	MAX	UNIT
V <sub>INP</sub> to GNDP	-0.3	6.0	V
EN, SYNC, SYNC_OK, to GNDP	-0.3	V <sub>INP</sub> + 0.3, ≤ 6.0	V
V <sub>ISO</sub> to GNDS	-0.3	6.0	V
SEL to GNDS	-0.3	V <sub>ISO</sub> + 0.3, ≤ 6.0	V
V <sub>ISO</sub> output power at T <sub>a</sub> = 25°C, P <sub>OUT_MAX</sub> <sup>(2)</sup>		675	mW
Operating junction temperature range, T <sub>J</sub>	-40	150	°C
Storage temperature, T <sub>stg</sub>	-65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) See the [Section 7.3.3](#) section for maximum rated values across temperature and V<sub>INP</sub> conditions for each different V<sub>ISO</sub> output mode.

### 6.2 ESD Ratings

			VALUE	UNIT
V <sub>(ESD)</sub>	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±3000	V
		Charged-device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>	±1000	

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V <sub>INP</sub>	Primary side supply voltage	4.5	5.0	5.5	V
V <sub>EN</sub>	EN pin input voltage	0		5.5	V
V <sub>SYNC</sub>	SYNC pin input voltage	0		5.5	V
V <sub>SYNC-OK</sub>	SYNC_OK pin drain pin voltage	0		5.5	V
V <sub>ISO</sub>	Isolated power supply voltage	0		5.7	V
V <sub>SEL</sub>	Input voltage	0		5.7	V
f <sub>SYNC</sub>	External DC/DC converter synchronization signal frequency	14.4	16.0	17.6	MHz
P <sub>VISO</sub>	V <sub>ISO</sub> output power at T <sub>a</sub> = 25°C <sup>(1)</sup>			500	mW
T <sub>a</sub>	Ambient temperature	-40		125	°C
T <sub>J</sub>	Junction temperature	-40		150	°C

- (1) See the [Section 7.3.3](#) section for maximum rated values across temperature and V<sub>INP</sub> conditions for each different V<sub>ISO</sub> output mode.

### 6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		DVE (SOIC)	UNIT
		16 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	63.8	°C/W
R <sub>θJC(top)</sub>	Junction-to-case (top) thermal resistance	21.4	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	38.5	°C/W
ψ <sub>JT</sub>	Junction-to-top characterization parameter	10.2	°C/W
ψ <sub>JB</sub>	Junction-to-board characterization parameter	37.2	°C/W

THERMAL METRIC <sup>(1)</sup>		DVE (SOIC)	UNIT
		16 PINS	
R <sub>θJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	—	°C/W

- (1) The value of R<sub>θJA</sub> given in this table is only valid for comparison with other packages and can not be used for design purposes. This value was calculated in accordance with JESD 51-7, and simulated on a 4-layer JEDEC board when P<sub>DP</sub> = 129 mW, P<sub>DC</sub> = 142 mW and P<sub>DT</sub> = 129 mW. The board temperature is taken from Pin 12. For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

## 6.5 Power Ratings

V<sub>INP</sub> = 5.0V, C<sub>INP</sub> = C<sub>OUT</sub> = 10 uF, T<sub>J</sub> = 150°C, Internal Clock mode

PARAMETER		TEST CONDITIONS	VALUE	UNIT
P <sub>D</sub>	Power dissipation	SEL connected to GNDS (3.3-V V <sub>ISO</sub> output mode), I <sub>ISO</sub> = 135 mA	460	mW
P <sub>DP</sub>	Power dissipation by driver side (primary)		148	mW
P <sub>DS</sub>	Power dissipation by rectifier side (secondary)		164	mW
P <sub>DT</sub>	Power dissipation by transformer		148	mW

## 6.6 Insulation Specifications

PARAMETER		TEST CONDITIONS	VALUE	UNIT
<b>GENERAL</b>				
CLR	External clearance <sup>(1)</sup>	Shortest terminal-to-terminal distance through air	> 8	mm
CPG	External creepage <sup>(1)</sup>	Shortest terminal-to-terminal distance across the package surface	> 8	mm
DTI	Distance through the insulation	Minimum internal gap (internal clearance)	> 120	μm
CTI	Comparative tracking index	DIN EN 60112 (VDE 0303-11); IEC 60112	> 600	V
	Material group	According to IEC 60664-1	I	
	Overvoltage Category	Rated mains voltage ≤ 300 V <sub>RMS</sub>	I-IV	
		Rated mains voltage ≤ 600 V <sub>RMS</sub>	I-IV	
		Rated mains voltage ≤ 1000 V <sub>RMS</sub>	I-III	
<b>DIN V VDE V 0884-11:2017-01<sup>(2)</sup></b>				
V <sub>IORM</sub>	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	1202	V <sub>PK</sub>
V <sub>IOWM</sub>	Maximum working isolation voltage	AC voltage (sine wave) Time dependent dielectric breakdown (TDDB) test	850	V <sub>RMS</sub>
		DC voltage	1202	V <sub>DC</sub>
V <sub>IOTM</sub>	Maximum transient isolation voltage	V <sub>TEST</sub> = V <sub>IOTM</sub> , t = 60s (qualification); V <sub>TEST</sub> = 1.2 × V <sub>IOTM</sub> , t = 1s (100% production)	4243	V <sub>PK</sub>
V <sub>IOSM</sub>	Maximum surge isolation voltage <sup>(3)</sup>	Test method per IEC 62368-1, 1.2/50 μs waveform, V <sub>TEST</sub> = 1.3 × V <sub>IOSM</sub> = 6500 V <sub>PK</sub> (qualification)	5000	V <sub>PK</sub>
q <sub>pd</sub>	Apparent charge <sup>(4)</sup>	Method a: After I/O safety test subgroup 2/3, V <sub>ini</sub> = V <sub>IOTM</sub> , t <sub>ini</sub> = 60 s; V <sub>pd(m)</sub> = 1.2 × V <sub>IORM</sub> = 1442 V <sub>PK</sub> , t <sub>m</sub> = 10 s	≤ 5	pC
		Method a: After environmental tests subgroup 1, V <sub>ini</sub> = V <sub>IOTM</sub> , t <sub>ini</sub> = 60 s; V <sub>pd(m)</sub> = 1.2 × V <sub>IORM</sub> = 1442 V <sub>PK</sub> , t <sub>m</sub> = 10 s	≤ 5	
		Method b1: At routine test (100% production) and preconditioning (type test) V <sub>ini</sub> = 1.2 × V <sub>IOTM</sub> , t <sub>ini</sub> = 1 s; V <sub>pd(m)</sub> = 1.5 × V <sub>IORM</sub> = 1803 V <sub>PK</sub> , t <sub>m</sub> = 1 s	≤ 5	
C <sub>IO</sub>	Barrier capacitance, input to output <sup>(5)</sup>	V <sub>IO</sub> = 0.4 sin (2πft), f = 1 MHz	~3.5	pF
R <sub>IO</sub>	Isolation resistance, input to output <sup>(5)</sup>	V <sub>IO</sub> = 500 V, T <sub>A</sub> = 25°C	> 10 <sup>12</sup>	Ω
		V <sub>IO</sub> = 500 V, 100°C ≤ T <sub>A</sub> ≤ 125°C	> 10 <sup>11</sup>	
		V <sub>IO</sub> = 500 V at T <sub>S</sub> = 150°C	> 10 <sup>9</sup>	
	Pollution degree		2	
	Climatic category		40/125/21	
<b>UL 1577</b>				

PARAMETER		TEST CONDITIONS	VALUE	UNIT
$V_{ISO}$	Withstand isolation voltage	$V_{TEST} = V_{ISO} = 3000 V_{RMS}$ , $t = 60$ s (qualification); $V_{TEST} = 1.2 \times V_{ISO} = 3600 V_{RMS}$ , $t = 1$ s (100% production)	3000	$V_{RMS}$

- Creepage and clearance requirements should be applied according to the specific equipment isolation standards of an application. Care should be taken to maintain the creepage and clearance distance of a board design to ensure that the mounting pads of the isolator on the printed-circuit board do not reduce this distance. Creepage and clearance on a printed-circuit board become equal in certain cases. Techniques such as inserting grooves and/or ribs on a printed circuit board are used to help increase these specifications.
- This coupler is suitable for *safe electrical insulation* only within the maximum operating ratings. Compliance with the safety ratings shall be ensured by means of suitable protective circuits.
- Testing is carried out in air or oil to determine the intrinsic surge immunity of the isolation barrier.
- Apparent charge is electrical discharge caused by a partial discharge (pd).
- All pins on each side of the barrier tied together creating a two-terminal device

## 6.7 Safety-Related Certifications

VDE	UL		CQC
Certified according to DIN V VDE V 0884-11:2017- 01	Certified according to IEC 60601-1	Certified according to IEC 60950-1 and IEC 62368- 1	Recognized under UL 1577 Component Recognition Program
Basic insulation Maximum transient isolation voltage, 4243 VPK; Maximum repetitive peak isolation voltage, 1202 VPK; Maximum surge isolation voltage, 5000 VPK	Basic insulation per UL 60601- 1:14 and IEC 60601-1 Ed.3+A1,AAMI ES 60601-1:2005/(R)2012 and A1:2012, C1:2009/(R)2012 and A2:2010/(R)2012 CSA C22.2 No. 60601-1:2014 IEC 60601-1:2012, 1 MOPP (Means of Patient Protection), 250 VRMS maximum working voltage	Basic insulation per UL 60950-1-07+A1+A2, IEC 60950-1 2nd Ed.+A1+A2, UL 62368-1- 14 and IEC 62368-1 2nd Ed., 850 VRMS maximum working voltage (pollution degree 1, material group I)	Single protection, 3000 VRMS
Certificate number: 40047657	Certificate number: US-36773-A1-UL	Certificate number: US-36706-UL, US-36707-UL	File number: E181974
			Certified according to GB4943.1-2011
			Basic insulation, Altitude $\leq$ 5000 m, Tropical Climate, 800 VRMS maximum working voltage
			Certificate number: CQC20001268791

## 6.8 Safety Limiting Values

Safety limiting intends to minimize potential damage to the isolation barrier upon failure of input or output circuitry.

PARAMETER		TEST CONDITIONS	MAX	UNIT
$I_S$	Safety input current <sup>(1)</sup>	$R_{\theta JA} = 63.8^\circ\text{C/W}$ , $V_I = 5.5$ V, $T_J = 150^\circ\text{C}$ , $T_A = 25^\circ\text{C}$	356	mA
		$R_{\theta JA} = 63.8^\circ\text{C/W}$ , $V_I = 4.5$ V, $T_J = 150^\circ\text{C}$ , $T_A = 25^\circ\text{C}$	435	
$P_S$	Safety input power	$R_{\theta JA} = 63.8^\circ\text{C/W}$ , $T_J = 150^\circ\text{C}$ , $T_A = 25^\circ\text{C}$	1960	mW
$T_S$	Safety temperature <sup>(1)</sup>		150	$^\circ\text{C}$

- The maximum safety temperature,  $T_S$ , has the same value as the maximum junction temperature,  $T_J$ , specified for the device. The  $I_S$  and  $P_S$  parameters represent the safety current and safety power respectively. The maximum limits of  $I_S$  and  $P_S$  should not be exceeded. These limits vary with the ambient temperature,  $T_A$ . The junction-to-air thermal resistance,  $R_{\theta JA}$ , in the [Thermal Information](#) table is that of a device installed on a high-K test board for leaded surface-mount packages. Use these equations to calculate the value for each parameter:  $T_J = T_A + R_{\theta JA} \times P$ , where P is the power dissipated in the device.  $T_{J(max)} = T_S = T_A + R_{\theta JA} \times P_S$ , where  $T_{J(max)}$  is the maximum allowed junction temperature.  $P_S = I_S \times V_I$ , where  $V_I$  is the maximum input voltage.

## 6.9 Electrical Characteristics

Over operating temperature range ( $T_J = -40^{\circ}\text{C}$  to  $150^{\circ}\text{C}$ ),  $V_{\text{INP}} = 4.5\text{V}$  to  $5.5\text{V}$ ,  $C_{\text{INP}} = C_{\text{OUT}} = 10\ \mu\text{F}$ , SEL connected to  $V_{\text{ISO}}$ , internal clock mode, unless otherwise noted. All typical values at  $T_J = 25^{\circ}\text{C}$  and  $V_{\text{INP}} = 5.0\text{V}$ .

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>INPUT SUPPLY</b>						
$I_{\text{VINQ}}$	VINP quiescent current, disabled	EN=LOW			100	$\mu\text{A}$
$I_{\text{VINO}}$	VINP operating current, no load	EN=HI; SEL shorted to $V_{\text{ISO}}$ (5.0V output)		50		mA
		EN=HI; SEL 100k $\Omega$ to $V_{\text{ISO}}$ (5.4V output)		45		
		EN=HI; SEL shorted to GNDS (3.3V output)		90		
		EN=HI; SEL 100k $\Omega$ to GNDS (3.7V output)		80		
$I_{\text{VIN\_SC}}$	DC current from VINP supply under short circuit on $V_{\text{ISO}}$	$V_{\text{ISO}}$ short to GNDS		245		mA
$V_{\text{UVPR}}$	VINP under-voltage lockout rising threshold			4.2		V
$V_{\text{UVPF}}$	VINP under-voltage lockout falling threshold			3.7		V
$V_{\text{UVPH}}$	VINP under-voltage lockout hysteresis			0.5		V
<b>EN, SYNC INPUT PINS</b>						
$V_{\text{IR}}$	Input voltage threshold, logic HIGH	Rising edge			2.2	V
$V_{\text{IF}}$	Input voltage threshold, logic LOW	Falling edge	0.8			V
$I_{\text{EN}}$	Enable Pin Input Current	$V_{\text{EN}} = 5.0\text{V}$		5	10	$\mu\text{A}$
$I_{\text{SYNC}}$	SYNC Pin Input Current	$V_{\text{SYNC}} = 5.0\text{V}$		0.02	1	$\mu\text{A}$
<b>SYNC_OK PIN</b>						
$V_{\text{OL}}$	SYNC_OK output low voltage	$I_{\text{SYNC\_OK}} = -2\text{mA}$		0.15		V
$I_{\text{LKG\_SYNC\_OK}}$	SYNC_OK pin leakage current	$V_{\text{SYNC\_OK}} = 5.0\text{V}$			1	$\mu\text{A}$
<b>DC/DC CONVERTER</b>						
$V_{\text{ISO}}$	Isolated supply output voltage	SEL shorted to $V_{\text{ISO}}$ (5.0V output); $I_{\text{ISO}} = 55\text{mA}$ (2)	4.7	5	5.3	V
		SEL 100k $\Omega$ to $V_{\text{ISO}}$ (5.4 V output); $I_{\text{ISO}} = 45\text{mA}$ (2)	5.1	5.4	5.7	V
		SEL shorted to GNDS (3.3V output); $I_{\text{ISO}} = 100\text{mA}$ (2)	3.1	3.3	3.5	V
		SEL 100k $\Omega$ to GNDS (3.7 V output); $I_{\text{ISO}} = 90\text{mA}$ (2)	3.5	3.7	3.9	V
$V_{\text{ISO(RIP)}}$	Voltage ripple on isolated supply output (pk-pk)	20-MHz bandwidth, CLOAD = 10 $\mu\text{F}$    0.1 $\mu\text{F}$ , SEL shorted to $V_{\text{ISO}}$ (5.0V output); $I_{\text{ISO}} = 100\text{mA}$		50		mV
		20-MHz bandwidth, CLOAD = 10 $\mu\text{F}$    0.1 $\mu\text{F}$ , SEL 100k $\Omega$ to $V_{\text{ISO}}$ (5.4V output); $I_{\text{ISO}} = 90\text{mA}$		50		mV
		20-MHz bandwidth, CLOAD = 10 $\mu\text{F}$    0.1 $\mu\text{F}$ , SEL shorted to GNDS (3.3V output); $I_{\text{ISO}} = 145\text{mA}$		50		mV
		20-MHz bandwidth, CLOAD = 10 $\mu\text{F}$    0.1 $\mu\text{F}$ , SEL shorted to GNDS (3.7V output); $I_{\text{ISO}} = 130\text{mA}$		50		mV
$V_{\text{ISO(LINE)}}$	$V_{\text{ISO}}$ DC line regulation	SEL shorted to $V_{\text{ISO}}$ (5.0 V output); $I_{\text{ISO}} = 50\text{mA}$ , $V_{\text{INP}} = 4.5\text{V}$ to $5.5\text{V}$		1%		
		SEL shorted to GNDS (3.3 V output); $I_{\text{ISO}} = 75\text{mA}$ , $V_{\text{INP}} = 4.5\text{V}$ to $5.5\text{V}$		1%		

Over operating temperature range ( $T_J = -40^\circ\text{C}$  to  $150^\circ\text{C}$ ),  $V_{INP} = 4.5\text{V}$  to  $5.5\text{V}$ ,  $C_{INP} = C_{OUT} = 10\ \mu\text{F}$ , SEL connected to  $V_{ISO}$ , internal clock mode, unless otherwise noted. All typical values at  $T_J = 25^\circ\text{C}$  and  $V_{INP} = 5.0\text{V}$ .

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{ISO(\text{LOAD})}$	$V_{ISO}$ DC load regulation	SEL shorted to $V_{ISO}$ (5.0 V output); $I_{ISO} = 0$ to 100 mA		1.5%		
	$V_{ISO}$ DC load regulation	SEL shorted to GNDS (3.3 V output); $I_{ISO} = 0$ to 145 mA		1.5%		
EFF	Efficiency at maximum recommended load <sup>(1)</sup>	SEL shorted to $V_{ISO}$ (5.0 V output); $I_{ISO} = 100$ mA		60%		
		SEL 100k $\Omega$ to $V_{ISO}$ (5.4V output); $I_{ISO} = 90$ mA		60%		
		SEL shorted to GNDS (3.3V output); $I_{ISO} = 145$ mA		50%		
		SEL 100k $\Omega$ to GNDS (3.7V output); $I_{ISO} = 130$ mA		53%		
$t_{RISE}$	$V_{ISO}$ rise time, 10% - 90%	EN = change from LO to HI, SEL shorted to $V_{ISO}$ (5.0V output); $I_{ISO} = 1$ mA		750		$\mu\text{s}$
		EN = change from LO to HI, SEL 100k $\Omega$ to GNDS (3.3V output); $I_{ISO} = 1$ mA		300		$\mu\text{s}$
<b>THERMAL SHUTDOWN</b>						
$TSD_{THR}$	Thermal shutdown threshold	Junction Temperature, Rising		165		$^\circ\text{C}$
$TSD_{HYST}$	Thermal shutdown hysteresis	Junction Temperature, Falling		27		$^\circ\text{C}$

(1) Efficiency calculation:  $EFF = (V_{ISO} \times I_{ISO}) / (V_{INP} \times I_{INP})$

(2) See the [Section 7.3.3](#) section for discussion of  $V_{ISO}$  regulation across load and temperature conditions for all output voltage settings.

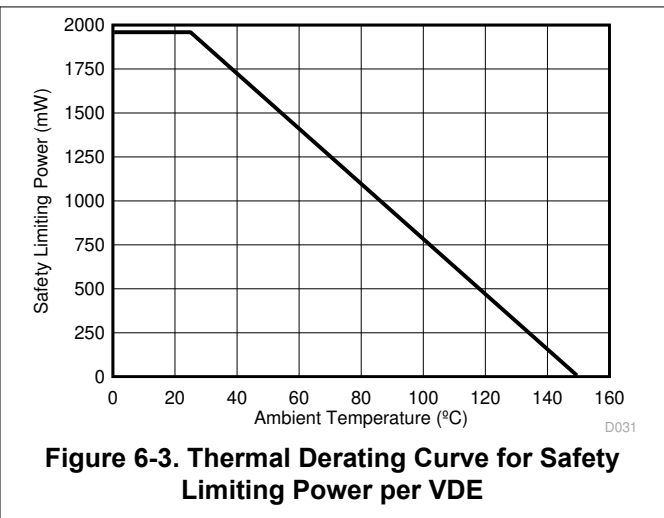
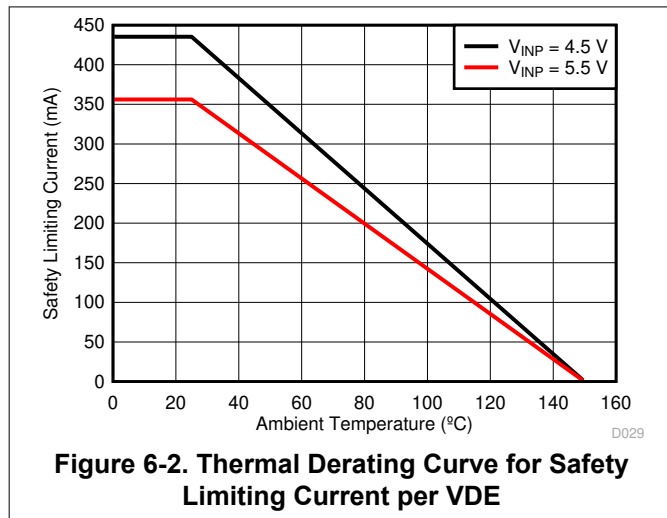
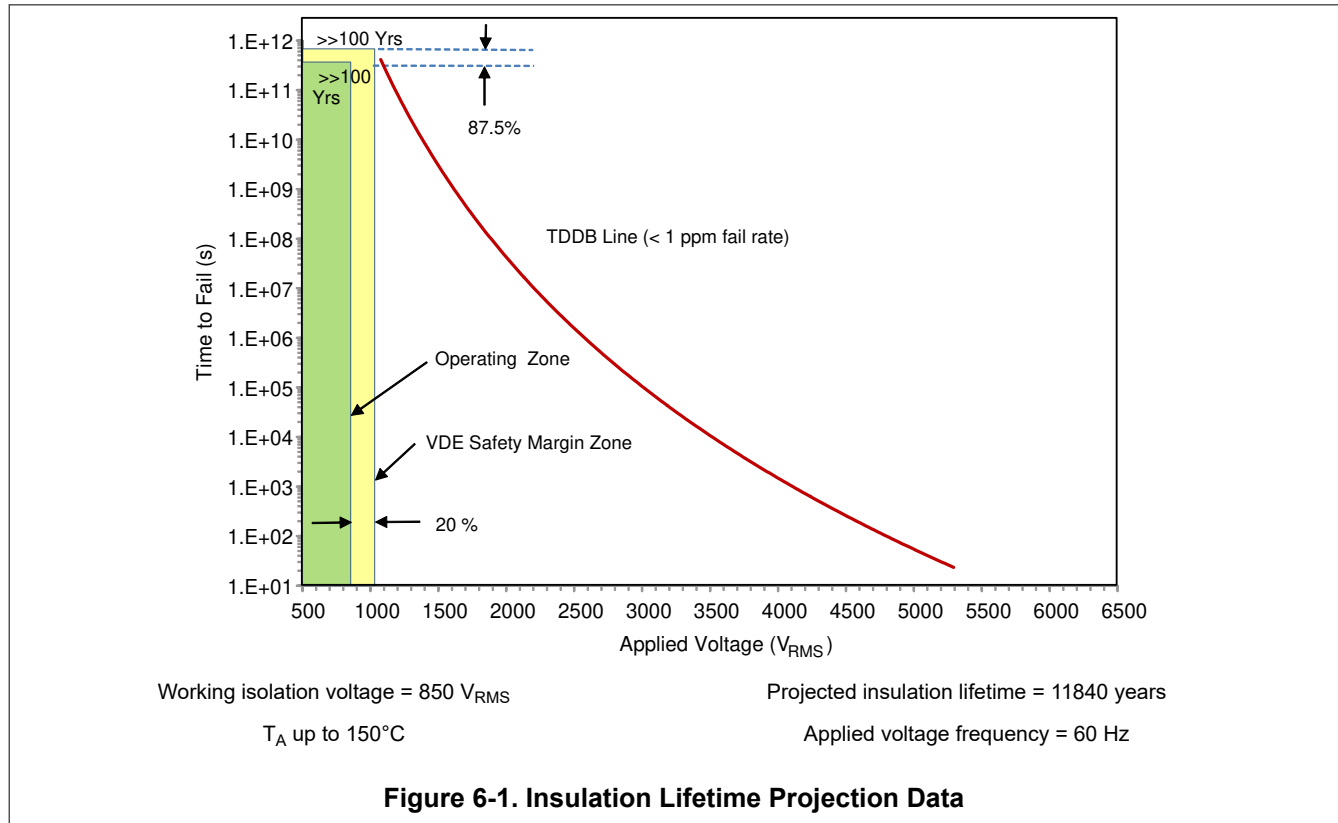
## 6.10 Switching Characteristics

Over operating temperature range ( $T_J = -40^\circ\text{C}$  to  $150^\circ\text{C}$ ),  $V_{INP} = 4.5\text{V}$  to  $5.5\text{V}$ ,  $C_{INP} = C_{OUT} = 10\ \mu\text{F}$ , SEL connected to  $V_{ISO}$ , internal clock mode, unless otherwise noted. All typical values at  $T_J = 25^\circ\text{C}$  and  $V_{INP} = 5.0\text{V}$ .

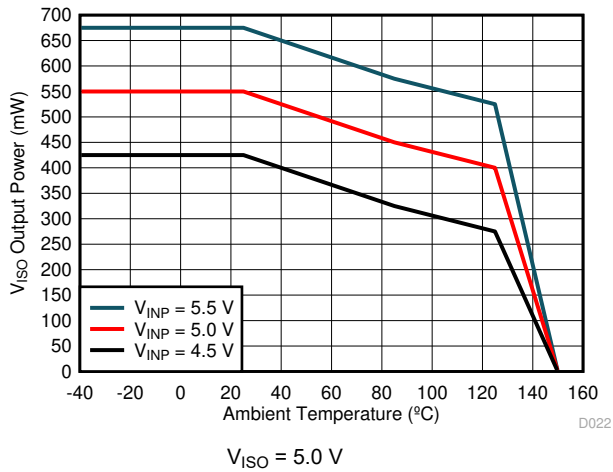
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$f_{SW\_INT}$	DC/DC Converter Clock	Internal clock mode	7.2	8	8.8	MHz
CMTI	Static common-mode transient immunity	Slew Rate of GNDP versus GNDS, $V_{CM} = 1000\ \text{V}$		100		V/ns



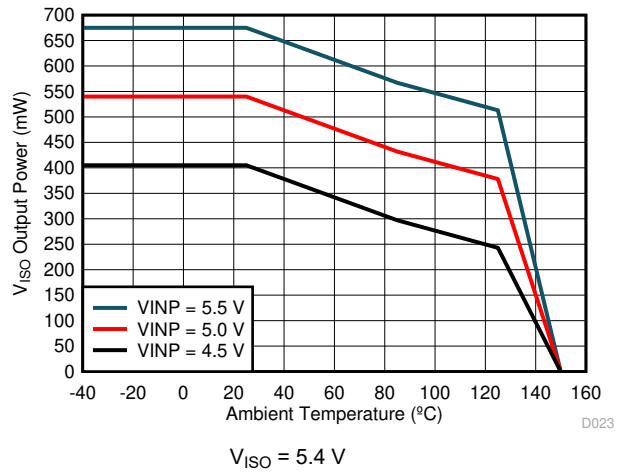
### 6.11 Insulation Characteristics Curves



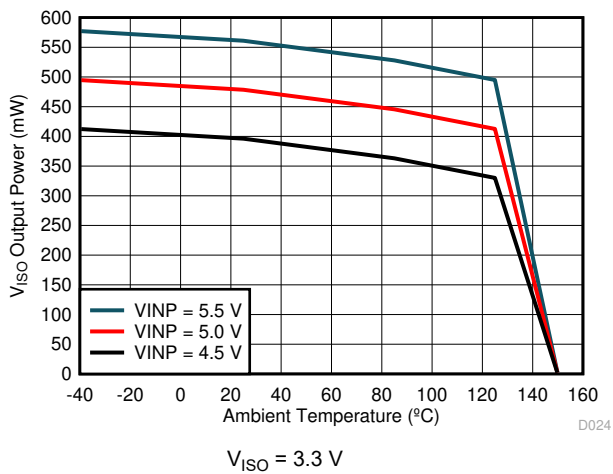
## 6.12 Typical Characteristics



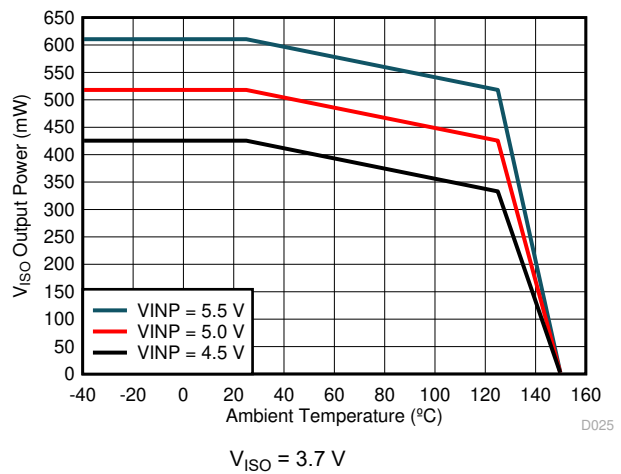
**Figure 6-4. Maximum V<sub>ISO</sub> Output Power vs. Temperature**



**Figure 6-5. Maximum V<sub>ISO</sub> Output Power vs. Temperature**



**Figure 6-6. Maximum V<sub>ISO</sub> Output Power vs. Temperature**



**Figure 6-7. Maximum V<sub>ISO</sub> Output Power vs. Temperature**

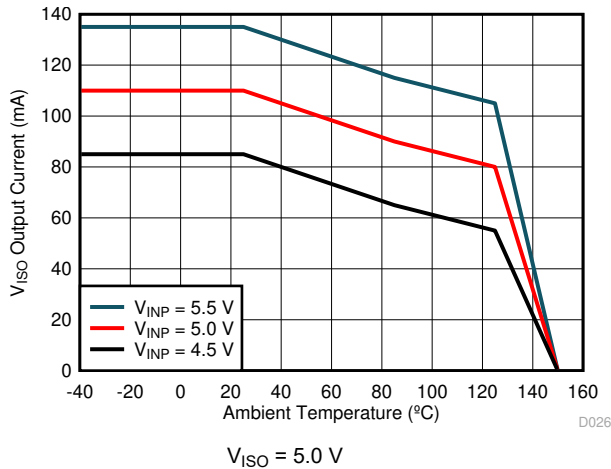


Figure 6-8. Maximum V<sub>ISO</sub> Output Current vs. Temperature

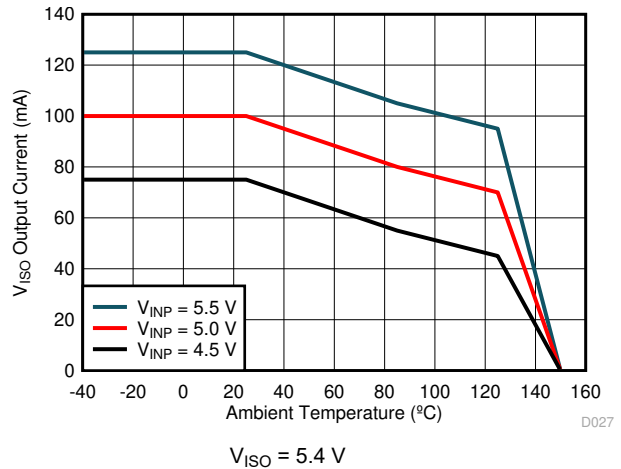


Figure 6-9. Maximum V<sub>ISO</sub> Output Current vs. Temperature

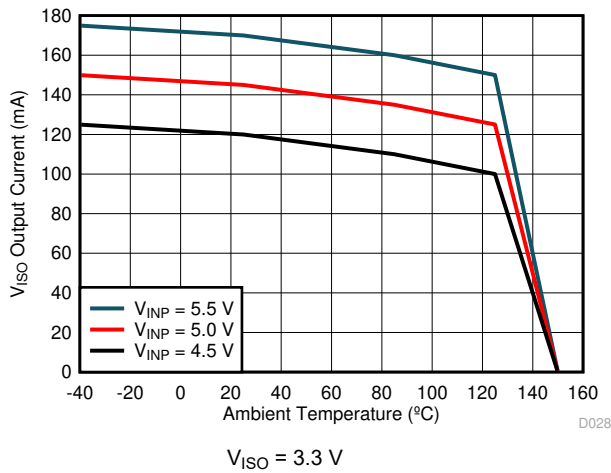


Figure 6-10. Maximum V<sub>ISO</sub> Output Current vs. Temperature

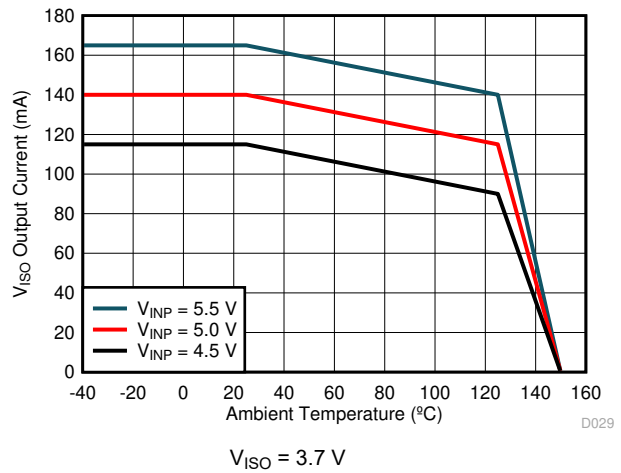


Figure 6-11. Maximum V<sub>ISO</sub> Output Current vs. Temperature

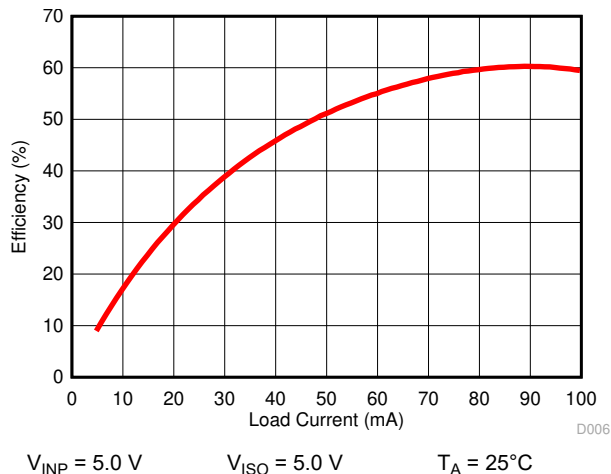


Figure 6-12. Power Supply Efficiency vs Load Current (I<sub>ISO</sub>)

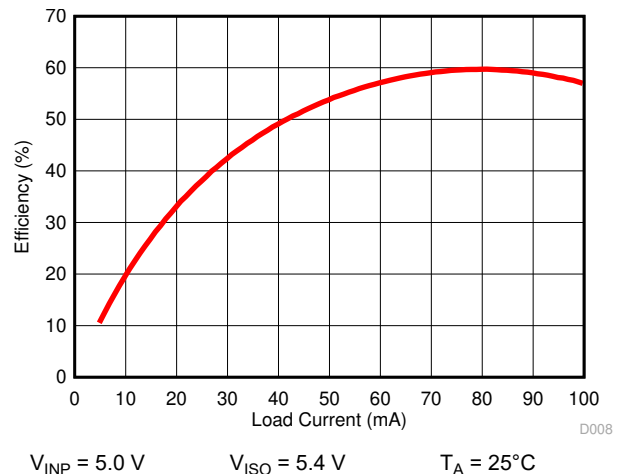
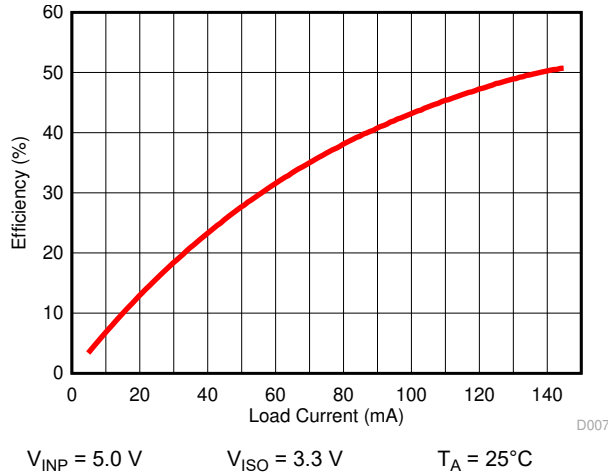
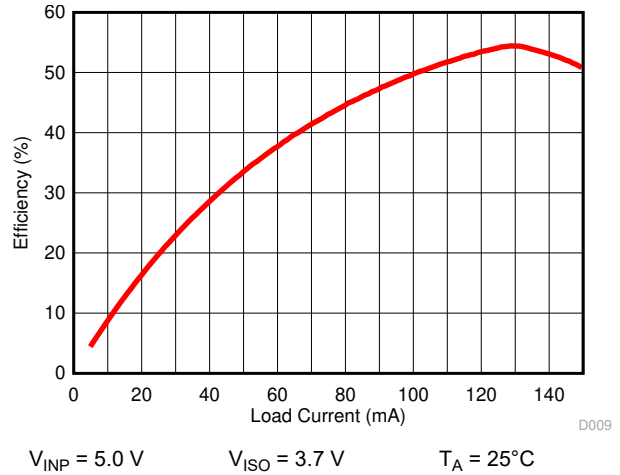


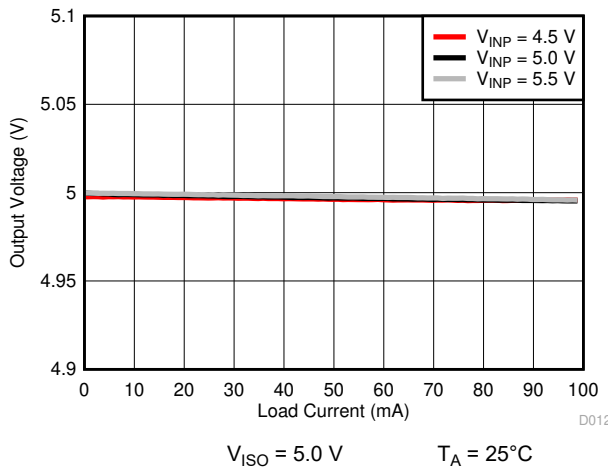
Figure 6-13. Power Supply Efficiency vs Load Current (I<sub>ISO</sub>)



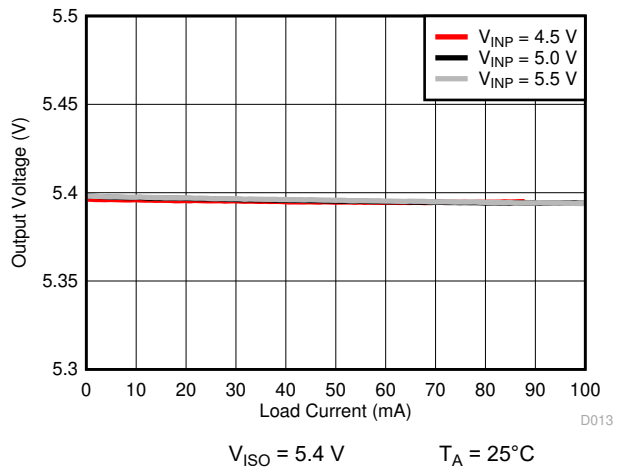
**Figure 6-14. Power Supply Efficiency vs Load Current ( $I_{ISO}$ )**



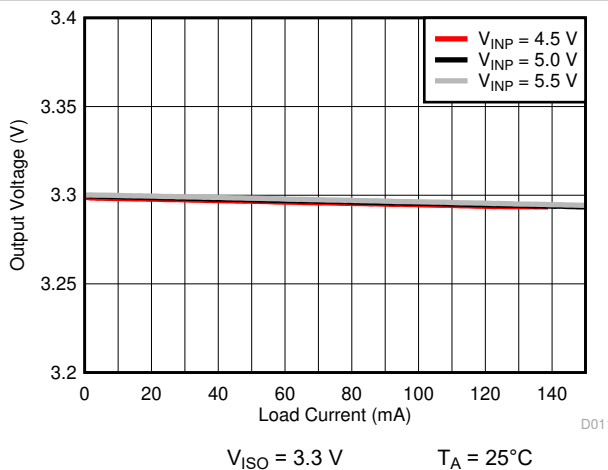
**Figure 6-15. Power Supply Efficiency vs Load Current ( $I_{ISO}$ )**



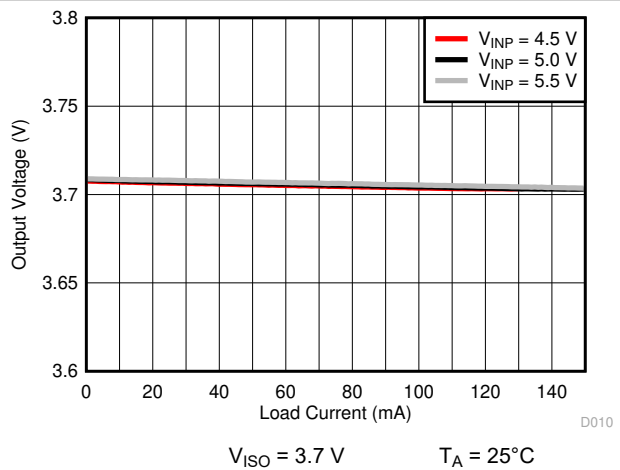
**Figure 6-16. Isolated Supply Voltage ( $V_{ISO}$ ) vs Load Current ( $I_{ISO}$ )**



**Figure 6-17. Isolated Supply Voltage ( $V_{ISO}$ ) vs Load Current ( $I_{ISO}$ )**



**Figure 6-18. Isolated Supply Voltage ( $V_{ISO}$ ) vs Load Current ( $I_{ISO}$ )**



**Figure 6-19. Isolated Supply Voltage ( $V_{ISO}$ ) vs Load Current ( $I_{ISO}$ )**

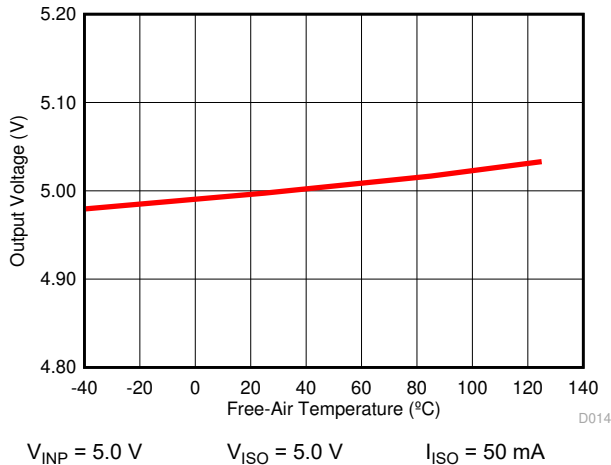


Figure 6-20. Isolated Supply Voltage ( $V_{ISO}$ ) vs Free-Air Temperature

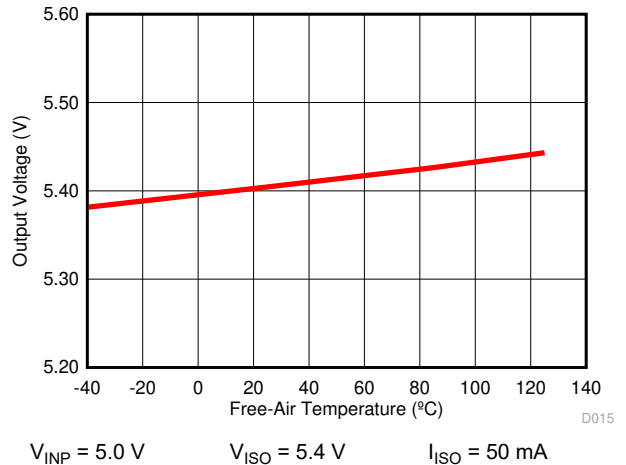


Figure 6-21. Isolated Supply Voltage ( $V_{ISO}$ ) vs Free-Air Temperature

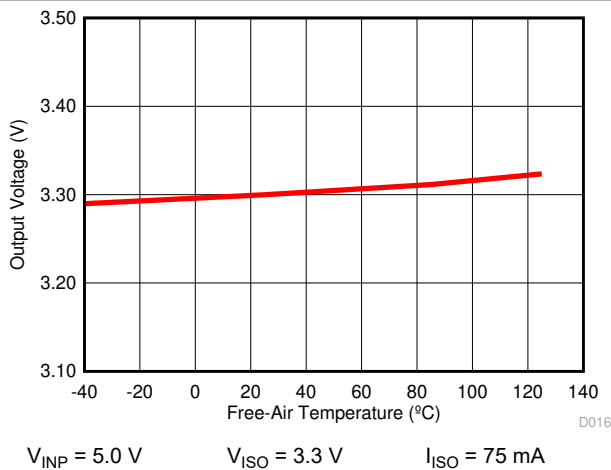


Figure 6-22. Isolated Supply Voltage ( $V_{ISO}$ ) vs Free-Air Temperature

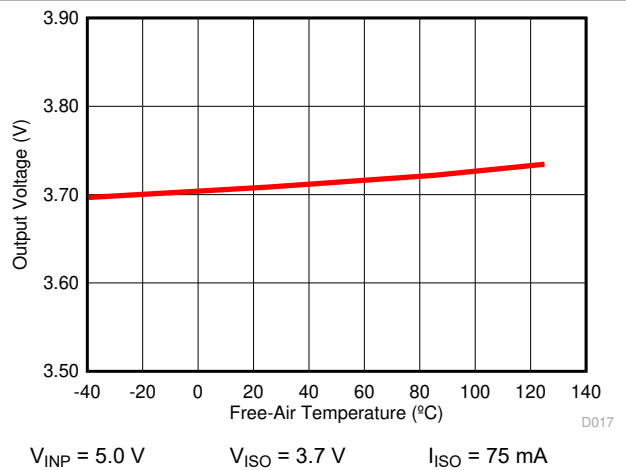


Figure 6-23. Isolated Supply Voltage ( $V_{ISO}$ ) vs Free-Air Temperature

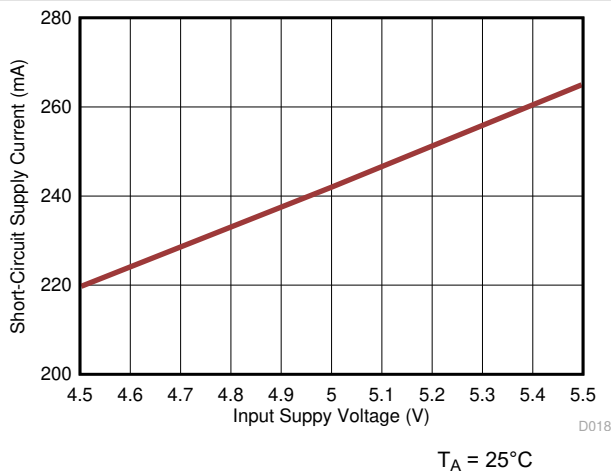


Figure 6-24. Short-Circuit Supply Current ( $I_{VIN\_SC}$ ) vs Supply Voltage ( $V_{INP}$ )

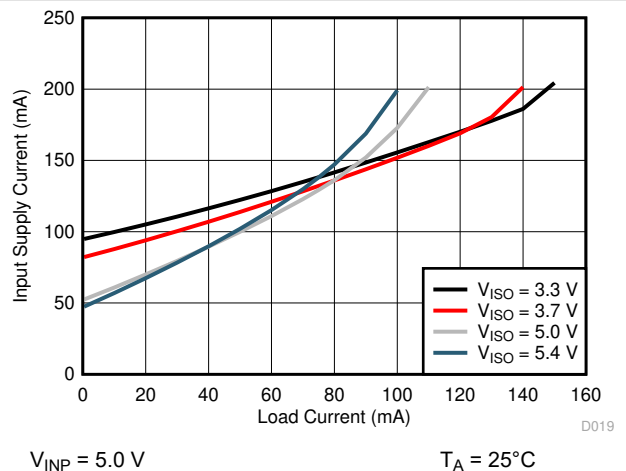


Figure 6-25. Input Supply Current ( $I_{VINP}$ ) vs Load Current ( $I_{ISO}$ )

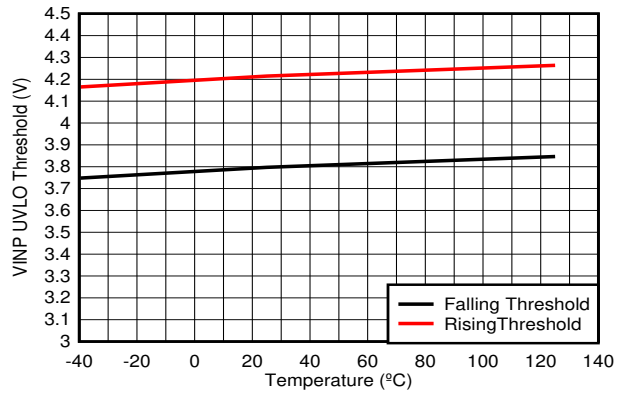


Figure 6-26. Typical  $V_{INP}$  UVLO Threshold vs Junction Temperature ( $T_J$ )

## 7 Detailed Description

### 7.1 Overview

The UCC12040 device integrates a high-efficiency, low-emissions isolated DC/DC converter. This approach provides typically 500 mW of clean, steady power across a 3000V<sub>RMS</sub> basic isolation barrier.

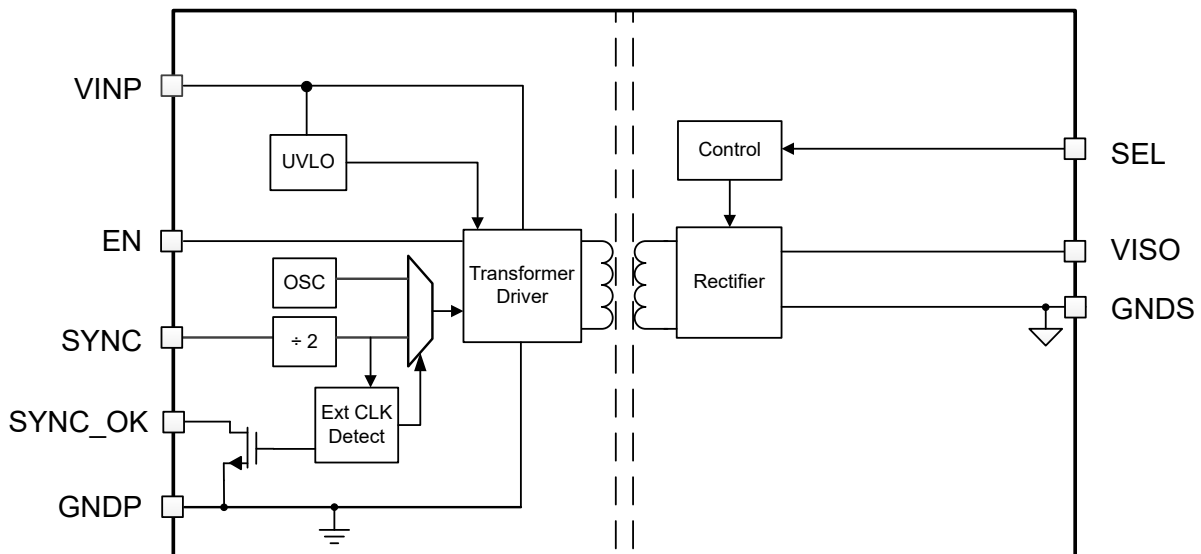
The integrated DC/DC converter uses switched mode operation and proprietary circuit techniques to reduce power losses and boost efficiency. Specialized control mechanisms, clocking schemes, and the use of an on-chip transformer provide high efficiency and low radiated emissions.

The VINP supply is provided to the primary power controller that switches the power stage connected to the integrated transformer. Power is transferred to the secondary side, rectified, and regulated to a level set by the SEL pin condition.

A fast feedback control loop monitors VISO and the output load, and ensures low overshoots and undershoots during load transients. Undervoltage lockout (UVLO) with hysteresis is integrated on the VINP supply, which ensures robust system performance under noisy conditions.

UCC12040 is suitable for applications that have limited board space and require more integration. These devices are also suitable for very-high voltage applications, where power transformers meeting the required isolation specifications are bulky and expensive.

### 7.2 Functional Block Diagram



### 7.3 Feature Description

#### 7.3.1 Enable and Disable

Forcing EN low disables the device, which greatly reduces the VINP power consumption. Pull the EN pin high to enable normal device functionality. The EN pin has a weak internal pull-down resistor, so the device floats to the disable state if the pin is left open.

#### 7.3.2 UVLO, Power-Up, and Power-Down Behavior

The UCC12040 has an undervoltage lockout (UVLO) on the VINP power supply. Upon power-up, while the VINP voltage is below the threshold voltage  $V_{UVPR}$ , the primary side transformer driver is disabled, and VISO output is off. The output powers up once the threshold is met. Likewise, if VINP falls below  $V_{UVPF}$ , the converter is disabled and there is no output at VISO. Both UVLO threshold voltages have hysteresis to avoid chattering.

### 7.3.3 $V_{ISO}$ Load Recommended Operating Area

Figure 7-1 depicts the device  $V_{ISO}$  regulation behavior across the output load range, including when the output is overloaded. For proper device operation, ensure that the device  $V_{ISO}$  output load does not exceed the maximum output current ( $I_{OUT\_MAX}$ ). The value for  $I_{OUT\_MAX}$  over different temperature and  $V_{INP}$  conditions are shown from Figure 6-8 to Figure 6-11. The following protection mechanisms will be engaged if the UCC12040 is loaded beyond the recommended operating area:

1. The device limits the maximum output power. If a load exceeding  $I_{OUT\_MAX}$  is applied,  $V_{ISO}$  drops accordingly to meet the maximum power limit.
2. If  $V_{ISO}$  drops below nominal 3.8 V while operating in the constant power limit region, the over-power fold-back feature will switch the power converter from active rectification to passive rectification, and the built-in recovery hysteresis will ensure the UCC12040 recovers at a lower output power. The device returns to active rectification when load drops and  $V_{ISO}$  increases above nominal 4.3V.
3. The device triggers a soft-start reset if  $V_{ISO}$  drops below the nominal 1.8-V threshold. This reset is designed to protect the device during  $V_{ISO}$  short-circuit conditions.
4. Thermal shutdown protection disables the converter if the device is operated in any of the above regions long enough to raise the silicon junction temperature above the thermal shutdown threshold. See the Section 7.3.4 section for more details on this device feature.

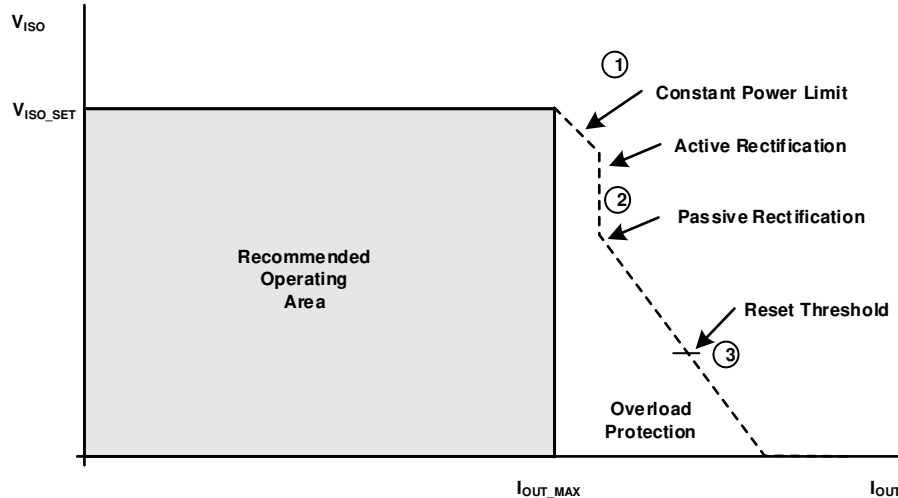


Figure 7-1.  $V_{ISO}$  Load Recommended Operating Area Description

### 7.3.4 Thermal Shutdown

Thermal protection is also integrated to help prevent the device from getting damaged during overload and short-circuit conditions on the isolated output. Under these conditions, the device temperature starts to increase. When the silicon junction temperature  $T_j$  sensed at the primary side die goes above the threshold  $TSD_{THR}$  (typical  $165^{\circ}C$ ), thermal shutdown activates and the primary controller turns off which removes the energy supplied to the  $V_{ISO}$  load, which causes the device to cool off. When the junction temperature drops approximately  $27^{\circ}C$  ( $TSD_{HYST}$ ) from the shutdown point, the device starts to function normally. If an overload or output short-circuit condition prevails, this protection cycle is repeated. Make sure the design prevents the device junction temperatures from reaching such high values.

### 7.3.5 External Clocking and Synchronization

The UCC12040 has an internal oscillator trimmed to drive the transformer at 8.0 MHz. An external clock may be applied at the SYNC pin to override the internal oscillator. This external clock will be divided by 2, so the target range for the external clock signal at SYNC is 16 MHz  $\pm 10\%$ . When a valid external clock signal is detected, the internal spread spectrum modulation (SSM) algorithm is disabled. This allows an external clock signal with a unique SSM to be applied. The depth and frequency of SSM is a tradeoff versus low frequency modulated  $V_{ISO}$  voltage ripple. The SYNC\_OK pin is asserted LOW if there is no external SYNC clock or one that is outside of



the operating range of the device is detected. In this state, the external clock is ignored and the DC/DC converter is clocked by the internal oscillator. The pin is in high impedance if a valid clock is applied on SYNC.

### 7.3.6 $V_{ISO}$ Output Voltage Selection

The SEL pin is monitored during power-up — within the first 1 ms after applying VINP above the UVLO rising threshold or enabling via the EN pin — to detect the desired regulation voltage for the VISO output. Note that after this initial monitoring, the SEL pin no longer affects the VISO output level. In order to change the output mode selection, either the EN pin must be toggled or the VINP power supply must be cycled off and back on. Section 6.4 provides more details on the SEL pin functionality.

### 7.3.7 Electromagnetic Compatibility (EMC) Considerations

UCC12040 devices use spread spectrum modulation algorithm for the internal oscillator and advanced internal layout scheme to minimize radiated emissions at the system level.

Many applications in harsh industrial environment are sensitive to disturbances such as electrostatic discharge (ESD), electrical fast transient (EFT), surge and electromagnetic emissions. These electromagnetic disturbances are regulated by international standards such as IEC 61000-4-x and CISPR 32. Although system-level performance and reliability depends, to a large extent, on the application board design and layout, the device incorporates many chip-level design improvements for overall system robustness.

## 7.4 Device Functional Modes

Table 7-1 lists the supply functional modes for this device.

**Table 7-1. Device Functional Modes**

INPUTS		Isolated Supply Output Voltage ( $V_{ISO}$ ) Setpoint
EN	SEL	
HIGH	Shorted to VISO	5.0 V
HIGH	100 k $\Omega$ to VISO	5.4 V
HIGH	Shorted to GNDS	3.3 V
HIGH	100 k $\Omega$ to GNDS	3.7 V
HIGH	OPEN <sup>(1)</sup>	UNSUPPORTED
LOW	X	0 V

(1) The SEL pin has an internal weak pull-down resistance to ground, but leaving this pin open is not recommended.

## 8 Application and Implementation

### Note

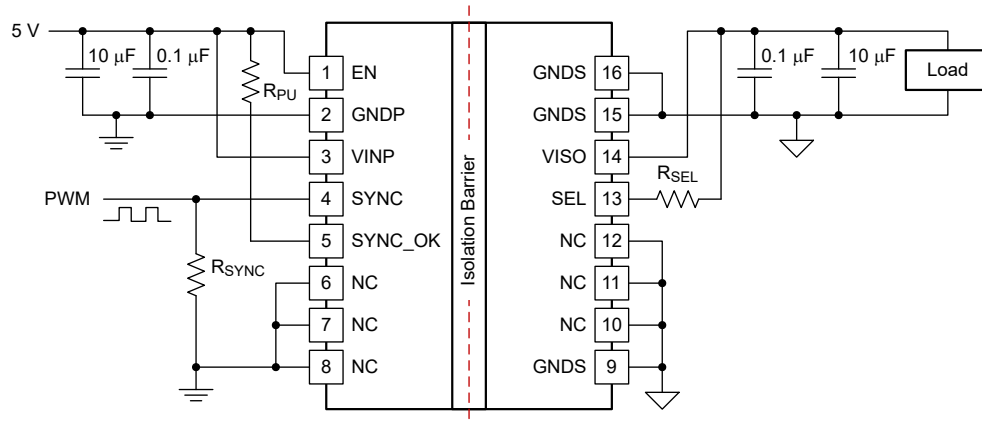
Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

### 8.1 Application Information

The UCC12040 device is suitable for applications that have limited board space and desire more integration. This device is also suitable for very high voltage applications, where power transformers meeting the required isolation specifications are bulky and expensive.

### 8.2 Typical Application

Figure 8-1 shows the typical application schematic for the UCC12040 device supplying an isolated load.



**Figure 8-1. Typical Application**

## 8.2.1 Design Requirements

To design using UCC12040, a few simple design considerations must be evaluated. [Table 8-1](#) shows some recommended values for a typical application. See [Section 9](#) and [Section 10](#) sections to review other key design considerations for the UCC12040.

**Table 8-1. Design Parameters**

PARAMETER	RECOMMENDED VALUE
Input supply voltage, $V_{INP}$	4.5 V to 5.5 V
Decoupling capacitance between $V_{INP}$ and GNDP	10 $\mu$ F, 16 V, $\pm$ 10%, X7R
Decoupling capacitance between $V_{ISO}$ and GNDS <sup>(1)</sup>	10 $\mu$ F, 16 V, $\pm$ 10%, X7R
Optional additional capacitance on $V_{ISO}$ or $V_{INP}$ to reduce high-frequency ripple	0.1 $\mu$ F, 50 V, $\pm$ 10%, X7R
Pull-up resistor from SYNC_OK to $V_{INP}$ , $R_{PU}$	100 k $\Omega$
Pull-up resistor from SEL to $V_{ISO}$ for 5.0V output voltage mode, $R_{SEL}$	0 $\Omega$
Pull-up resistor from SEL to $V_{ISO}$ for 5.4V output voltage mode, $R_{SEL}$	100 k $\Omega$
Optional SYNC signal impedance-matching resistor, $R_{SYNC}$	Match source — typical values are 50 $\Omega$ , 75 $\Omega$ , 100 $\Omega$ , or 1 k $\Omega$
External clock signal applied on SYNC	16 MHz

(1) See [Section 8.2.2.1](#) section.

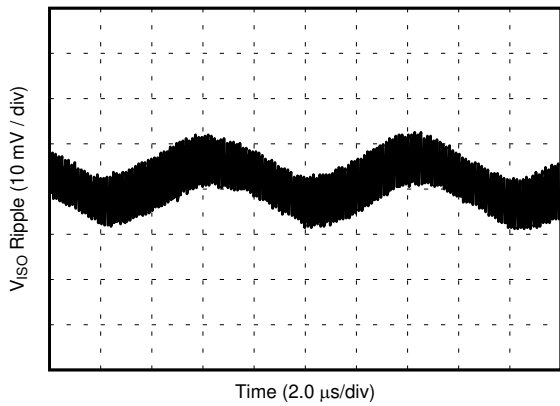
## 8.2.2 Detailed Design Procedure

Place ceramic decoupling capacitors as close as possible to the device pins. For the input supply, place the capacitor(s) between pin 3 ( $V_{INP}$ ) and pin 2 (GNDP). For the isolated output supply, place the capacitor(s) between pin 14 ( $V_{ISO}$ ) and pin 15 (GNDS). This location is of particular importance to the input decoupling capacitor, because this capacitor supplies the transient current associated with the fast switching waveforms of the power drive circuits. The recommended capacitor value is 10  $\mu$ F. Ensure the capacitor dielectric material is compatible with the target application temperature.

### 8.2.2.1 $V_{ISO}$ Output Capacitor Selection

The UCC12040 is optimized to run with an effective output capacitance of 5  $\mu$ F to 20  $\mu$ F. A ceramic capacitor is recommended. Ceramic capacitors have DC-Bias and temperature derating effects, which both have influence the final effective capacitance. Choose the right capacitor carefully in combination with considering its package size, dielectric and voltage rating. It is good design practice to include one 0.1- $\mu$ F capacitor close to the device for high-frequency noise reduction.

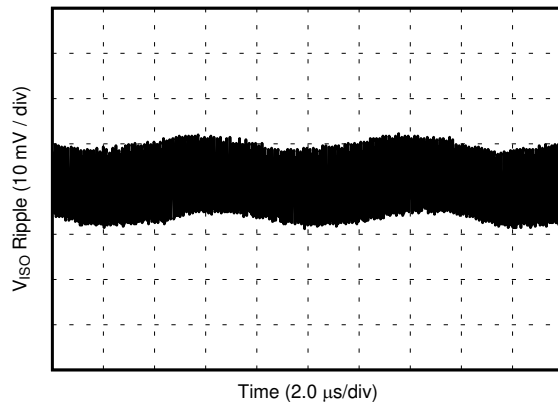
### 8.2.3 Application Curves



D033

$V_{INP} = 5.0\text{ V}$     $V_{ISO} = 5.4\text{ V}$    Bandwidth = 20 MHz

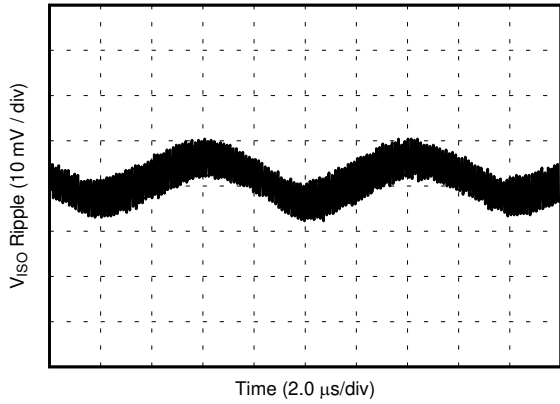
**Figure 8-2.  $V_{ISO}$  Ripple, 5.4-V Output, 10% Load**



D034

$V_{INP} = 5.0\text{ V}$     $V_{ISO} = 5.4\text{ V}$    Bandwidth = 20 MHz

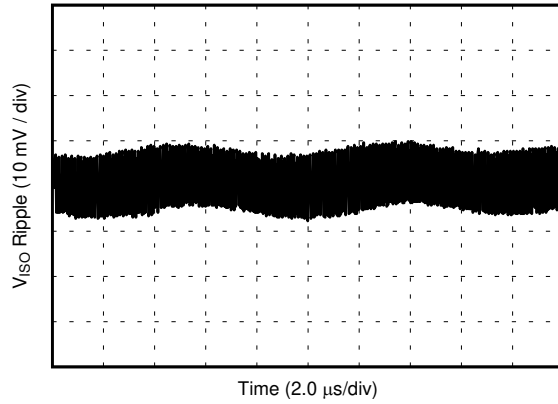
**Figure 8-3.  $V_{ISO}$  Ripple, 5.4-V Output, 90% Load**



D035

$V_{INP} = 5.0\text{ V}$     $V_{ISO} = 5.0\text{ V}$    Bandwidth = 20 MHz

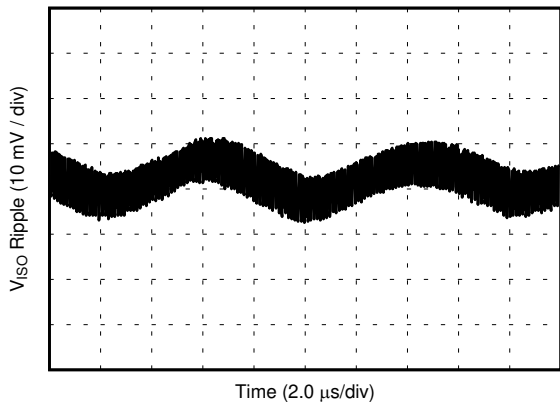
**Figure 8-4.  $V_{ISO}$  Ripple, 5.0-V Output, 10% Load**



D036

$V_{INP} = 5.0\text{ V}$     $V_{ISO} = 5.0\text{ V}$    Bandwidth = 20 MHz

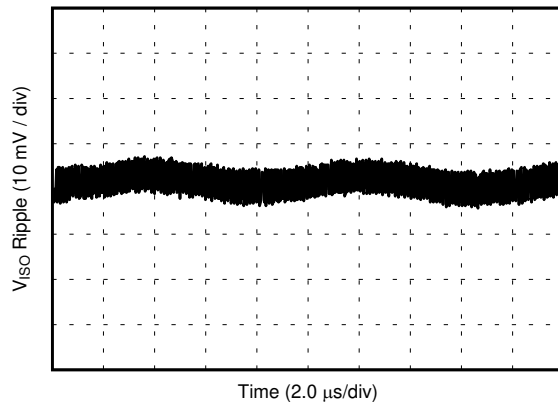
**Figure 8-5.  $V_{ISO}$  Ripple, 5.0-V Output, 90% Load**



D037

$V_{INP} = 5.0\text{ V}$     $V_{ISO} = 3.7\text{ V}$    Bandwidth = 20 MHz

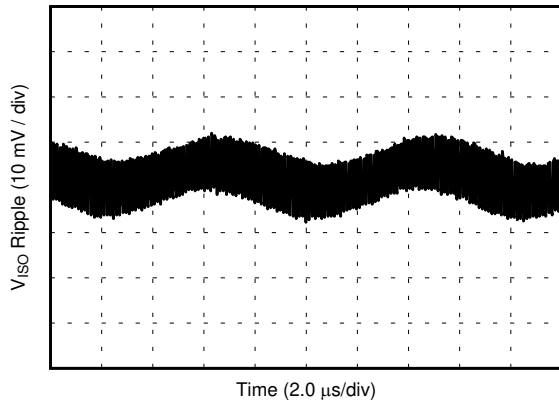
**Figure 8-6.  $V_{ISO}$  Ripple, 3.7-V Output, 10% Load**



D038

$V_{INP} = 5.0\text{ V}$     $V_{ISO} = 3.7\text{ V}$    Bandwidth = 20 MHz

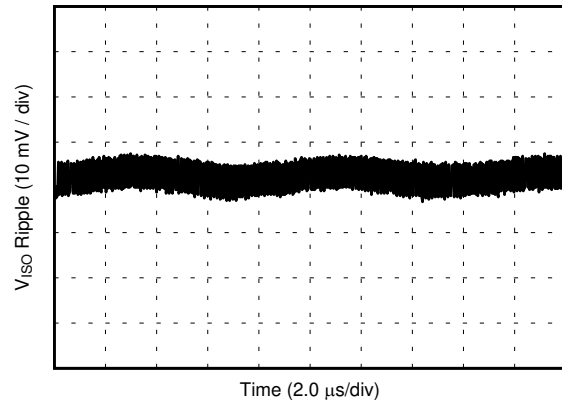
**Figure 8-7.  $V_{ISO}$  Ripple, 3.7-V Output, 90% Load**



D039

$V_{INP} = 5.0\text{ V}$     $V_{ISO} = 3.3\text{ V}$    Bandwidth = 20 MHz

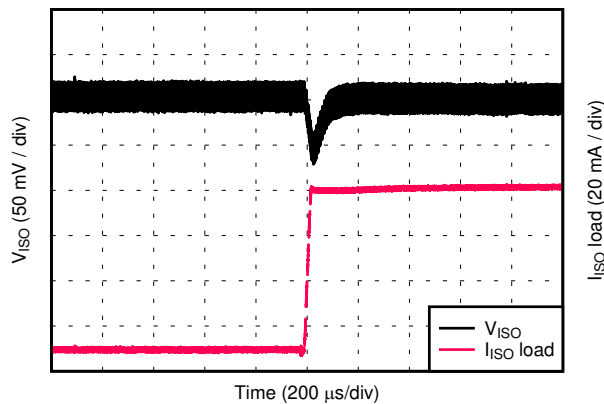
**Figure 8-8.  $V_{ISO}$  Ripple, 3.3-V Output, 10% Load**



D040

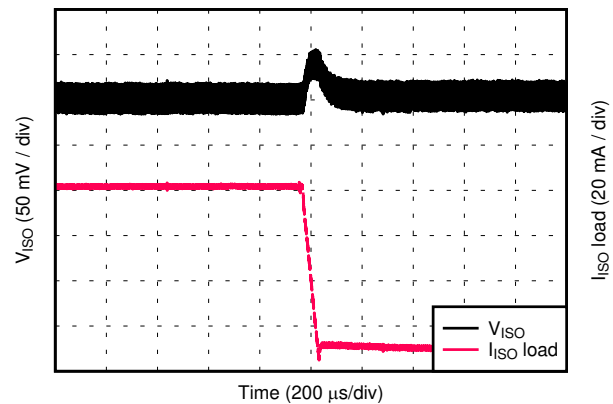
$V_{INP} = 5.0\text{ V}$     $V_{ISO} = 3.3\text{ V}$    Bandwidth = 20 MHz

**Figure 8-9.  $V_{ISO}$  Ripple, 3.3-V Output, 90% Load**



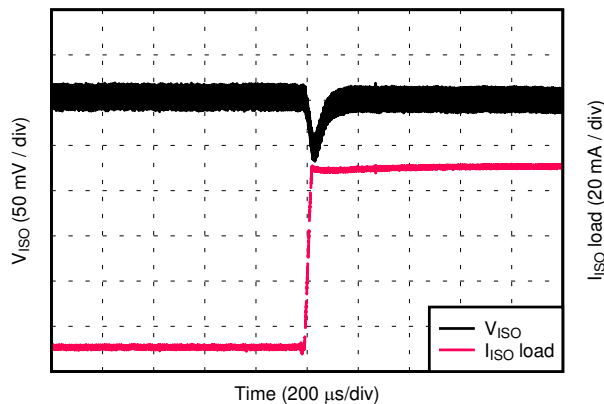
D041

**Figure 8-10.  $V_{ISO}$  Load Transient Response, 10% to 90% Load Step, 5.0-V Input, 5.4-V Output**



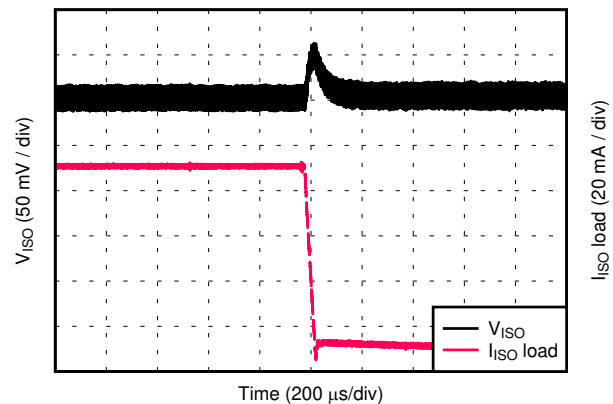
D042

**Figure 8-11.  $V_{ISO}$  Load Transient Response, 90% to 10% Load Step, 5.0-V Input, 5.4-V Output**



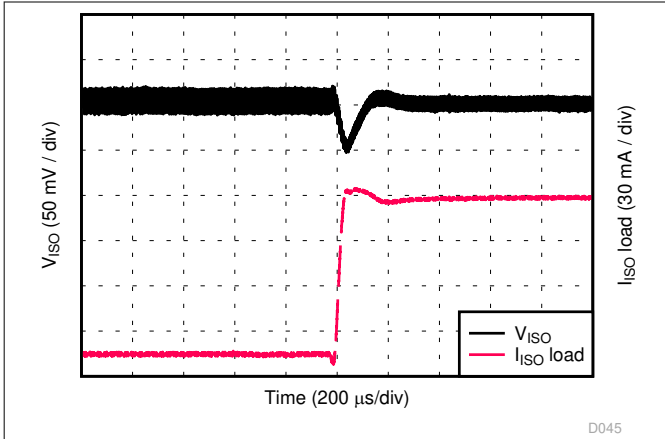
D043

**Figure 8-12.  $V_{ISO}$  Load Transient Response, 10% to 90% Load Step, 5.0-V Input, 5.0-V Output**

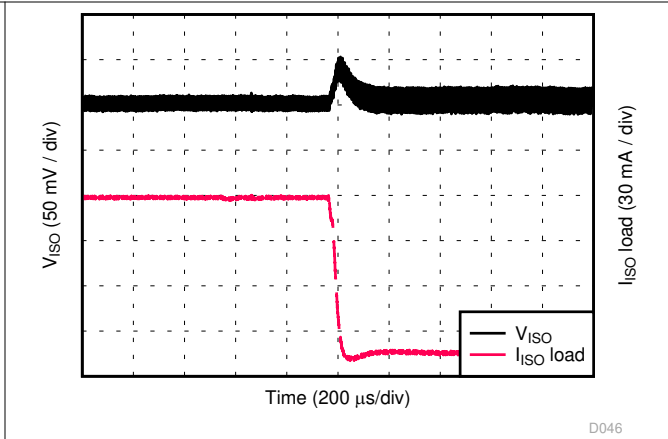


D044

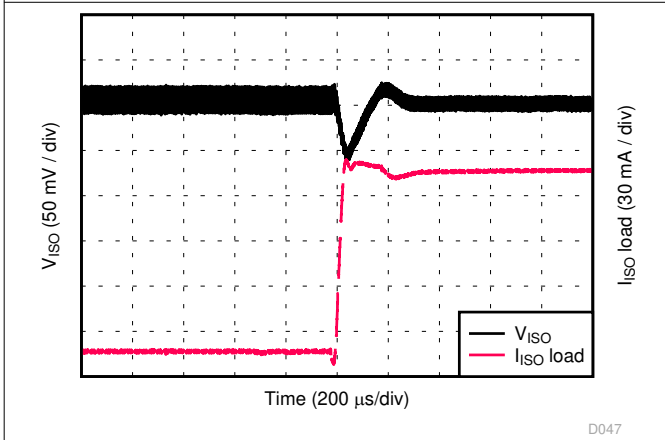
**Figure 8-13.  $V_{ISO}$  Load Transient Response, 90% to 10% Load Step, 5.0-V Input, 5.0-V Output**



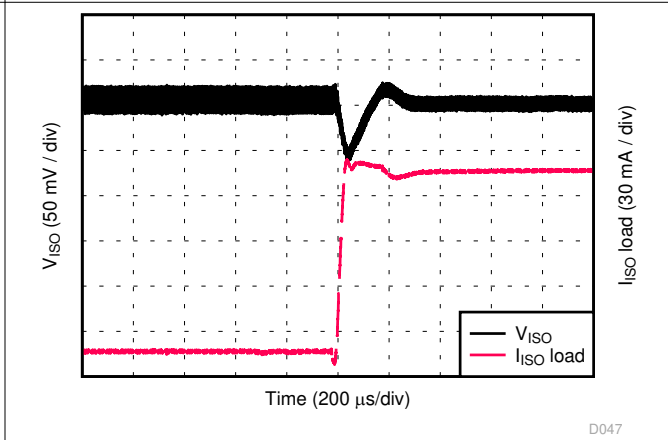
**Figure 8-14.  $V_{ISO}$  Load Transient Response, 10% to 90% Load Step, 5.0-V Input, 3.7-V Output**



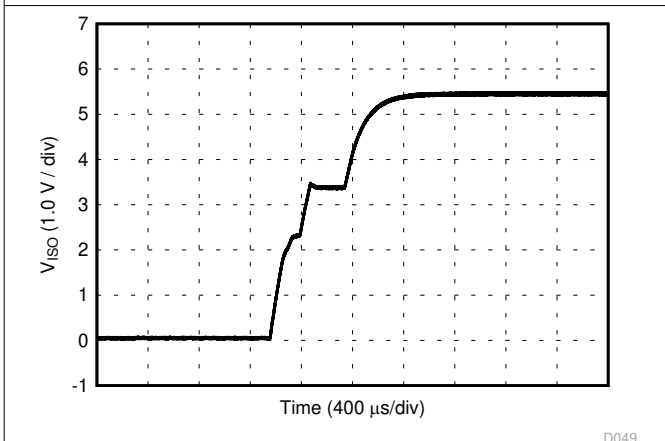
**Figure 8-15.  $V_{ISO}$  Load Transient Response, 90% to 10% Load Step, 5.0-V Input, 3.7-V Output**



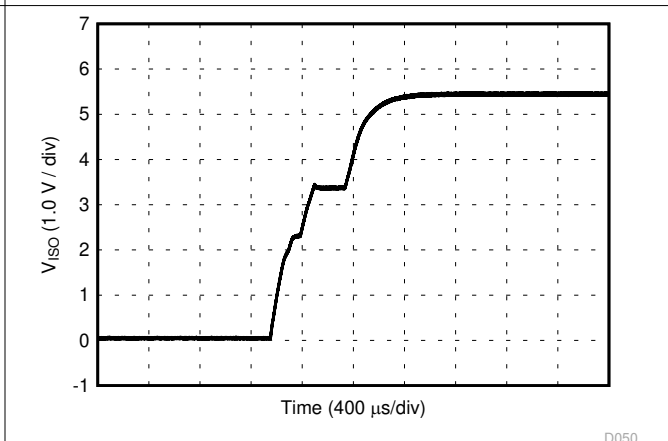
**Figure 8-16.  $V_{ISO}$  Load Transient Response, 10% to 90% Load Step, 5.0-V Input, 3.3-V Output**



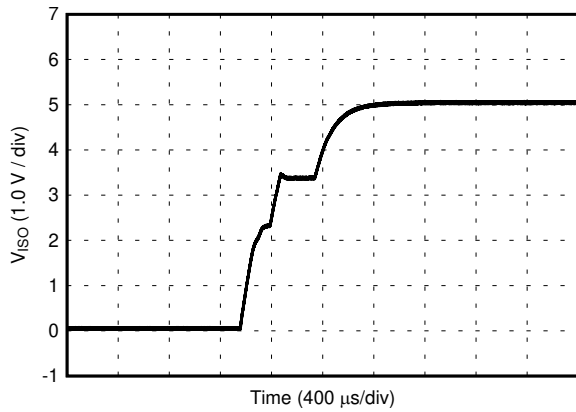
**Figure 8-17.  $V_{ISO}$  Load Transient Response, 90% to 10% Load Step, 5.0-V Input, 3.3-V Output**



**Figure 8-18.  $V_{ISO}$  Soft Start at 10% Rated Load, 5.0-V Input, 5.4-V Output**

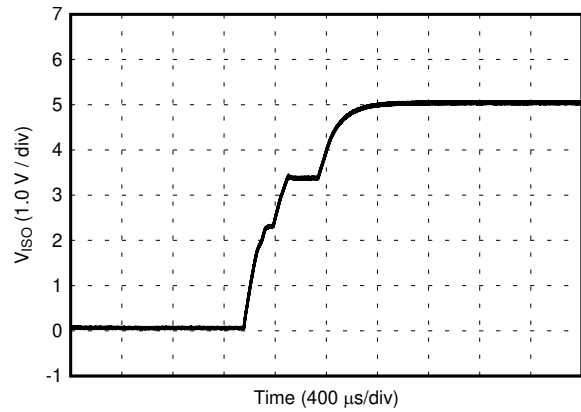


**Figure 8-19.  $V_{ISO}$  Soft Start at 90% Rated Load, 5.0-V Input, 5.4-V Output**



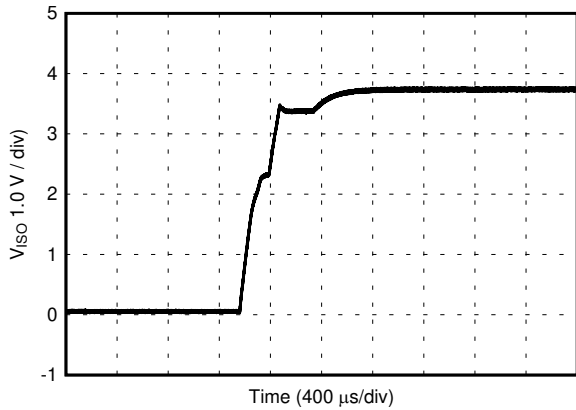
D051

**Figure 8-20.  $V_{ISO}$  Soft Start at 10% Rated Load, 5.0-V Input, 5.0-V Output**



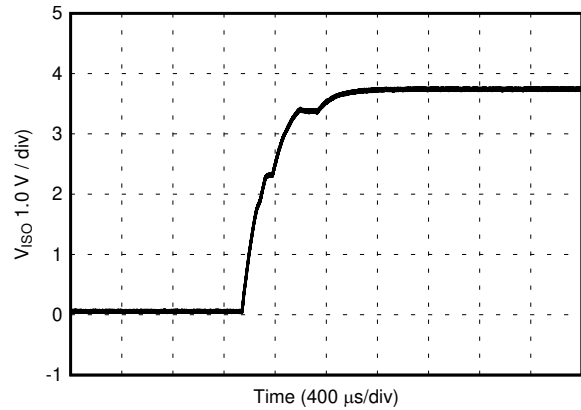
D052

**Figure 8-21.  $V_{ISO}$  Soft Start at 90% Rated Load, 5.0-V Input, 5.0-V Output**



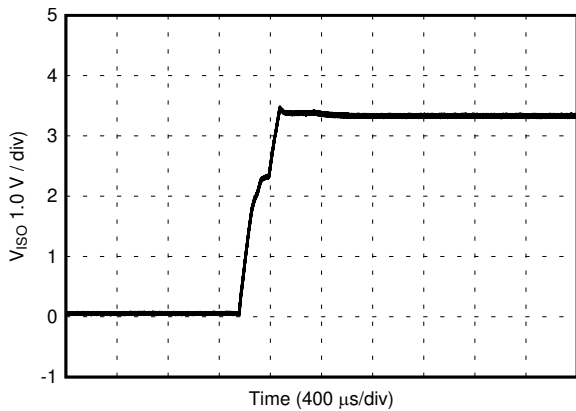
D053

**Figure 8-22.  $V_{ISO}$  Soft Start at 10% Rated Load, 5.0-V Input, 3.7-V Output**



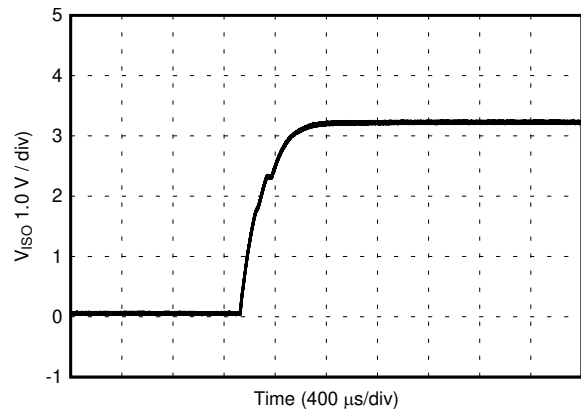
D054

**Figure 8-23.  $V_{ISO}$  Soft Start at 90% Rated Load, 5.0-V Input, 3.7-V Output**



D055

**Figure 8-24.  $V_{ISO}$  Soft Start at 10% Rated Load, 5.0-V Input, 3.3-V Output**



D056

**Figure 8-25.  $V_{ISO}$  Soft Start at 90% Rated Load, 5.0-V Input, 3.3-V Output**



## 9 Power Supply Recommendations

The recommended input supply voltage (VINP) for the UCC12040 is between 4.5 V and 5.5 V. To help ensure reliable operation, adequate decoupling capacitors must be located as close to supply pins as possible. Place local bypass capacitors between the VINP and GNDP pins at the input, and between VISO and GNDS at the isolated output supply. Low ESR, ceramic surface mount capacitors are recommended. It is further suggested that one place two such capacitors: one with a value of 10  $\mu$ F for supply bypassing, and an additional 100-nF capacitor in parallel for high frequency filtering. The input supply must have an appropriate current rating to support output load required by the end application.

## 10 Layout

### 10.1 Layout Guidelines

The UCC12040 integrated isolated power solution simplifies system design and reduces board area usage. Proper PCB layout is important in order to achieve optimum performance. Here is a list of recommendations:

1. Place decoupling capacitors as close as possible to the device pins. For the input supply, place the capacitor(s) between pin 3 (VINP) and pin 2 (GNDP). For the isolated output supply, place the capacitor(s) between pin 14 (VISO) and pin 15 (GNDS). This location is of particular importance to the input decoupling capacitor, because this capacitor supplies the transient current associated with the fast switching waveforms of the power drive circuits.
2. Because the device does not have a thermal pad for heat-sinking, the device dissipates heat through the respective GND pins. Ensure that enough copper (preferably a connection to the ground plane) is present on all GNDP and GNDS pins for best heat-sinking.
3. If space and layer count allow, it is also recommended to connect the VINP, GNDP, VISO and GNDS pins to internal ground or power planes through multiple vias of adequate size. Alternatively, make traces for these nets as wide as possible to minimize losses.
4. TI also recommends grounding the no-connect pins (NC) to their respective ground planes. For pins 6, 7, and 8, connect to GNDP. For pins 10, 11, and 12, connect to GNDS. This will allow more continuous ground planes and larger thermal mass for heat-sinking.
5. A minimum of four layers is recommended to accomplish a low-EMI PCB design. Inner layers can be spaced closer than outer layers and used to create a high-frequency bypass capacitor between GNDP and GNDS to reduce radiated emissions. Ensure proper spacing, both inter-layer and layer-to-layer, is implemented to avoid reducing isolation capabilities. These spacings will vary based on the printed circuit board construction parameters, such as dielectric material and thickness.
6. Pay close attention to the spacing between primary ground plane (GNDP) and secondary ground plane (GNDS) on the PCB outer layers. The effective creepage and or clearance of the system will be reduced if the two ground planes have a lower spacing than that of the device package.
7. To ensure isolation performance between the primary and secondary side, avoid placing any PCB traces or copper below the UCC12040 device on the outer copper layers.

### 10.2 Layout Example

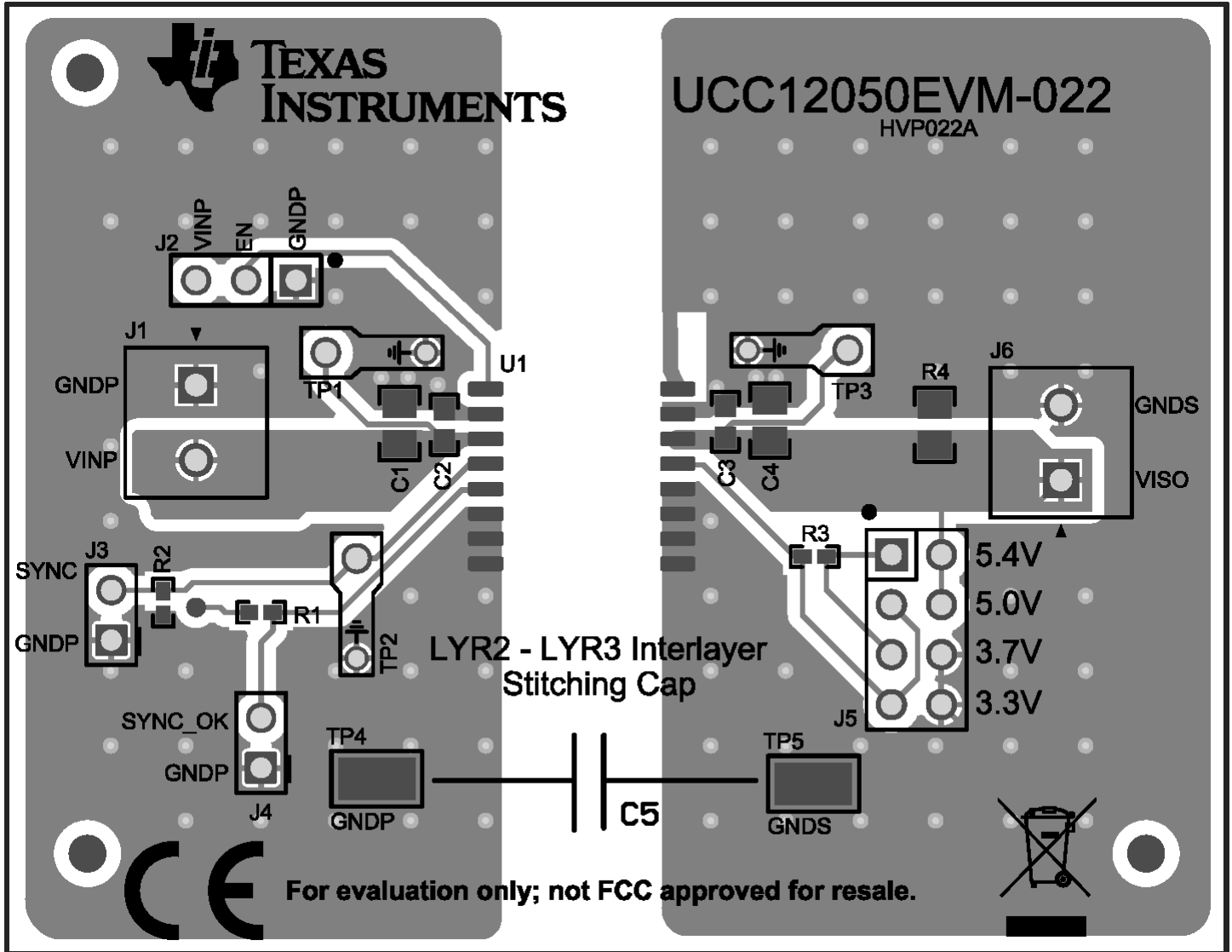


Figure 10-1. Layout Example

## 11 Device and Documentation Support

### 11.1 Device Support

#### 11.1.1 Development Support

For development support, refer to:

- [High-efficiency, low-emission, isolated DC/DC converter-based analog input module reference design](#)
- [Isolated delta-sigma modulator based AC/DC voltage and current measurement module reference design](#)
- [Isolated power architecture reference design for communication and analog input/output modules](#)

### 11.2 Documentation Support

#### 11.2.1 Related Documentation

For related documentation see the following:

- [UCC12050 Evaluation Module User Guide](#)
- [Isolation Glossary](#)
- [A Reinforced-isolated Analog Input Chain for Space-constrained Applications](#)

### 11.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on [ti.com](#). Click on *Subscribe to updates* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

### 11.4 Support Resources

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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### 11.5 Trademarks

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### 11.6 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

### 11.7 Glossary

[TI Glossary](#) This glossary lists and explains terms, acronyms, and definitions.

## 12 Mechanical and Packaging Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
UCC12040DVE	ACTIVE	SO-MOD	DVE	16	40	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 125	UCC12040	<a href="#">Samples</a>
UCC12040DVER	ACTIVE	SO-MOD	DVE	16	2000	RoHS & Green	NIPDAU	Level-3-260C-168 HR	-40 to 125	UCC12040	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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**TAPE AND REEL INFORMATION**

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
UCC12040DVER	SO-MOD	DVE	16	2000	330.0	16.4	10.75	10.7	2.7	12.0	16.0	Q1

**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
UCC12040DVER	SO-MOD	DVE	16	2000	350.0	350.0	43.0

**TUBE**


\*All dimensions are nominal

Device	Package Name	Package Type	Pins	SPQ	L (mm)	W (mm)	T (μm)	B (mm)
UCC12040DVE	DVE	SO-MOD	16	40	506.98	12.7	4826	6.6



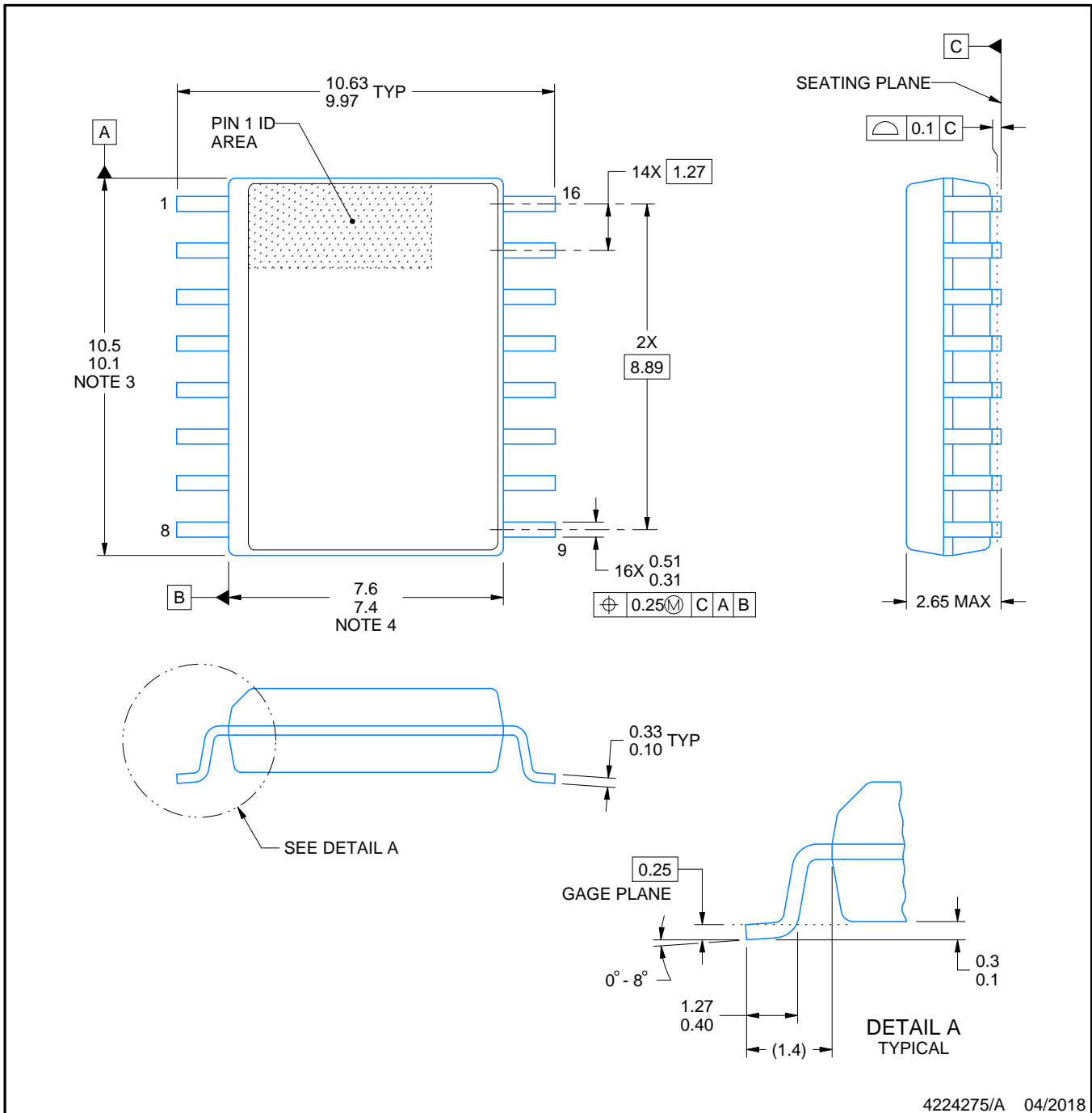


# PACKAGE OUTLINE

## DVE0016A

### SO-MOD - 2.65 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



4224275/A 04/2018

#### NOTES:

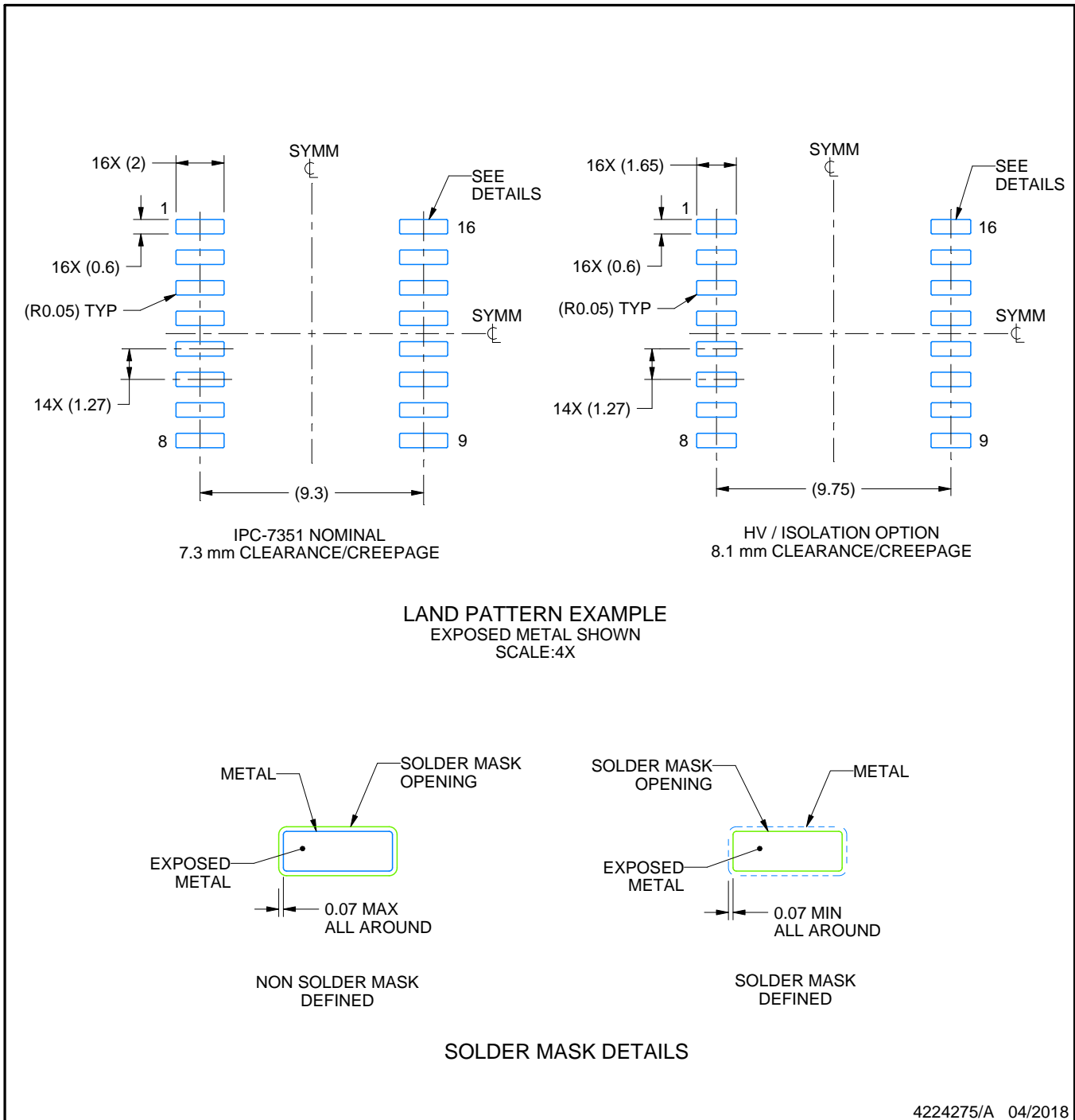
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm, per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm, per side.
5. Reference JEDEC registration MS-013.

# EXAMPLE BOARD LAYOUT

DVE0016A

SO-MOD - 2.65 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

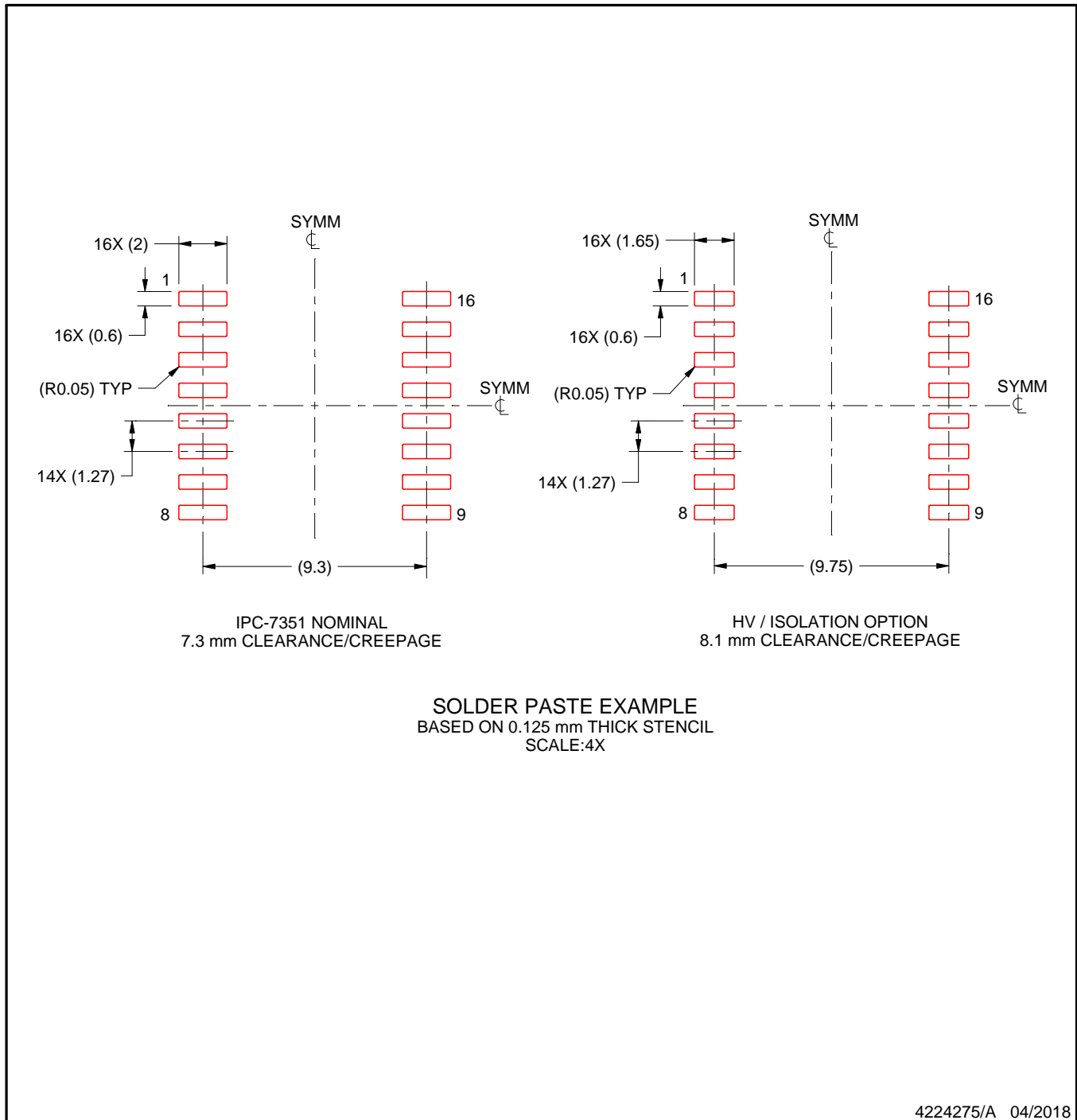
- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DVE0016A

SO-MOD - 2.65 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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