

## 5.5V, 1MHz, 2A Synchronous Step-Down Converter

### FEATURES

- High Efficiency: Up to 95% (@3.3V<sub>OUT</sub>)
- 1MHz Constant Frequency Operation
- 2A Output Current
- 2.5V to 5.5V Input Voltage Range
- Output Voltage as Low as 0.6V
- PFM Mode for High Efficiency in Light Load
- 100% Duty Cycle in Dropout Operation
- Low Quiescent Current: 40μA
- Short Hiccup Protection
- Thermal Fault Protection
- Inrush Current Limit and Soft Start
- Input over voltage protection (OVP)
- <1μA Shutdown Current
- SOT23-5 Package

### APPLICATIONS

- Set Top Box
- Wireless and DSL Modems
- Portable Instruments
- Digital Still and Video Cameras
- PC Cards

### GENERAL DESCRIPTION

The TMI3410 is a 1MHz constant frequency, current mode step-down converter. It is ideal for portable equipment requiring very high current up to 2A from single-cell Lithium-ion batteries or other input source from 2.5V to 5.5V input voltage and the output voltage can be regulated as low as 0.6V. The TMI3410 also can run at 100% duty cycle for low dropout operation, extending battery life in portable systems while light load operation provides very low output ripple for noise sensitive applications. The high switching frequency minimizes the size of external components while keeping switching losses low. The internal slope compensation setting allows the device to operate with smaller inductor values to optimize size and provide efficient operation. The TMI3410 is offered in a 5-pin, SOT package, and is available in an adjustable version. This device offers two operation modes, PWM control and PFM Mode switching control, which allows a high efficiency over the wider range of the load.

### TYPICAL APPLICATION

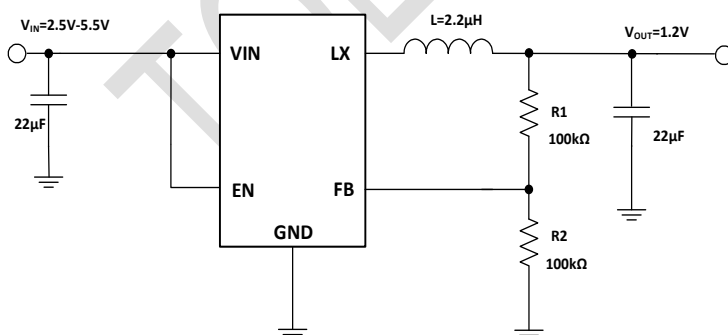
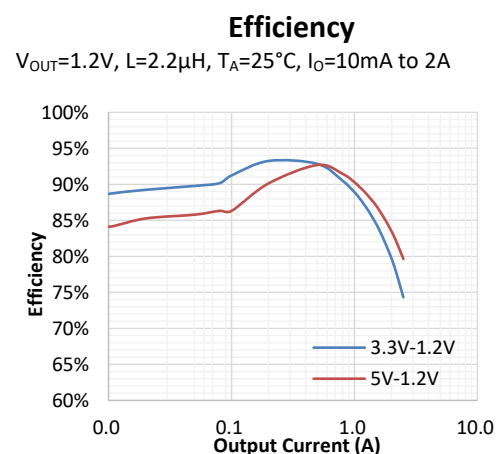


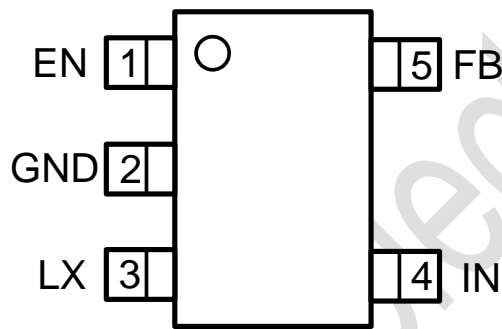
Figure 1. Basic Application Circuit



## ABSOLUTE MAXIMUM RATINGS (Note 1)

Parameter	Min	Max	Unit
Input Supply Voltage	-0.3	6.0	V
LX Voltages	-0.3	6.0	V
EN, FB Voltage	-0.3	6.0	V
Storage Temperature Range	-65	150	°C
Junction Temperature (Note2)	-	155	°C
Power Dissipation	-	600	mW
Lead Temperature (Soldering,10s)	-	260	°C

## PACKAGE/ORDER INFORMATION



SOT23-5

Top Mark: S15BXXX (S15B: Device Code, XXX: Inside Code)

Part Number	Package	Top Mark	Quantity/ Reel
TMI3410	SOT23-5	S15BXXX	3000

TMI3410 devices are Pb-free and RoHS compliant.

**PIN DESCRIPTION**

Pin	Name	Function
1	EN	Chip Enable Pin. Drive EN above 1.5V to turn on the part. Drive EN below 0.4V to turn it off. Do not leave EN floating.
2	GND	Ground pin.
3	LX	Power Switch Output. It is the switch node connection to Inductor. This pin connects to the drains of the internal P-ch and N-ch MOSFET switches.
4	IN	Power supply input pin.
5	FB	Output Voltage Feedback Pin.

**ESD RATING**

Items	Description	Value	Unit
$V_{ESD\_HBM}$	Human Body Model for all pins	$\pm 2000$	V
$V_{ESD\_CDM}$	Charge Device Model for all pins	$\pm 1000$	V

JEDEC specification JS-001

**RECOMMENDED OPERATING CONDITIONS**

Items	Description	Min	Max	Unit
Voltage Range	IN	2.5	5.5	V
$T_J$	Operating Junction Temperature Range	-40	125	°C

**THERMAL RESISTANCE** (Note 3)

Items	Description	Value	Unit
$\theta_{JA}$	Junction-to-ambient thermal resistance	200	°C/W
$\theta_{JC}$	Junction-to-case thermal resistance	62	°C/W

## ELECTRICAL CHARACTERISTICS

( $V_{IN}=V_{EN}=3.6V$ ,  $V_{OUT}=1.8V$ ,  $T_A = 25^{\circ}C$ , unless otherwise noted.)

Parameter	Test Conditions	Min	Typ	Max	Unit
Input Voltage Range		2.5		5.5	V
Input OVP Threshold	$V_{IN}$ rising		5.9	6.0	V
UVLO Threshold	$V_{IN}$ rising		2.35		V
UVLO Hysteresis			0.4		V
Quiescent Current	$V_{EN}=2.0V$ , $I_{OUT}=0A$ , $V_{FB}=V_{REF} \times 105\%$		40	100	$\mu A$
Shutdown Current	$V_{EN}=0V$		0.2	1.0	$\mu A$
Feedback Voltage Accuracy	$T_A = 25^{\circ}C$ , PWM Operation	0.588	0.600	0.612	V
Oscillation Frequency	$V_{OUT}=100\%$		1.0		MHz
	$V_{OUT}=0V$ , During Hiccup Mode		350		kHz
On Resistance of PMOS	$I_{LX}=100mA$		120		m $\Omega$
On Resistance of NMOS	$I_{LX}=-100mA$		70		m $\Omega$
Peak Current Limit	$V_{IN}=5V$ , $V_{OUT}=90\%$		2.5		A
EN High Level Input Voltage		1.5			V
EN Low Level Input Voltage				0.4	V
EN Leakage Current				1.0	$\mu A$
LX Leakage Current	$V_{EN}=0V$ , $V_{IN}=V_{LX}=5V$			1.0	$\mu A$
Thermal Shutdown Threshold <small>(Note 4)</small>			150		$^{\circ}C$
Thermal Shutdown Hysteresis <small>(Note 4)</small>			20		$^{\circ}C$

**Note 1:** Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

**Note 2:**  $T_J$  is calculated from the ambient temperature  $T_A$  and power dissipation  $P_D$  according to the following formula:  $T_J = T_A + (P_D) \times \theta_{JA}$ .

**Note 3:** Measured on JESD51-7, 4-layer PCB.

**Note 4:** Thermal shutdown threshold and hysteresis are guaranteed by design.

## FUNCTION DESCRIPTION

### Overview

The TMI3410 is a high output current switch mode step-down DC-DC converter. The device operates at a fixed 1MHz switching frequency, and uses a slope compensated current mode architecture.

This step-down DC-DC converter can supply up to 2A output current at  $V_{IN}=5V$  and has an input voltage range from 2.5V to 5.5V. It minimizes external component size and optimizes efficiency at the heavy load range. The slope compensation allows the device to remain stable over a wider range of inductor values so that smaller values with lower DCR can be used to achieve higher efficiency. Only a small bypass input capacitor is required at the output.

In light and no-load condition, TMI3410 are operating in PFM mode for power saving. In PFM mode, the device ramps up its output voltage with several SW switching pulse, while the error amplifier output voltage  $V_{COMP}$  drops. The device stops switching when  $V_{COMP}$  voltage drops down the inner threshold, so the FB voltage in PFM mode is a little bit higher than normal 0.6V reference voltage in PWM operation.

The adjustable output voltage can be programmed with external feedback dividers, ranging from 0.6V to near the input voltage. It uses internal MOSFETs to achieve high efficiency and can generate very low output voltages by using an internal reference of 0.6V. At dropout operation, the converter duty cycle increases to 100% and the output voltage tracks the input voltage minus the low  $R_{DS(ON)}$  drop of the P-channel high-side MOSFET and the inductor DCR. The internal error amplifier and compensation provides excellent transient response, load and line regulation. Internal soft start eliminates any output voltage overshoot when the device is enabled or the input voltage is applied.

### Input Over Voltage Protection

TMI3410 has input side over voltage protection function. When input voltage is higher than input OVP threshold 5.9V typical, TMI3410 stops switching operation to protect device works with high input voltage. When input voltage is recovered from OVP and drops down input OVP threshold with OVP hysteresis typical 140mV, the device starts to switch as normal operation automatically. This function protects device from switching in abnormal high input voltage and input surge condition.

### Input Under Voltage Lockout

TMI3410 implements input under voltage lockout function to avoid mis-operation at low input voltages. When the input voltage is lower than input UVLO threshold with UVLO hysteresis, the device is shut down. The typical 400mV input UVLO hysteresis value of TMI3410 is useful to prevent device from abnormal switching caused by input voltage oscillation around UVLO threshold during input voltage power-up and power-down with high load condition.

### Soft Start

TMI3410 has built-in soft-start circuits to control output voltage rise rate to avoids excessive inrush current during IC start up. The typical soft-start time is 0.8ms.

## Over Current Limit and Output Short Protection

TMI3410 has high side switching current limit function and prevents the device from high load current condition. The typical high side peak current limit value is 2.7A. When output load current increases and inductor current peak value reaches peak current limit value, high side MOSFET is turned off immediately and the output voltage drops down according to load condition. If output voltage keeps falling down, once the  $V_{FB}$  voltage is lower than 200mV typical, the device enters into output short hiccup protection condition in order to reduce power consumption and device thermal rise in the condition of output short to GND. In output short hiccup protection condition, the device hiccup cycle period is 16ms typical and the switching operation time during hiccup mode is 2ms. the switching frequency during hiccup mode is 350kHz.

## Thermal Shutdown

TMI3410 enters into thermal shutdown once the junction temperature exceeds thermal shutdown threshold 150°C typically. Once the device junction temperature falls below the threshold with hysteresis, TMI3410 returns to normal operation automatically.

## FUNCTIONAL BLOCK DIAGRAM

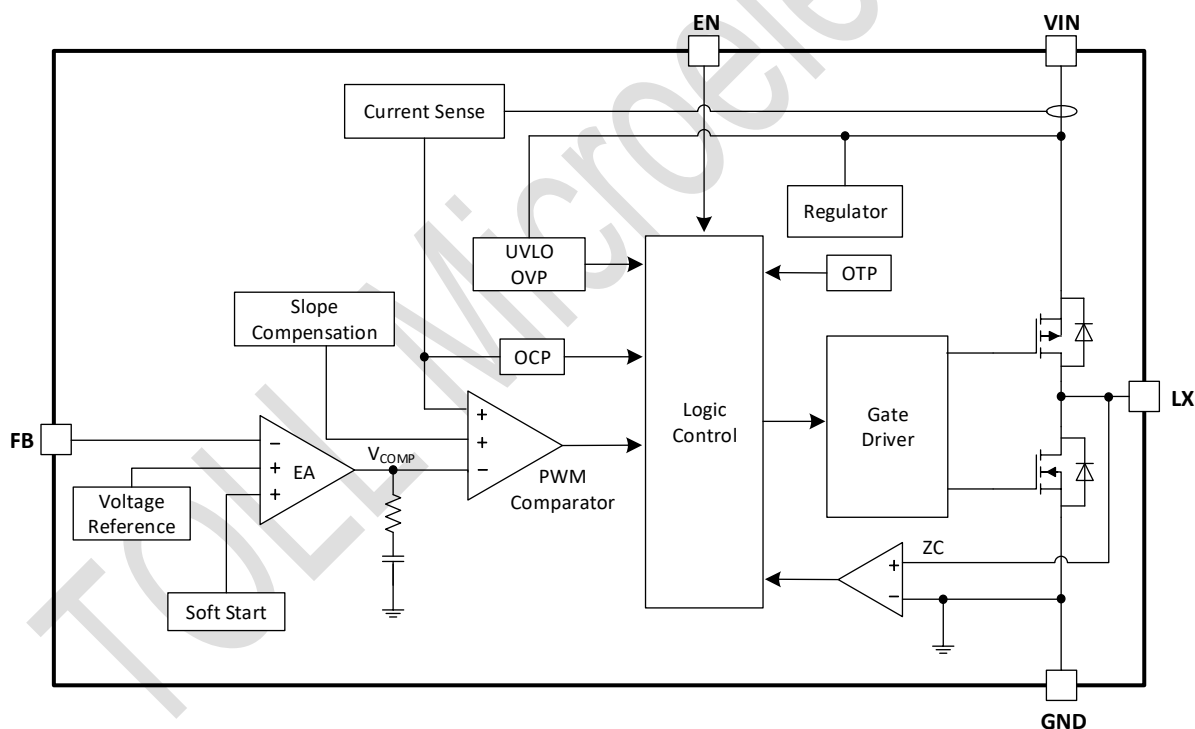
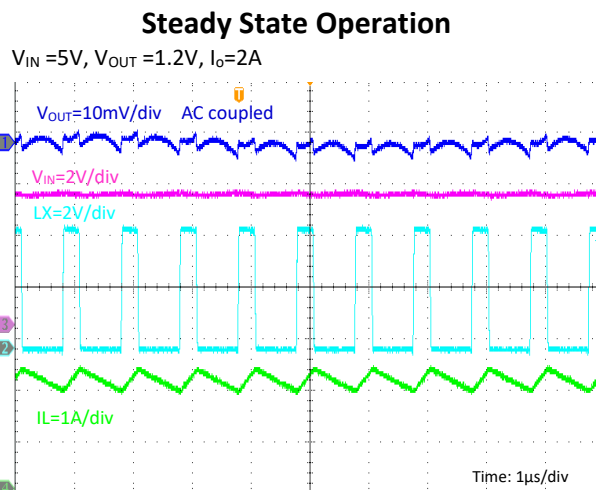
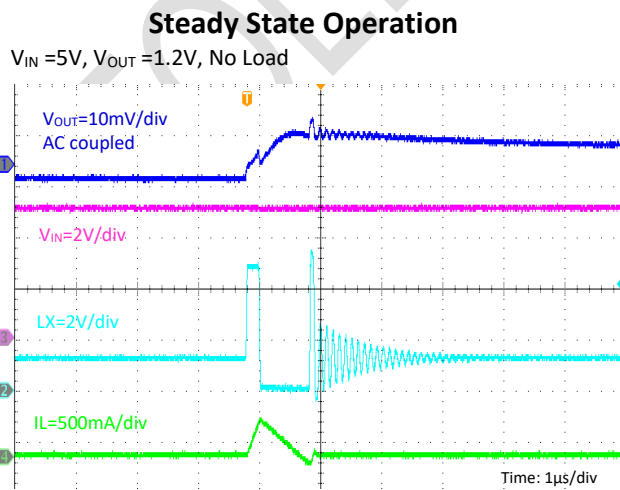
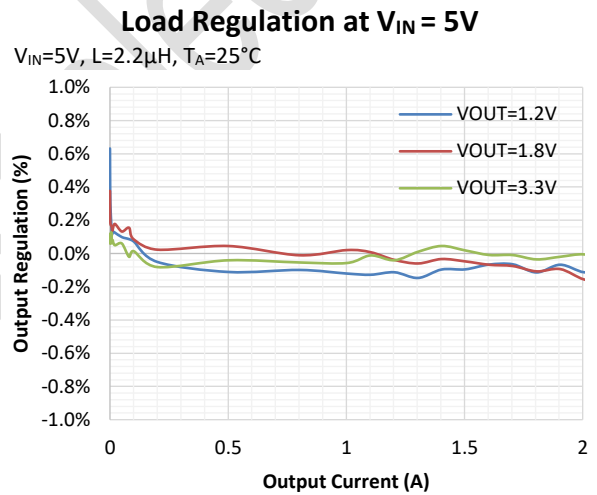
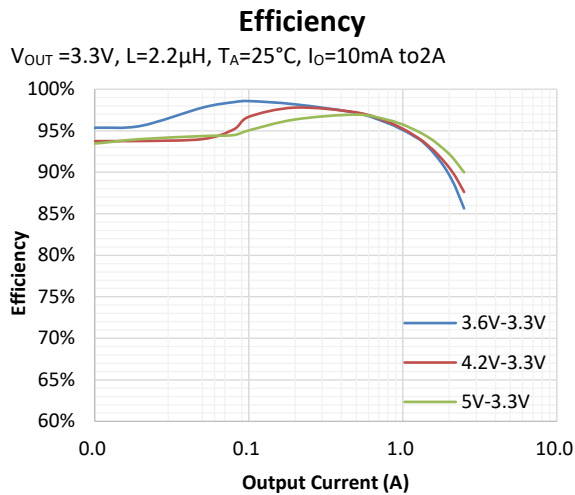
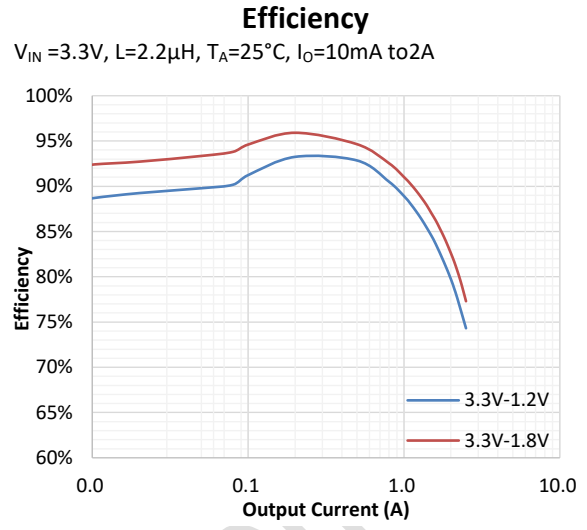
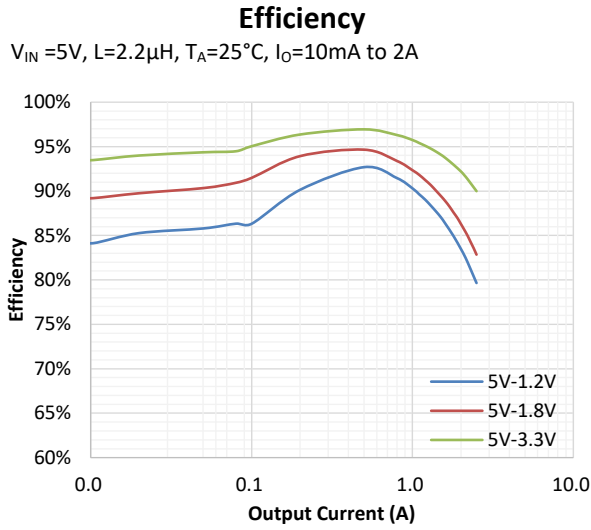


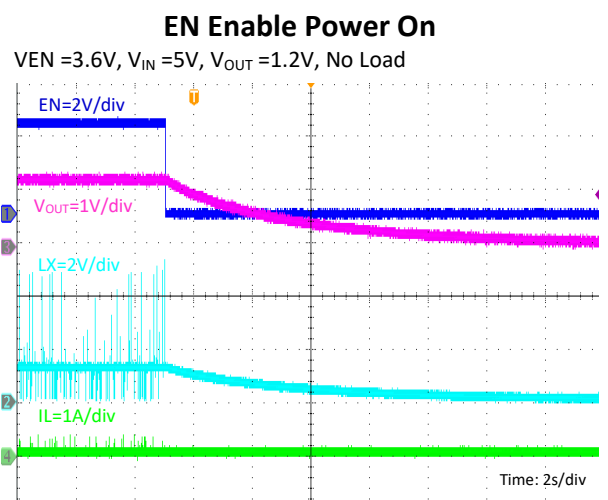
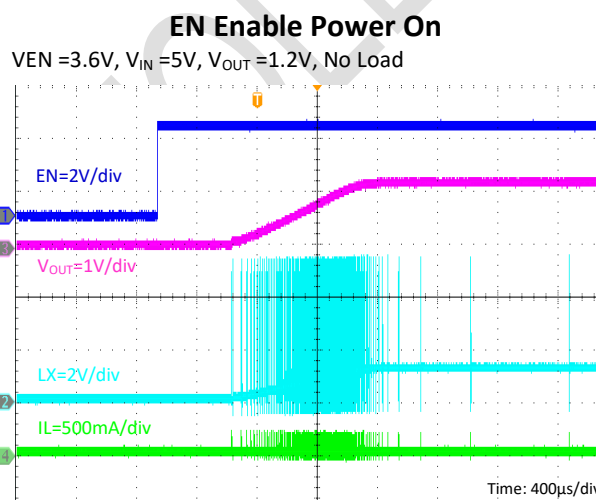
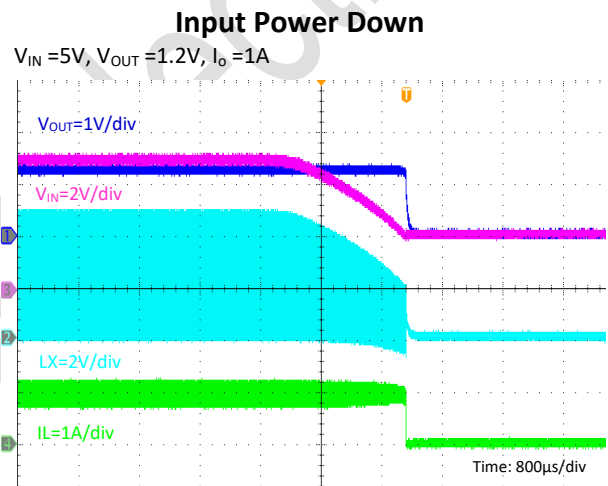
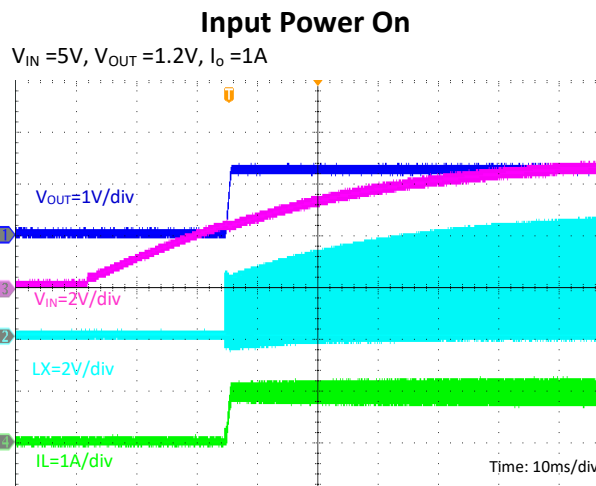
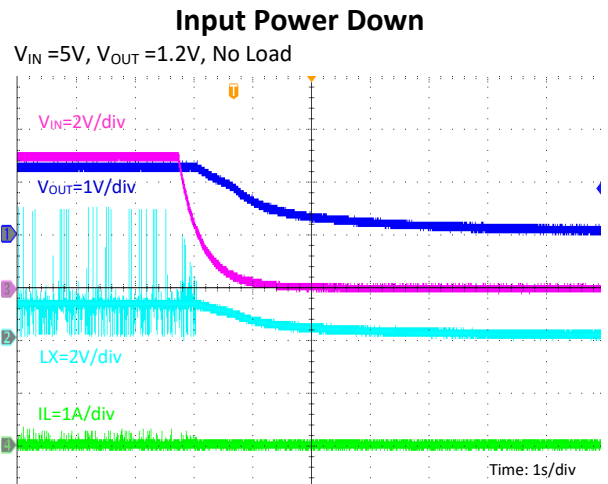
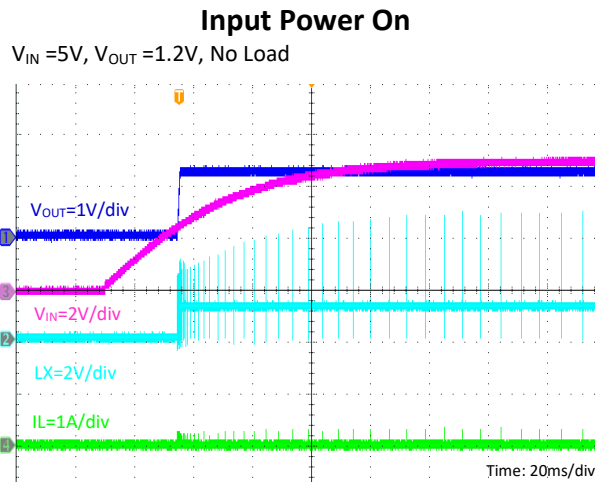
Figure 2. TMI3410 Block Diagram

**TYPICAL PERFORMANCE CHARACTERISTICS °C**

Test condition:  $V_{IN}=5V$ ,  $V_{OUT}=1.2V$ ,  $L=2.2\mu H$ ,  $T_A=+25^\circ C$ , unless other noted.



## TYPICAL PERFORMANCE CHARACTERISTICS (continued)

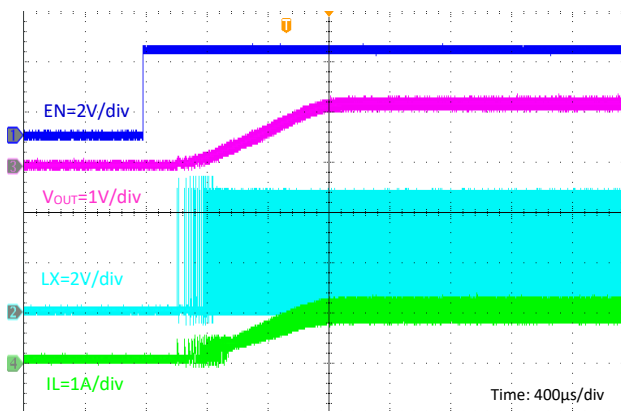




**TYPICAL PERFORMANCE CHARACTERISTICS** (continued)

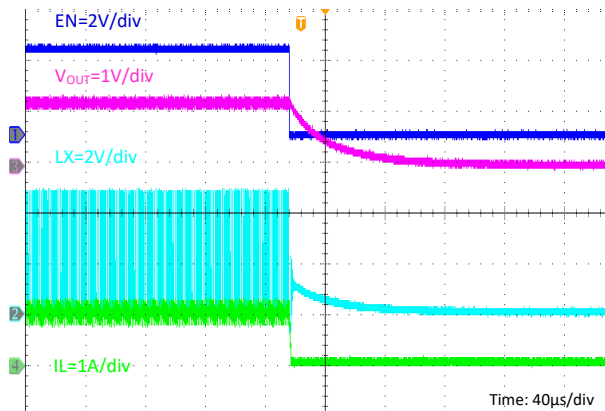
**EN Enable Power On**

$V_{EN}=3.6V, V_{IN}=5V, V_{OUT}=1.2V, I_o=2A$



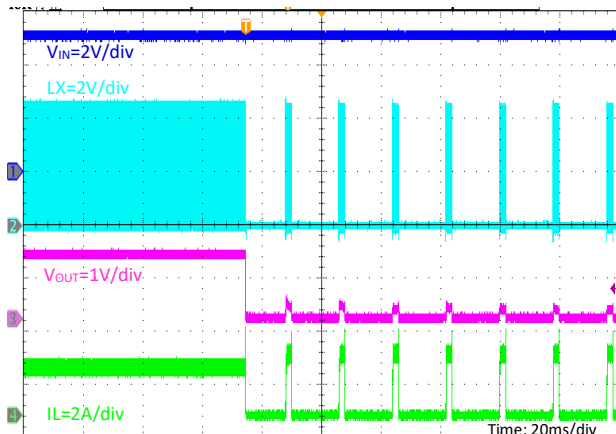
**EN Disable Power down**

$V_{EN}=3.6V, V_{IN}=5V, V_{OUT}=1.2V, I_o=2A$



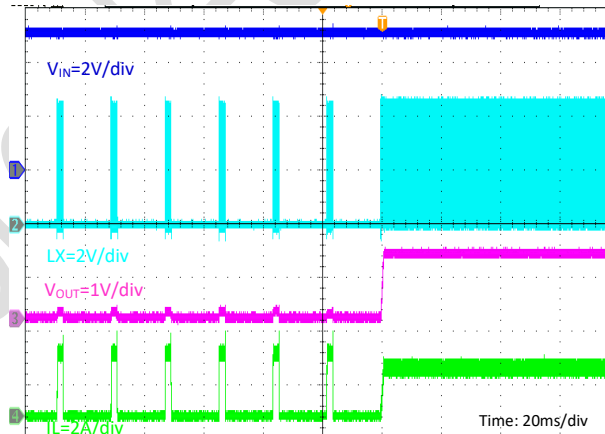
**Output Short Entry**

$V_{IN}=5V, V_{OUT}=1.2V, I_o=2A$



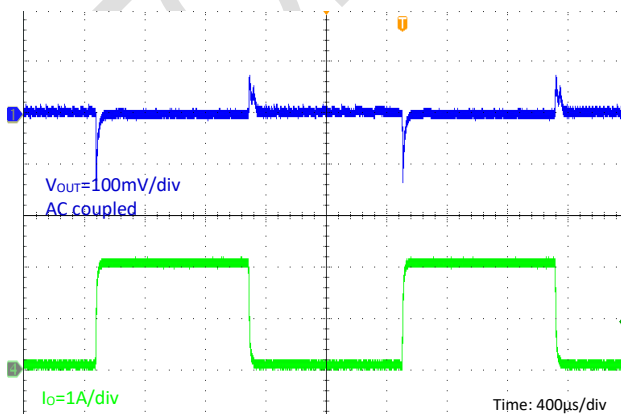
**Output Short Recovery**

$V_{IN}=5V, V_{OUT}=1.2V, I_o=2A$



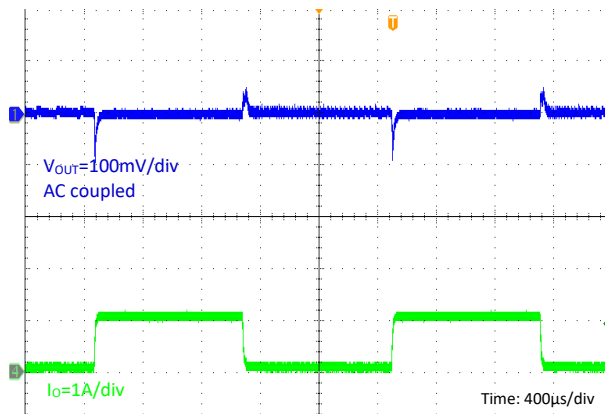
**Load Transient**

$V_{IN}=5V, V_{OUT}=1.2V, I_o=0A \text{ to } 2A$

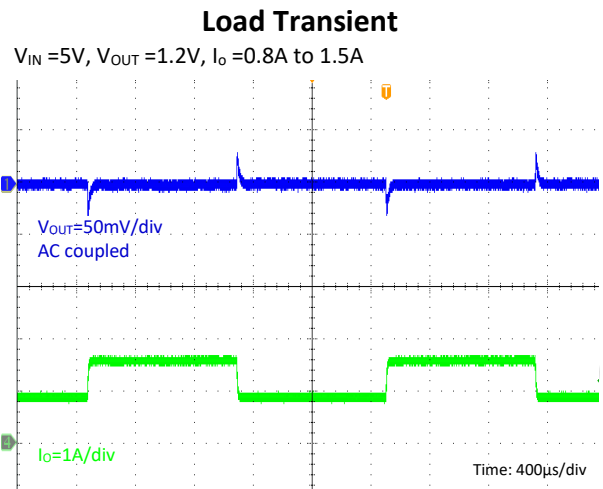
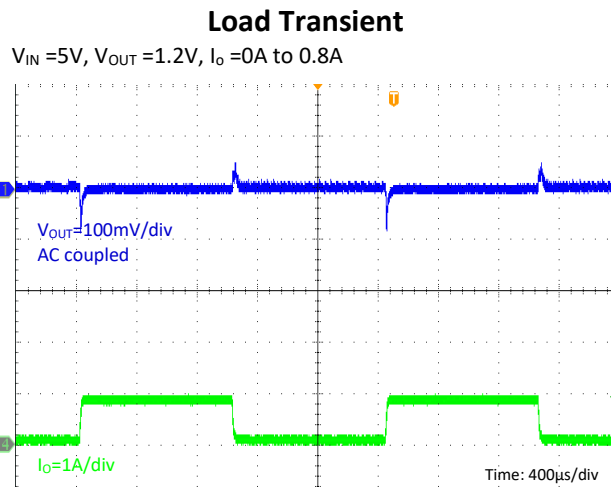


**Load Transient**

$V_{IN}=5V, V_{OUT}=1.2V, I_o=0A \text{ to } 1A$



## TYPICAL PERFORMANCE CHARACTERISTICS (continued)



## APPLICATION INFORMATION

### Setting the Output Voltage

Figure 1 shows the basic application circuit for the TMI3410. The TMI3410 can be externally programmed. Resistors R1 and R2 in Figure 1 program the output to regulate at a voltage higher than 0.6V. The external resistor sets the output voltage according to the following equation:

$$V_{OUT} = 0.6 \times \left(1 + \frac{R_1}{R_2}\right)$$

$$R_1 = (V_{OUT} / 0.6 - 1) \times R_2$$

### Inductor Selection

For most designs, 2.2μH inductance can satisfy most application conditions. Inductance value is related to inductor ripple current value, input voltage, output voltage setting and switching frequency. The inductor value can be derived from the following equation:

$$L = \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times \Delta I_L \times f_{OSC}}$$

Where  $\Delta I_L$  is inductor ripple current. Large value inductors result in lower ripple current and small value inductors result in high ripple current, so inductor value has effect on output voltage ripple value. DC resistance of inductor which has impact on efficiency of DC/DC converter should be taken into account when selecting the inductor.

### Input Capacitor Selection

The input capacitor reduces the surge current drawn from the input and switching noise from the device. The input capacitor impedance at the switching frequency should be less than input source impedance to prevent high frequency switching current passing to the input.

A low ESR input capacitor sized for maximum RMS current must be used. Ceramic capacitors with X5R or X7R dielectrics are highly recommended because of their low ESR and small temperature coefficients.

A 10μF ceramic capacitor for most applications is sufficient. A large value may be used for improved input voltage filtering. Input capacitor must be closed to IN and GND pin of the device.

### Output Capacitor Selection

The output capacitor is required to keep the output voltage ripple small and to ensure regulation loop stability. The output capacitor must have low impedance at the switching frequency. Ceramic capacitors with X5R or X7R dielectrics are recommended due to their low ESR and high ripple current ratings. The output ripple  $\Delta V_{OUT}$  is determined by:

$$\Delta V_{OUT} \leq \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times f_{OSC} \times L} \times \left( ESR + \frac{1}{8 \times f_{osc} \times C3} \right)$$

A 22μF ceramic can satisfy most applications. DC voltage derating of ceramic capacitor must be considered in applications, especially for 5V and 3.3V output voltage.

## Layout Consideration

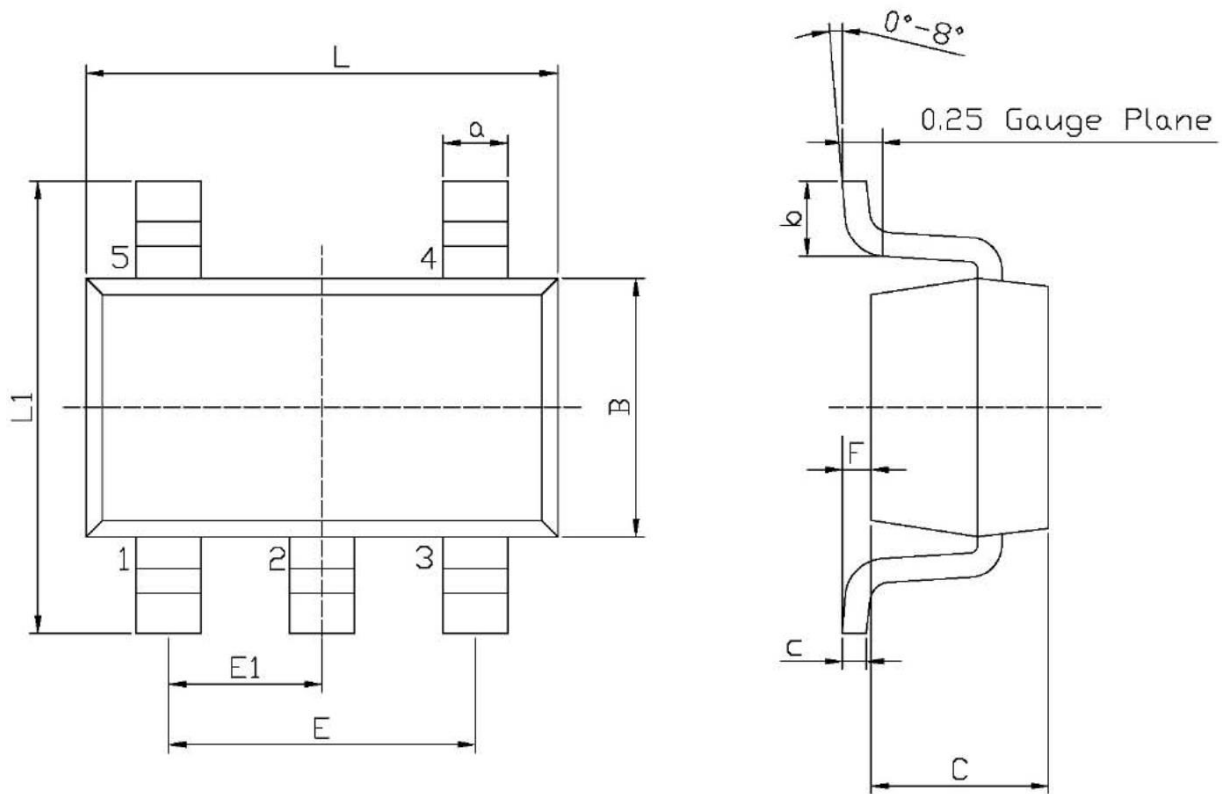
When laying out the printed circuit board, the following checking should be used to ensure proper operation of the TMI3410. Check the following in your layout:

1. The power traces, consisting of the GND trace, the LX trace and the IN trace should be kept short, direct and wide.
2. Does the (+) plates of Cin connect to Vin as closely as possible. This capacitor provides the AC current to the internal power MOSFETs.
3. Keep the switching node, LX, away from the sensitive VOUT node.
4. Keep the (-) plates of Cin and Cout as close as possible

TOLL Microelectronic

**PACKAGE INFORMATION**

**SOT23-5**



Unit: mm

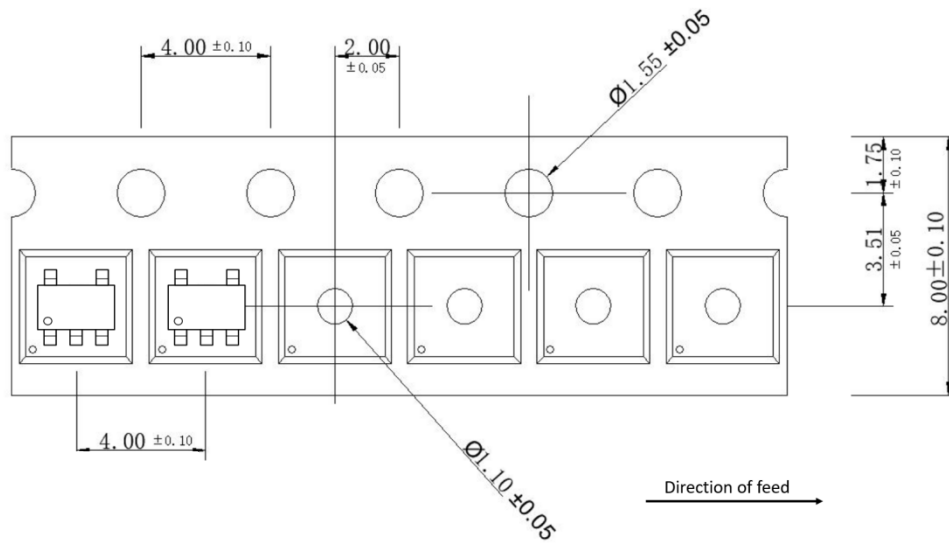
Symbol	Dimensions In Millimeters		Symbol	Dimensions In Millimeters	
	Min	Max		Min	Max
L	2.82	3.02	E1	0.85	1.05
B	1.50	1.70	a	0.35	0.50
C	0.90	1.30	c	0.10	0.20
L1	2.60	3.00	b	0.35	0.55
E	1.80	2.00	F	0	0.15

**Note:**

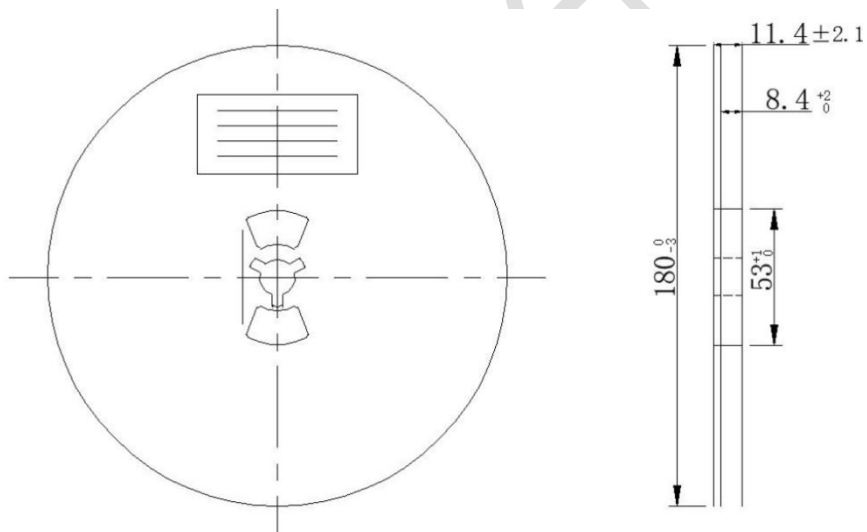
- 1) All dimensions are in millimeters.
- 2) Package length does not include mold flash, protrusion or gate burr.
- 3) Package width does not include inter lead flash or protrusion.
- 4) Lead popularity (bottom of leads after forming) shall be 0.10 millimeters max.
- 5) Pin 1 is lower left pin when reading top mark from left to right.

## TAPE AND REEL INFORMATION

### TAPE DIMENSIONS: SOT23-5



### REEL DIMENSIONS: SOT23-5



#### Note:

- 1) All Dimensions are in Millimeter
- 2) Quantity of Units per Reel is 3000
- 3) MSL level is level 3.

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[MB39A136PFT-G-BND-ERE1](#) [NCP1256BSN100T1G](#) [LV5768V-A-TLM-E](#) [NCP1365BABCYDR2G](#) [NCP1365AABCYDR2G](#)  
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