ETR05057-003

Ultra Low Quiescent Current Synchronous Step-Down PFM DC/DC Converter for Low Output Voltage

#### GENERAL DESCRIPTION

☆GreenOperation compatible

XC9272 series are Ultra Low Quiescent Current synchronous-rectification for Low Output Voltage type PFM step down DC/DC converters with a built-in 0.4Ω (TYP.) Pch driver and 0.4Ω (TYP.) Nch synchronous switching transistor, designed to allow the use of ceramic capacitor.

PFM control enables a low quiescent current, making these products ideal for battery operated devices that require high efficiency and long battery life.

Only inductor, C<sub>IN</sub> and C<sub>L</sub> capacitors are needed as external parts to make a step down DC/DC circuit.

Operation voltage range is from 2.0V to 6.0V. This product has fixed output voltage from 0.6V to 0.95V(accuracy: ±20mV) in increments of 0.05V.

During stand-by, all circuits are shutdown to reduce consumption to as low as 0.1µA(TYP.) or less.

With the built-in UVLO (Under Voltage Lock Out) function, the internal P-channel MOS driver transistor is forced OFF when input voltage gets lower than UVLO detection voltage. Besides, XC9272 series has UVLO release voltage of 1.8V (Typ.). The product with CL discharge function, XC9272B type, can discharge CL capacitor during stand-by mode due to the internal resistance by turning on the internal switch between VOUT-GND. This enables output voltage restored to GND level fast.

## APPLICATIONS

- Electric devices with GPS
- Wearable devices
- **Energy Harvest devices**
- Backup power supply circuits
- Devices with 1 Lithium cell

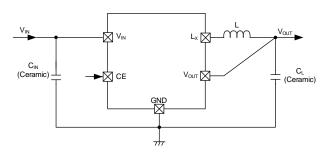
FEATURE	S
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Input Voltage Range	:	2.0V~6.0V
Output Voltage Setting	:	0.6V~0.95V (±20mV, 0.05V step increments)
Output Current	:	50mA
Driver Transistor	:	0.4Ω (Pch Driver Tr)
		$0.4\Omega$ (Nch Synchronous rectifier Switch Tr)
Supply Current	:	0.50µA @ Vout(t)=0.7V (TYP.)
Control Method	:	PFM control
High Speed Transient	:	50mV (VIN=3.6V, VOUT=0.7V, IOUT=10 $\mu$ A $\rightarrow$ 50mA)
PFM Switching Current	:	180mA
Function	:	Short Protection function
		C∟Discharge(XC9272B type)
		UVLO function
		Ceramic Capacitor Compatible
Operation Ambient Temperature	:	-40~+85°C
Package	:	SOT-25, USP-6EL
Environmentally Friendly	:	EU RoHS compliant, Pb Free

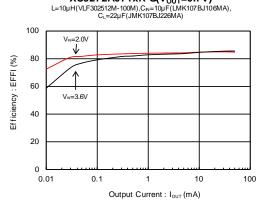
## ■ TYPICAL APPLICATION CIRCUIT

## ■TYPICAL PERFORMANCE **CHARACTERISTICS**

Efficiency vs. Output Current

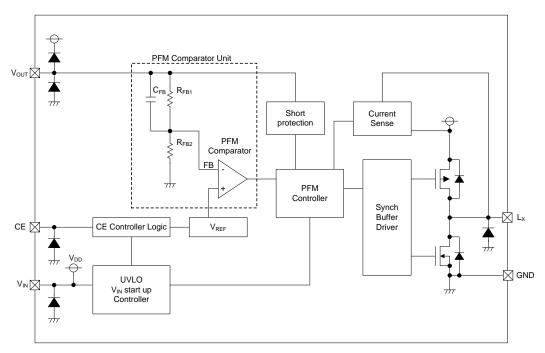


# XC9272A071xR-G(V<sub>OUT</sub>=0.7V)

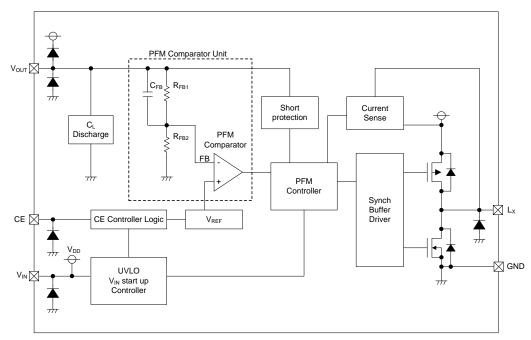


## BLOCK DIAGRAM

XC9272A Type



\* Diodes inside the circuit are an ESD protection diode and a parasitic diode.



XC9272B Type

\* Diodes inside the circuit are an ESD protection diode and a parasitic diode.

# ■ PRODUCT CLASSIFICATION

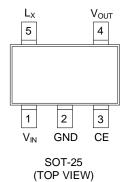
#### Ordering information

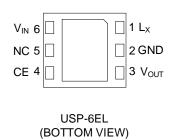
XC9272123456-7

DESIGNATOR	ITEM	SYMBOL	DESCRIPTION
			Without C∟ Discharge
1	Product Type	В	With CL Discharge
23	Output Voltage	06 ~ 09	Output Voltage : e.g. V <sub>OUT</sub> =0.7V⇒②=0, ③=7 Output Voltage Range: 0.6V~0.95V (0.05V step)
4		1	Output Voltage {x.x0V} (the 2nd decimal place is "0")
4	(4) Output Voltage Type		Output Voltage {x.x5V} (the 2nd decimal place is "5")
	)-⑦ <sup>(*1)</sup> Packages (Order Unit) -	4R-G	USP-6EL (3,000pcs/Reel)
(5)(6)-(7) ( 1)		MR-G	SOT-25 (3,000pcs/Reel)

<sup>(1)</sup> The "-G" suffix denotes Halogen and Antimony free as well as being fully EU RoHS compliant.

## ■PIN CONFIGURATION





\* The dissipation pad for the USP-6EL package should be solder-plated in reference mount pattern and metal masking so as to enhance mounting strength and heat release. The mount pattern should be connected to GND pin (No.2).

#### ■ PIN ASSIGNMENT

PIN NU	JMBER	PIN NAME	FUNCTIONS
USP-6EL	SOT-25		FUNCTIONS
1	5	Lx	Switching
2	2	GND	Ground
3	4	Vout	Output Voltage
4	3	CE	Chip Enable
5	-	NC	No Connection
6	1	V <sub>IN</sub>	Power Input

#### ■ CE PIN FUNCTION

PIN NAME	SIGNAL	STATUS
CE	Н	Operation (All Series)
	L	Standby (All Series)

\* Please do not leave the CE pin open.

## ■ABSOLUTE MAXIMUM RATINGS

PARAMETER		PARAMETER		SYMBOL	RATINGS	UNITS
V <sub>IN</sub> Pin V	oltage	Vin	-0.3 ~ +7.0	V		
L <sub>X</sub> Pin Ve	oltage	VLX	-0.3 ~ V <sub>IN</sub> +0.3 or +7.0 <sup>(*1)</sup>	V		
Vout Pin V	/oltage	Vout	-0.3 ~ V <sub>IN</sub> +0.3 or +7.0 <sup>(*1)</sup>	V		
CE Pin V	CE Pin Voltage		-0.3 ~ +7.0	V		
Lx Pin Current		ILX	1000	mA		
	SOT-25		250			
<b>Dowor Dissinction</b>	501-25	Pd	600 (40mm x 40mm Standard board) (*2)	mW		
Power Dissipation	USP-6EL	Pu	120	TIVV		
	(DAF)		750 (40mm x 40mm Standard board) $^{(*2)}$			
Operating Ambient Temperature		Topr	-40 ~ +85	°C		
Storage Temperature		Tstg	-55 ~ +125	°C		

\* All voltages are described based on the GND.

 $^{(*1)}$  The maximum value is the lower of either VIN + 0.3 or +7.0.

<sup>('2)</sup> The power dissipation figure shown is PCB mounted and is for reference only. The mounting condition is please refer to PACKAGING INFORMATION. Ta=25°C

Ta=25°C

## ■ELECTRICAL CHARACTERISTICS

#### ●XC9272A Type, without C<sub>L</sub> discharge function

PARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNITS	CIRCUIT
Input Voltage	V <sub>IN</sub>	-	2.0	-	6.0	V	1
Output Voltage	V <sub>OUT(E)</sub> <sup>(*2)</sup>	Resistor connected with $L_X$ pin. Voltage which $L_X$ pin changes "L" to "H" level while $V_{OUT}$ is decreasing.		E1		V	2
UVLO Release Voltage	V <sub>UVLO(E)</sub>	$V_{CE}=V_{IN}, V_{OUT}=0V.$ Resistor connected with $L_x$ pin. Voltage which $L_x$ pin changes "L" to "H" level while $V_{IN}$ is increasing.	1.65	1.8	1.95	V	2
UVLO Hysteresis Voltage	V <sub>HYS(E)</sub>	$\begin{split} &V_{CE}{=}V_{IN},V_{OUT}{=}0V. \text{ Resistor connected with }L_X \text{ pin.} \\ &V_{UVLO(E)} \text{ - Voltage which }L_X \text{ pin changes "H" to "L"} \\ &\text{level while }V_{IN} \text{ is decreasing.} \end{split}$	0.1	0.15	0.23	V	2
Supply Current	lq	$V_{\text{IN}} = V_{\text{CE}} = 2.0 \text{V},  V_{\text{OUT}} = V_{\text{OUT}(\text{T})} + 0.5 \text{V}^{(^*1)},  L_{\text{X}} = \text{Open}.$	-	0.5	0.8	μΑ	3
Standby Current	I <sub>STB</sub>	V <sub>IN</sub> =5.0V, V <sub>CE</sub> =V <sub>OUT</sub> =0V, L <sub>X</sub> =Open.	-	0.1	1.0	μA	3
L <sub>x</sub> SW "H" Leak Current	I <sub>LEAKH</sub>	$V_{IN}$ =5.0V, $V_{CE}$ = $V_{OUT}$ =0V, $V_{LX}$ =0V.	-	0.1	1.0	μA	3
L <sub>x</sub> SW "L" Leak Current	I <sub>LEAKL</sub>	$V_{IN}$ =5.0V, $V_{CE}$ = $V_{OUT}$ =0V, $V_{LX}$ =5.0V.	-	0.1	1.0	μA	3
PFM Switching Current	I <sub>PFM</sub>	$V_{IN}=V_{CE}=V_{OUT(T)}+2.0V^{(*1)}, I_{OUT}=10mA.$	115	180	250	mA	1
Efficiency (*3)	EFFI	$V_{IN}=V_{CE}=3.6V,$ $V_{OUT(T)}=0.7V^{(*1)}, I_{OUT}=30mA.$	-	85	-	%	1
LX SW "Pch" ON Resistance (*4)	R <sub>LXP</sub>	$V_{IN}=V_{CE}=5.0V$ , $V_{OUT}=0V$ , $I_{LX}=50mA$ .	-	0.4	0.65	Ω	4
LX SW "Nch" ON Resistance	R <sub>LXN</sub>	$V_{IN}=V_{CE}=5.0V.$	-	0.4 (*5)	-	Ω	-
Output Voltage Temperature Characteristics	Δν <sub>ουτ</sub> , (V <sub>ουτ</sub> • Δ <b>Τορr</b> )	-40°C≦Topr≦85°C.	-	±100	-	ppm/°C	2
CE "High" Voltage	V <sub>CEH</sub>	$V_{OUT}=0V$ . Resistor connected with $L_x$ pin. Voltage which $L_x$ pin changes "L" to "H" level while $V_{CE}=0.2 \rightarrow 1.5V$ .	1.2	-	6.0	V	5
CE "Low" Voltage V <sub>CEL</sub> V <sub>OUT</sub> =0V. Resi Voltage which		$V_{OUT}$ =0V. Resistor connected with L <sub>x</sub> pin. Voltage which L <sub>x</sub> pin changes "H" to "L" level while $V_{CE}$ =1.5 $\rightarrow$ 0.2V.	GND	-	0.3	V	5
CE "High" Current	I <sub>CEH</sub>	$V_{IN}=V_{CE}=5.0V, V_{OUT}=0V, L_X=Open.$	-0.1	-	0.1	μA	5
CE "Low" Current		V <sub>IN</sub> =5.0V, V <sub>CE</sub> =V <sub>OUT</sub> =0V, L <sub>X</sub> =Open.	-0.1	-	0.1	μA	5
Short Protection Threshold Voltage	V <sub>SHORT</sub>	Resistor connected with $L_X$ pin. Voltage which $L_X$ pin changes "H" to "L" level while $V_{OUT} = V_{OUT(T)} + 0.1 V \rightarrow 0 V^{(*1)}$ .	0.14	0.3	0.48	V	2

Unless otherwise stated, V\_{IN}=V\_{CE}=5.0V

(\*1) Vout(T)=Nominal Output Voltage

(\*2) VOUT(E)=Effective Output Voltage

The actual output voltage value  $V_{OUT(E)}$  is the PFM comparator threshold voltage in the IC.

Therefore, the DC/DC circuit output voltage, including the peripheral components, is boosted by the ripple voltage average value. Please refer to the characteristic example.

(\*3) EFFI=[{ (Output Voltage)x(Output Current)] / [(Input Voltage)x(Input Current)}]x100

 $^{(^*\!4)}$  LX SW "Pch" ON resistance = (V\_{IN}-V\_{LX}\,pin measurement voltage) / 50mA

(\*5) ) Designed value

# ■ELECTRICAL CHARACTERISTICS (Continued)

#### ●XC9272B Type, with C<sub>L</sub> discharge function

PARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNITS	CIRCUI
Input Voltage	V <sub>IN</sub>	-	2.0	-	6.0	V	1
Output Voltage	V <sub>OUT(E)</sub> <sup>(*2)</sup>	Resistor connected with $L_x$ pin. Voltage which $L_x$ pin changes "L" to "H" level while $V_{OUT}$ is decreasing.		E1		V	2
UVLO Release Voltage	V <sub>UVLO(E)</sub>	$V_{CE}=V_{IN}$ , $V_{OUT}=0V$ . Resistor connected with $L_x$ pin. Voltage which $L_x$ pin changes "L" to "H" level while $V_{IN}$ is increasing.	1.65	1.8	1.95	V	2
UVLO Hysteresis Voltage	$V_{\text{HYS}(\text{E})}$	$\label{eq:Vce} \begin{split} V_{CE} = & V_{IN}, \ V_{OUT} = 0V. \ Resistor \ connected \ with \ L_x \ pin. \\ V_{UVLO(E)} \ - \ Voltage \ which \ L_x \ pin \ changes \ "H" \ to \ "L" \\ level \ while \ V_{IN} \ is \ decreasing. \end{split}$	0.1	0.15	0.23	V	2
Supply Current	lq	$V_{IN}=V_{CE}=2.0V, V_{OUT}=V_{OUT(T)}+0.5V^{(*1)}, L_{X}=Open.$	-	0.5	0.8	μΑ	3
Standby Current	I <sub>STB</sub>	V <sub>IN</sub> =5.0V, V <sub>CE</sub> =V <sub>OUT</sub> =0V, L <sub>X</sub> =Open.	-	0.1	1.0	μA	3
L <sub>X</sub> SW "H" Leak Current	I <sub>LEAKH</sub>	$V_{IN}$ =5.0V, $V_{CE}$ = $V_{OUT}$ =0V, $V_{LX}$ =0V.	-	0.1	1.0	μA	3
L <sub>x</sub> SW "L" Leak Current	I <sub>LEAKL</sub>	$V_{IN}=5.0V, V_{CE}=V_{OUT}=0V, V_{LX}=5.0V.$	-	0.1	1.0	μA	3
PFM Switching Current	I <sub>PFM</sub>	$V_{IN}=V_{CE}=V_{OUT(T)}+2.0V^{(*1)}, I_{OUT}=10mA.$	115	180	250	mA	1
Efficiency (*3)	EFFI	V <sub>IN</sub> =V <sub>CE</sub> =3.6V, V <sub>OUT(T)</sub> =0.7V <sup>(*1)</sup> , I <sub>OUT</sub> =30mA.	-	85	-	%	1
LX SW "Pch" ON Resistance (*4)	R <sub>LXP</sub>	$V_{IN}=V_{CE}=5.0V, V_{OUT}=0V, I_{LX}=50mA.$		0.4	0.65	Ω	4
LX SW "Nch" ON Resistance	R <sub>LXN</sub>	$V_{IN}=V_{CE}=5.0V.$	-	0.4 (*5)	-	Ω	-
Output Voltage Temperature Characteristics	Δ/ <sub>ουτ</sub> / (V <sub>ουτ</sub> • ΔΓορr)	-40°C≦Topr≦85°C.	-	±100	-	ppm/°C	2
CE "High" Voltage	V <sub>CEH</sub>	$V_{OUT}=0V$ . Resistor connected with $L_x$ pin. Voltage which $L_x$ pin changes "L" to "H" level while $V_{CE}=0.2 \rightarrow 1.5V$ .	1.2	-	6.0	V	5
CE "Low" Voltage	V <sub>CEL</sub>	$V_{OUT}$ =0V. Resistor connected with L <sub>x</sub> pin. Voltage which L <sub>x</sub> pin changes "H" to "L" level while $V_{CE}$ =1.5 $\rightarrow$ 0.2V.	GND	-	0.3	V	5
CE "High" Current	I <sub>CEH</sub>	$V_{IN}=V_{CE}=5.0V, V_{OUT}=0V, L_X=Open.$	-0.1	-	0.1	μA	5
CE "Low" Current		$V_{IN}$ =5.0V, $V_{CE}$ = $V_{OUT}$ =0V, $L_X$ =Open.	-0.1	-	0.1	μA	(5)
Short Protection Threshold Voltage		Resistor connected with $L_x$ pin. Voltage which $L_x$ pin changes "H" to "L" level while $V_{OUT} = V_{OUT(T)} + 0.1 V \rightarrow 0 V^{(*1)}$ .	0.14	0.3	0.48	V	2
C <sub>L</sub> Discharge	R <sub>DCHG</sub>	$V_{IN}=V_{OUT}=5.0V, V_{CE}=0V, L_{X}=Open.$	55	80	105	Ω	3

Unless otherwise stated, VIN=VCE=5.0V

(\*1) VOUT(T)=Nominal Output Voltage

(\*2) VOUT(E)=Effective Output Voltage

The actual output voltage value VOUT(E) is the PFM comparator threshold voltage in the IC.

Therefore, the DC/DC circuit output voltage, including the peripheral components, is boosted by the ripple voltage average value. Please refer to the characteristic example.

('3) EFFI=[{ (Output Voltage)×(Output Current)] / [(Input Voltage)×(Input Current)}]×100

 $^{(^*\!4)}$  LX SW "Pch" ON resistance = (V\_{IN} - V\_{LX} pin measurement voltage) / 50mA

(\*5) Designed value

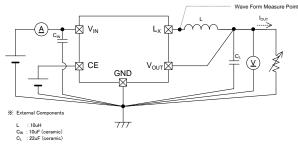
## ■ ELECTRICAL CHARACTERISTICS (Continued)

XC9272 series voltage specification chart

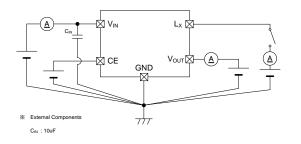
SYMBOL	E	1	
PARAMETER	Output Voltage		
UNITS: V	UNIT	rs: v	
OUTPUT VOLTAGE	MIN.	MAX.	
0.60	0.58	0.62	
0.65	0.63	0.67	
0.70	0.68	0.72	
0.75	0.73	0.77	
0.80	0.78	0.82	
0.85	0.83	0.87	
0.90	0.88	0.92	
0.95	0.93	0.97	

## TEST CIRCUITS

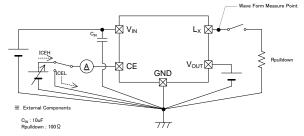
#### < Test Circuit No.1 >



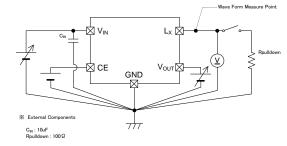
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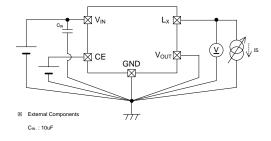
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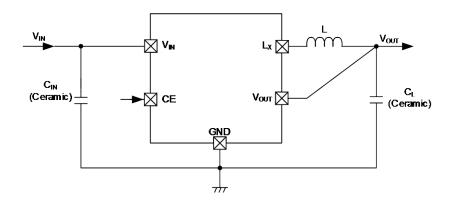
< Test Circuit No.2 >



< Test Circuit No.4 >



## ■TYPICAL APPLICATION CIRCUIT



#### [Typical Examples]

/		MANUFACTURE	PRODUCT NUMBER	VALUE
		TDK	VLF302512M-100M	10µH
	L	Coilcraft	LPS3015-103MRB	10µH
		Murata	DFE201610E-100M	10µH
	CIN	TAIYO YUDEN	LMK107BJ106MA	10µF/10V
	CL	TAIYO YUDEN	JMK107BJ226MA	22µF/6.3V

\* Take capacitance loss, withstand voltage, and other conditions into consideration when selecting components.

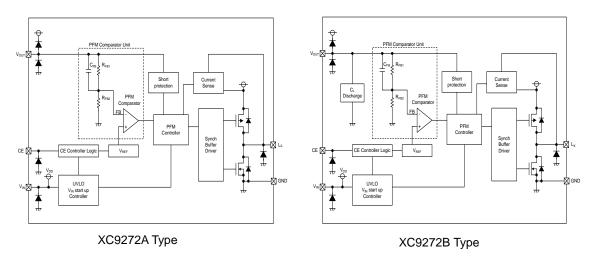
\* Characteristics are dependent on deviations in the coil inductance value. Test fully using the actual device.

\* A value of 10µH is recommended for the coil inductance.

\* If a tantalum or electrolytic capacitor is used for the load capacitance C<sub>L</sub>, ripple voltage will increase, and there is a possibility that operation will become unstable. Test fully using the actual device.

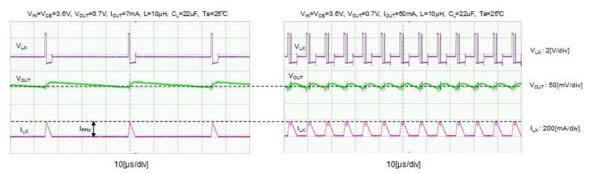
## ■OPERATIONAL EXPLANATION

The XC9272 series consists of a reference voltage supply, PFM comparator, Pch driver Tr, Nch synchronous rectification switch Tr, current sensing circuit, PFM control circuit, CE control circuit, and others. (Refer to the block diagram below.)



An ultra-low quiescent current circuit and synchronous rectification enable a significant reduction of dissipation in the IC, and the IC operates with high efficiency at both light loads and heavy loads. Current limit PFM is used for the control method, and even when switching current superposition occurs, increases of output voltage ripple are suppressed, allowing use over a wide voltage and current range. The IC is compatible with low-capacitance ceramic capacitors, and a small, high-performance step-down DC-DC converter can be created.

The actual output voltage  $V_{OUT(E)}$  in the electrical characteristics is the threshold voltage of the PFM comparator in the block diagram. Therefore the average output voltage of the step-down circuit, including peripheral components, depends on the ripple voltage. Before use, test fully using the actual device



<Reference voltage supply (VREF)> Reference voltage for stabilization of the output voltage of the IC.

#### <PFM control>

(1) The feedback voltage (FB voltage) is the voltage that results from dividing the output voltage with the IC internal dividing resistors  $R_{FB1}$  and  $R_{FB2}$ . The PFM comparator compares this FB voltage to  $V_{REF}$ . When the FB voltage is lower than  $V_{REF}$ , the PFM comparator sends a signal to the buffer driver through the PFM control circuit to turn on the Pch driver Tr. When the FB voltage is higher than  $V_{REF}$ , the PFM comparator sends a signal to prevent the Pch driver Tr from turning on.

(2) When the Pch driver Tr is on, the current sense circuit monitors the current that flows through the Pch driver Tr connected to the Lx pin. When the current reaches the set PFM switching current (IPFM), the current sense circuit sends a signal to the buffer driver through the PFM control circuit. This signal turns off the Pch driver Tr and turns on the Nch synchronous rectification switch Tr.

(3) The on time of the Nch synchronous rectification switch Tr is dynamically optimized inside the IC. After the off time elapses and the PFM comparator detects that the  $V_{OUT}$  voltage is higher than the set voltage, the PFM comparator sends a signal to the PFM control circuit that prevents the Pch driver Tr from turning on. However, if the  $V_{OUT}$  voltage is lower than the set voltage, the PFM comparator starts Pch driver Tr on.

By continuously adjusting the interval of the linked operation of (1), (2) and (3) above in response to the load current, the output voltage is stabilized with high efficiency from light loads to heavy loads.

## ■OPERATIONAL EXPLANATION (Continued)

#### <PFM Switching Current >

The PFM switching current monitors the current that flows through the Pch driver Tr, and is a value that limits the Pch driver Tr current. The Pch driver Tr remains on until the coil current reaches the PFM switching current ( $I_{PFM}$ ). An approximate value for this on-time ton can be calculated using the following equation: ton = L ×  $I_{PFM}$  / ( $V_{IN} - V_{OUT}$ )

#### <Maximum on-time function>

To avoid excessive ripple voltage in the event that the coil current does not reach the PFM switching current within a certain interval even though the Pch driver Tr has turned on and the FB voltage is above  $V_{REF}$ , the Pch driver Tr can be turned off at any timing using the maximum on-time function of the PFM control circuit. If the Pch driver Tr turns off by the maximum on-time function instead of the current sense circuit, the Nch synchronous rectification switch Tr will not turn on and the coil current will flow to the V<sub>OUT</sub> pin by means of the parasite diode of the Nch synchronous rectification switch Tr.

#### <VIN start mode>

When the  $V_{IN}$  voltage rises,  $V_{IN}$  start mode stops the short-circuit protection function during the interval until the FB voltage approaches  $V_{REF}$ . After the  $V_{IN}$  voltage rises and the FB voltage approaches  $V_{REF}$  by step-down operation,  $V_{IN}$  start mode is released. In order to prevent an excessive rush current while  $V_{IN}$  start mode is activated, Nch synchronous rectification switch Tr will not turn on and the coil current flows to the  $V_{OUT}$  pin by means of the parasitic diode of the Nch synchronous rectification Tr. In  $V_{IN}$  start mode as well, the coil current is limited by the PFM switching current.

#### <Short-circuit protection function>

The short-circuit protection function monitors the  $V_{OUT}$  voltage. In the event that the  $V_{OUT}$  pin is accidentally shorted to GND or an excessive load current causes the  $V_{OUT}$  voltage to drop below the set short-circuit protection voltage, the short-circuit protection function activates, and turns off and latches the Pch driver Tr at any selected timing. Once in the latched state, the IC is turned off and then restarted from the CE pin, or operation is started by re-applying the  $V_{IN}$  voltage.

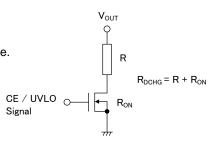
#### <UVLO function>

When the  $V_{IN}$  pin voltage drops below the UVLO detection voltage, the IC stops switching operation at any selected timing, turns off the Pch driver Tr and Nch synchronous rectification switch Tr (UVLO mode). When the  $V_{IN}$  pin voltage recovers and rises above the UVLO release voltage, the IC restarts operation.

#### <C<sub>L</sub> discharge function>

On the XC9272 series, a C<sub>L</sub> discharge function is available as an option (XC9272B type). This function enables quick discharging of the C<sub>L</sub> load capacitance when "L" voltage is input into the CE pin by the Nch Tr connected between the V<sub>OUT</sub>-GND pins, or in UVLO mode. This prevents malfunctioning of the application in the event that a charge remains on C<sub>L</sub> when the IC is stopped. The discharge time is determined by C<sub>L</sub> and the C<sub>L</sub> discharge resistance R<sub>DCHG</sub>, including the Nch Tr (refer to the diagram below). Using this time constant  $\tau = C_L \times R_{DCHG}$ , the discharge time of the output voltage is calculated by means of the equation below.

$$\begin{split} \mathsf{V} &= \mathsf{V}_{\mathsf{OUT}} \times e^{-t/\tau}, \text{ or in terms of } t, t = \tau \ln(\mathsf{V}_{\mathsf{OUT}} / \mathsf{V}) \\ \mathsf{V} : \mathsf{Output voltage after discharge} \\ \mathsf{V}_{\mathsf{OUT}} : \mathsf{Set output voltage} \\ t: \mathsf{Discharge time} \\ \mathsf{C}_{\mathsf{L}} : \mathsf{Value of load capacitance} (\mathsf{C}_{\mathsf{L}}) \\ \mathsf{R}_{\mathsf{DCHG}} : \mathsf{Value of } \mathsf{C}_{\mathsf{L}} \text{ discharge resistance Varies by power supply voltage.} \\ \tau : \mathsf{C}_{\mathsf{L}} \times \mathsf{R}_{\mathsf{DCHG}} \end{split}$$

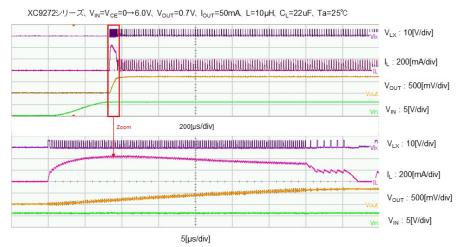


The CL discharge function is not available on the XC9272A type.

#### ■NOTE ON USE

1. Be careful not to exceed the absolute maximum ratings for externally connected components and this IC.

- 2. The DC/DC converter characteristics greatly depend not only on the characteristics of this IC but also on those of externally connected components, so refer to the specifications of each component and be careful when selecting the components. Be especially careful of the characteristics of the capacitor used for the load capacity C<sub>L</sub> and use a capacitor with B characteristics (JIS Standard) or an X7R/X5R (EIA Standard) ceramic capacitor.
- 3. Use a ground wire of sufficient strength. Ground potential fluctuation caused by the ground current during switching could cause the IC operation to become unstable, so reinforce the area around the GND pin of the IC in particular.
- 4. Mount the externally connected components in the vicinity of the IC. Also use short, thick wires to reduce the wire impedance.
- 5. When the voltage difference between V<sub>IN</sub> and V<sub>OUT</sub> is small, switching energy increases and there is a possibility that the ripple voltage will be too large. Before use, test fully using the actual device.
- 6. The CE pin does not have an internal pull-up or pull-down, etc. Apply the prescribed voltage to the CE pin.
- 7. If other than the inductance and capacitance values listed in the "Typical example" are used, excessive ripple voltage or a drop in efficiency may result.
- 8. If other than the inductance and capacitance values listed in the "Typical example" are used, a drop of output voltage at load transient may cause the short-circuit protection function to activate. Before use, test fully using the actual device.
- 9. At high temperature, excessive ripple voltage may occur and cause a drop in output voltage and efficiency. Before using at high temperature, test fully using the actual device
- 10. At light loads or when IC operation is stopped, leakage current from the Pch driver Tr may cause the output voltage to rise.
- 11. The average output voltage may vary due to the effects of output voltage ripple caused by the load current. Before use, test fully using the actual device.
- 12. If V<sub>IN</sub> voltage is high or the C<sub>L</sub> capacitance or load current is large, the output voltage rise time will lengthen when the IC is started, and coil current overlay may occur during the interval until the output voltage reaches the set voltage (refer to the diagram below).



- 13. When the IC is started, the short-circuit protection function does not operate during the interval until the V<sub>OUT</sub> voltage reaches a value near the set voltage.
- 14. If the load current is excessively large when the IC is started, it is possible that the V<sub>OUT</sub> voltage will not rise to the set voltage. Before use, test fully using the actual device.

## ■NOTE ON USE (Continued)

- 15. In actual operation, the maximum on-time depends on the peripheral components, input voltage, and load current. Before use, test fully using the actual device.
- 16. When the V<sub>IN</sub> voltage is turned on and off continuously, excessive rush current may occur while the voltage is on. Before use, test fully using the actual device.
- 17. When the V<sub>IN</sub> voltage is high, the Pch driver may change from on to off before the coil current reaches the PFM switching current (I<sub>PFM</sub>), or before the maximum on-time elapses. Before use, test fully using the actual device.
- 18. For temporary, transitional voltage drop or voltage rising phenomenon, the IC is liable to malfunction should the ratings be

exceeded.

19. Torex places an importance on improving our products and their reliability.

We request that users incorporate fail-safe designs and post-aging protection treatment when using Torex products in their

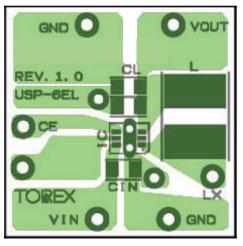
systems.

## ■NOTE ON USE (Continued)

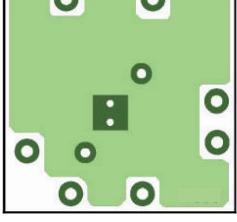
#### Instructions of pattern layouts

- 1. To suppress fluctuations in the V<sub>IN</sub> potential, connect a bypass capacitor (C<sub>IN</sub>) in the shortest path between the V<sub>IN</sub> pin and ground pin.
- 2. Please mount each external component as close to the IC as possible.
- 3. Wire external components as close to the IC as possible and use thick, short connecting traces to reduce the circuit impedance.
- 4. Make sure that the ground traces are as thick as possible, as variations in ground potential caused by high ground currents at the time of switching may result in instability of the IC.
- 5. Internal driver transistors bring on heat because of the transistor current and ON resistance of the driver transistors.

Reference Pattern Layout (USP-6EL)

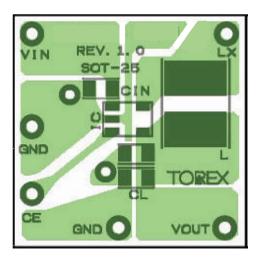


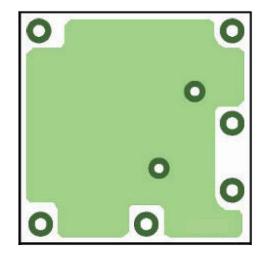
Top view



Bottom view

●Reference Pattern Layout (SOT-25)



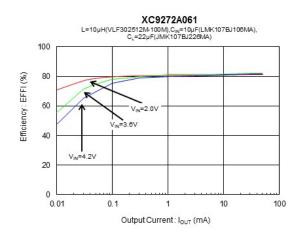


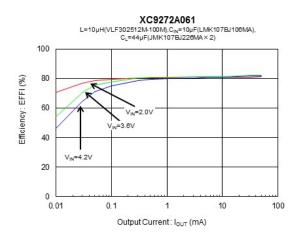
Top view

Bottom view

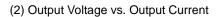
## TYPICAL PERFORMANCE CHARACTERISTICS

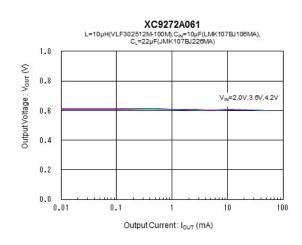
(1) Efficiency vs. Output Current



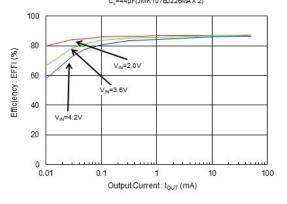


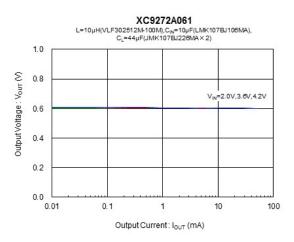
XC9272A09B  $\begin{array}{l} L=\!10\mu H(VLF302512M\!\cdot\!100M), C_{IN}\!\!=\!\!10\mu F(LMK107BJ108MA), \\ C_{L}\!\!=\!\!22\mu F(JMK107BJ228MA) \end{array}$ 100 80 Efficiency: EFFI (%) 60 3.6V 40 V<sub>IN</sub>=4.2V 20 0 0.01 0.1 1 10 100 Output Current : I<sub>OUT</sub> (mA)





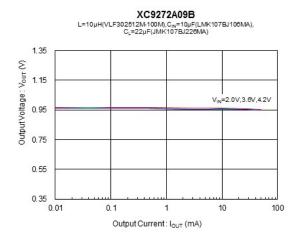
XC9272A09B L=10µH(VLF302612M-100M),C<sub>N</sub>=10µF(LMK107BJ108MA), C<sub>L</sub>=44µF(JMK107BJ226MA×2)

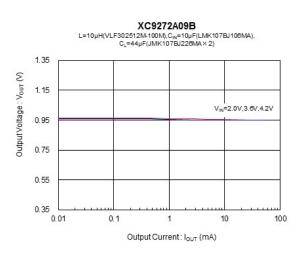




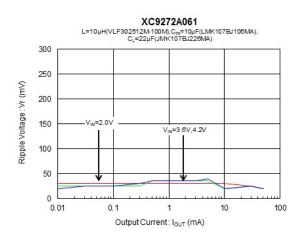
## ■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

#### (2) Output Voltage vs. Output Current





#### (3) Ripple Voltage vs. Output Current



 XC9272A061

 L=10µH(VLF302612M-100M), C<sub>R</sub>=10µF(LMK107BJ106MA), C<sub>R</sub>=40µF(JMK107BJ226MA×2)

 300

 250

 250

 250

 250

 150

 150

 50

 0

 0

 0

 0

 0

 0

 0

 0

 0

 0

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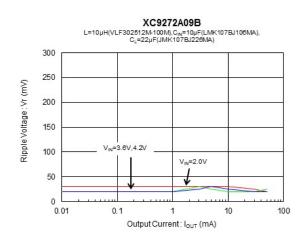
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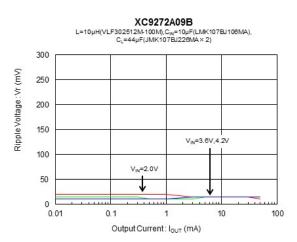
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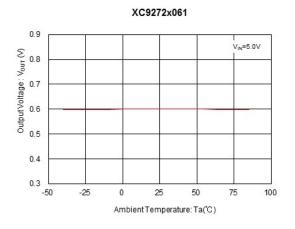


# TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

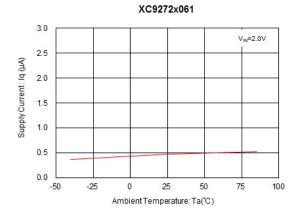
1.25

3.0

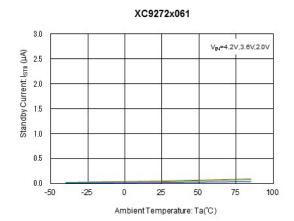
(4) Output Voltage vs. Ambient Temperature



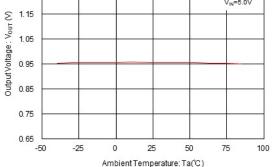
#### (5) Supply Current vs. Ambient Temperature

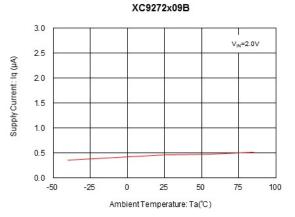


#### (6) Standby Current vs. Ambient Temperature

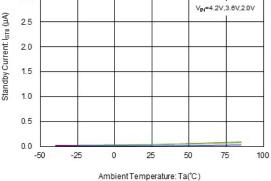


XC9272x09B V<sub>IN</sub>=5.0∨



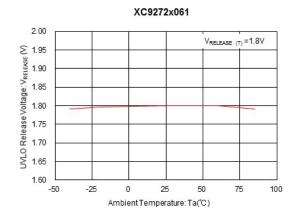




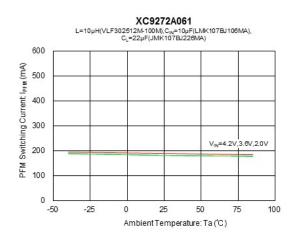


## TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

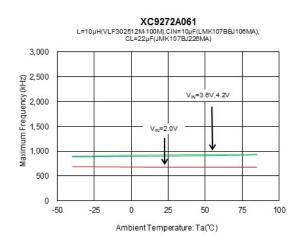
(7) UVLO Release Voltage vs. Ambient Temperature

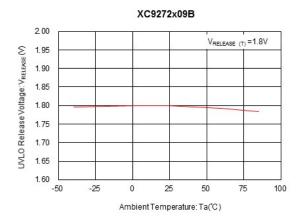


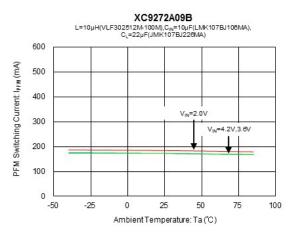
#### (8) PFM Switching Current vs. Ambient Temperature

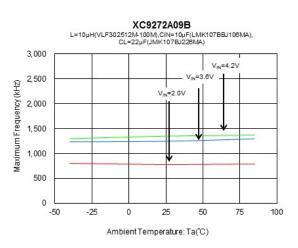


#### (9) Maximum Frequency vs. Ambient Temperature





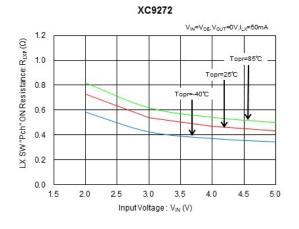




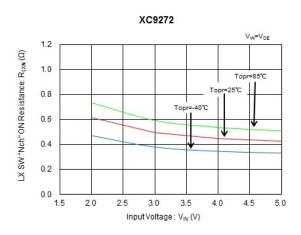
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## TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

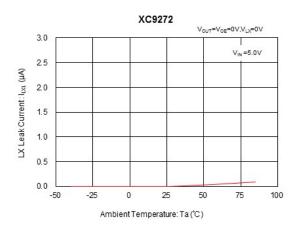
(10) Pch Driver ON Resistance vs. Ambient Temperature



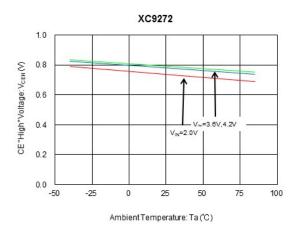




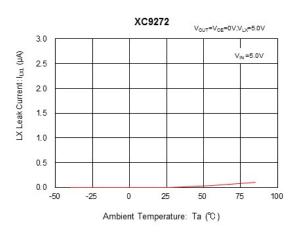
#### (12) Lx SW "H" Leakage Current vs. Ambient Temperature



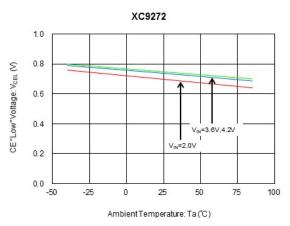
(14) CE "High" Voltage vs. Ambient Temperature



(13) Lx SW "L" Leakage Current vs. Ambient Temperature

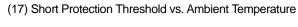


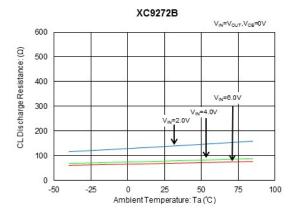
(15) CE "Low" Voltage vs. Ambient Temperature



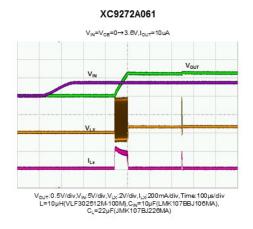
## TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

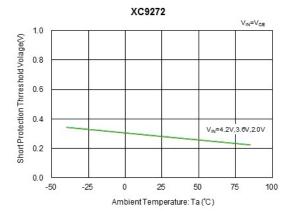
(16) CL Discharge vs. Ambient Temperature





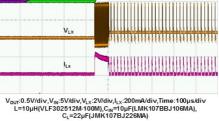
#### (18) Rising Output Voltage

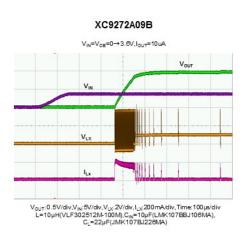


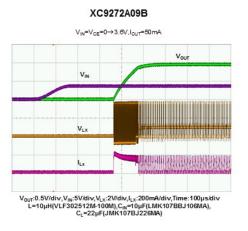


# V<sub>IN</sub>=V<sub>OE</sub>=0→3.6V.I<sub>OUT</sub>=50mA

XC9272A061

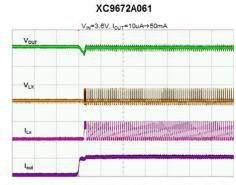




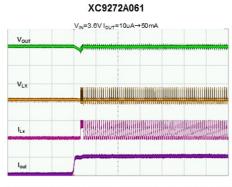


## ■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

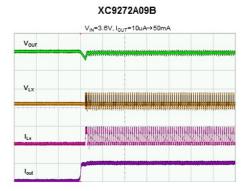
(19) Load Transient Response



 $\begin{array}{l} V_{0ur;} 200mV/div, I_{0ur;} 100mA/div, V_{Lx;} 5V/div, I_{Lx}; 200mA/div, Time: 100 \mu s/div, \\ L=10 \mu H(VLF302512M-100M), C_{N}=10 \mu F(LMK107BJ106MA), \\ C_L=22 \mu F(JMK107BJ226MA) \end{array}$ 

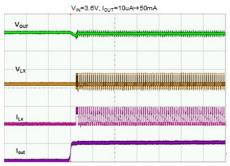


 $\begin{array}{l} V_{00T}: 200mV/div, I_{0UT}: 100mA/div, V_{LX}: 5V/div, I_{LX}: 200mA/div, Time:: 100 \mu s/div \\ L=10 \mu H(VLF302512M:100M), C_{N}=10 \mu F(LMK107BJ106MA), \\ C_{L}=44 \mu F(JMK107BJ226MA \times 2) \end{array}$ 



 $\begin{array}{l} V_{0UT}: 200mV/div, I_{0UT}: 50mA/div, V_{LX}: 5V/div, J_{LX}: 200mA/div, Time: 100 \mu s/div \\ L=10 \mu H(VLF302512M\cdot100M), C_{IN}= 10 \mu F(LMK107BJ106MA), \\ C_{L}=22 \mu F(JMK107BJ226MA) \end{array}$ 





 $\begin{array}{l} V_{00T}: 200mV/div, J_{00T}: 50mA/div, V_{LX}: 51/ldiv, J_{LX}: 200mA/div, Time: 100 \mu s/div \\ L=10 \mu H(VLF302512M-100M), C_{IN}=10 \mu F(LMK107BJ106MA), \\ C_L=44 \mu F(JMK107BJ226MA \times 2) \end{array}$ 

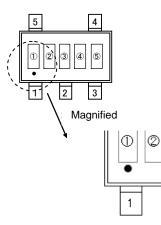
## ■ PACKAGING INFORMATION

For the latest package information go to, www.torexsemi.com/technical-support/packages

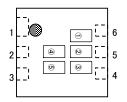
PACKAGE	OUTLINE / LAND PATTERN	THERMAL CHARACTERISTICS			
SOT-25	SOT-25 PKG	Standard Board	SOT-25 Power Dissipation		
USP-6EL(DAF)	USP-6EL PKG	Standard Board	USP-6EL Power Dissipation		

#### MARKING RULE

SOT-25(Under dot)



#### OUSP-6EL



#### 1 represents product series

MARK	PRODUCT SERIES	
С	XC9272A/B****-G	
WOOT OF HELE		

SOT-25 Under dot

#### ② represents output voltage

MARK	OUTPUT VOLTAGE		PRODUCT SERIES	
Ν	0.6	0.65	XC9272*06***-G	
Р	0.7	0.75	XC9272*07***-G	
R	0.8	0.85	XC9272*08***-G	
S	0.9	0.95	XC9272*09***-G	

③ represents product type and output voltage type

MARK	PRODUCT TYPE	OUTPUT VOLTAGE TYPE	PRODUCT SERIES
N	Without C <sub>L</sub> Discharge	Output Voltage {x.x0V} (the 2nd decimal place is "0")	XC9272A**1**-G
Р	Without C <sub>L</sub> Discharge	Output Voltage {x.x5V} (the 2nd decimal place is "5")	XC9272A**B**-G
R	With C <sub>L</sub> Discharge	Output Voltage {x.x0V} (the 2nd decimal place is "0")	XC9272B**1**-G
S	With $C_L$ Discharge	Output Voltage {x.x5V} (the 2nd decimal place is "5")	XC9272B**B**-G

(4)5 represents production lot number

01~09、0A~0Z、11~9Z、A1~A9、AA~AZ、B1~ZZ

(G, I, J, O, Q, W excluded)

\* No character inversion used.

- 1. The product and product specifications contained herein are subject to change without notice to improve performance characteristics. Consult us, or our representatives before use, to confirm that the information in this datasheet is up to date.
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- 5. Although we make continuous efforts to improve the quality and reliability of our products; nevertheless Semiconductors are likely to fail with a certain probability. So in order to prevent personal injury and/or property damage resulting from such failure, customers are required to incorporate adequate safety measures in their designs, such as system fail safes, redundancy and fire prevention features.
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