XDL603/XDL604 Series

ETR44003-002

18V Operation 500mA Inductor Built-in Step-down "micro DC/DC" Converter

☆AEC-Q100 Grade2

■GENERAL DESCRIPTION

The XDL603/XDL604 series is an ultra compact step-down DC / DC converter that integrates a coil and a control IC in one tiny package. By adding a ceramic capacitor for input / output and a resistor for output voltage setting to external parts, a power supply circuit of up to 500mA can be created. An internal coil simplifies the circuit and enables minimization of noise and other operational trouble due to the circuit wiring.

XDL603/XDL604 series has operating voltage range of 3.0V~18.0V and it can support 500mA as an output current with high-efficiency. They use synchronous rectification at an operating frequency of 2.2MHz. The output voltage can be set to a value from 1.0V to 5.0V using external resistors.

They have a fixed internal soft start time which is 1.0ms(TYP.), additionally the time can be extended by using an external resistor and capacitor.The output state can be monitored using the power good function. Over current protection. short-circuit protection and thermal shutdown are embedded and they secure a safety operation even with a short-circuit.

The XDL603/XD604 series employ the wettable flank plated packaging. This provides a visual indicator of solderability and lowers the inspection time.

■APPLICATIONS

● Automotive Body Control
● Automotive Infotainment
● Automotive accessories
・Drive recorder
・Car-mounted camera
・ETC
● Industrial Equipment

■FEATURES

|		
Input Voltage Range	:	3.0V ~ 18.0V (Absolute Max 20V)
Output Voltage Range		1.0V ~ 5.0V
FB Voltage	:	0.75V ± 1.5%
Oscillation Frequency	:	2.2MHz
Output Current	:	500mA
Quiescent Current		13.5µA (XDL604)
Control Methods	:	PWM control (XDL603)
		PWM/PFM Auto (XDL604)
		Efficiency 81%@12V→5V, 300mA
Function	:	Soft-start External settings
		Power good
Protection Function		Over Current Protection
	:	- Automatic recovery (B Type)
		- Integral Latch (A Type)
		Thermal Shutdown
		UVLO
Output Capacitor	:	Ceramic Capacitor
Operating Ambient Temperature	:	-40℃ ~ 105℃
Packages	:	DFN3625-11B (Wettable Flank)
Environmentally Friendly	:	EU RoHS Compliant, Pb Free

■TYPICAL APPLICATION
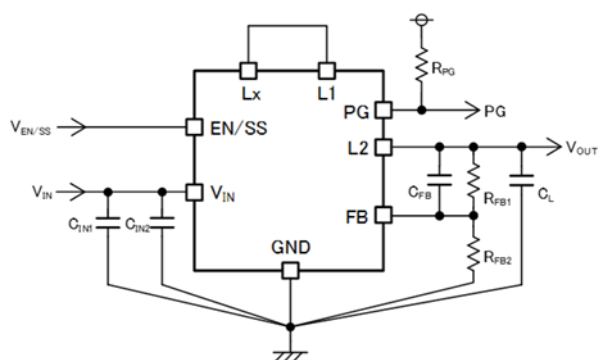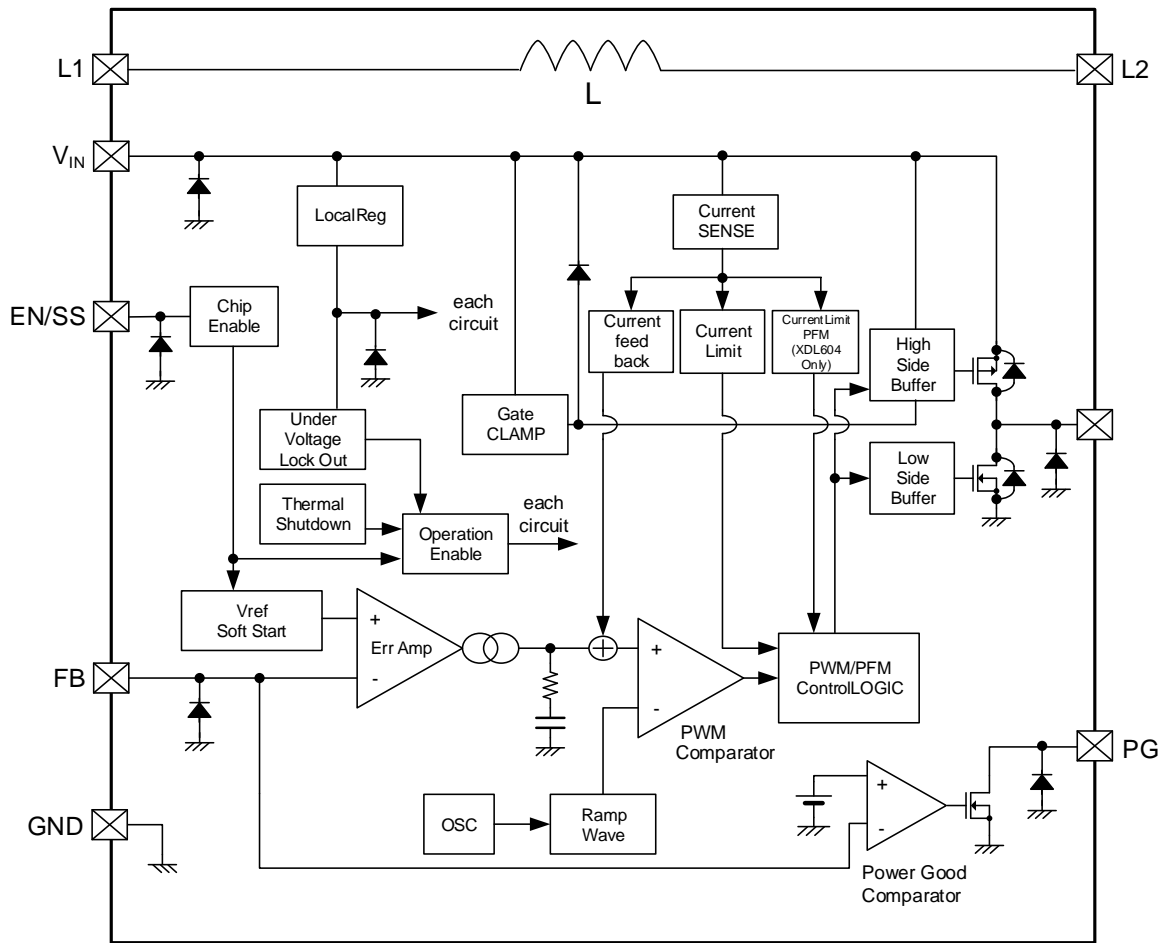

■ TYPICAL PERFORMANCE CHARACTERISTICS

XDL603x75D/XDL604x75D
(V_IN=12V, V_OUT=5V)

C_IN=2.2µF(CGA4J3X7R1E225K125AB),
C_L=10µF×2 (CGA5L1X7R1C106K160AC)
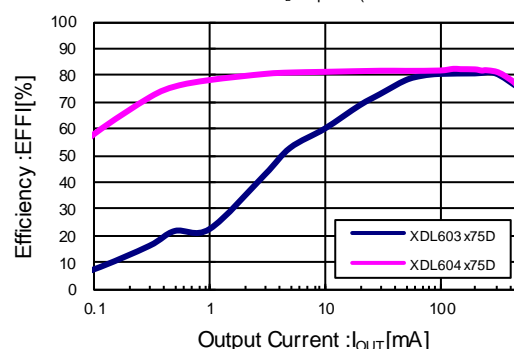
■BLOCK DIAGRAM

* Diodes inside the circuit are an ESD protection diodes and a parasitic diode.

■PRODUCT CLASSIFICATION

●Ordering Information

XDL603①②③④⑤⑥-⑦   PWM control
XDL604①②③④⑤⑥-⑦   PWM/PFM Auto.

DESIGNATOR	ITEM	SYMBOL	DESCRIPTION
①	Type	A	Refer to Selection Guide
		B	
②③	FB Voltage	75	Output voltage can be adjusted in 1.0V to 5.0V
④	Oscillation Frequency	D	2.2MHz
⑤⑥-⑦[*1]	Packages (Order Unit)	82-Q	DFN3625-11B (2,000pcs/Reel)[*2]

[*1] The "-Q" suffix denotes "AEC-Q100" compliant.

[*2] "Halogen and Antimony free" as well as being fully EU RoHS compliant.
   The products are shipped in a moisture-proof packing.

●Selection Guide

FUNCTION	A TYPE	B TYPE
Chip Enable	Yes	Yes
UVLO	Yes	Yes
Thermal Shutdown	Yes	Yes
Soft Start	Yes	Yes
Power-Good	Yes	Yes
Current Limiter (Automatic Recovery)	-	Yes
Current Limiter (Latch Protection [*3])	Yes	-

[*3] The over-current protection latch is an integral latch type.

XDL603/XDL604 Series

■PIN CONFIGURATION

DFN3625-11B
(BOTTOM VIEW)
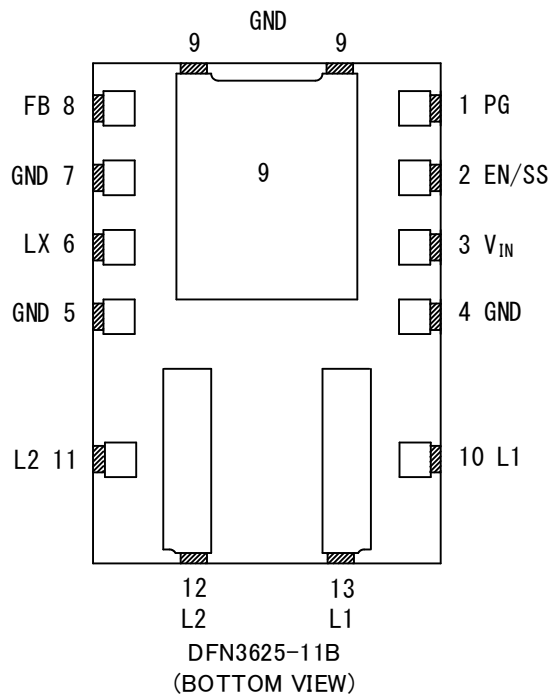
* The dissipation pad(No.9) pin for the DFN3625-11B package should be solder-plated in recommended mount pattern and metal masking so as to enhance mounting strength and heat release. If the pad needs to be connected to other pins, it should be connected to the GND (No. 4,5,7) pin.

■ PIN ASSIGNMENT

PIN NUMBER	PIN NAME	FUNCTIONS
1	PG	Power-good Output
2	EN/SS	Enable Soft-start
3	V_{IN}	Power Input
4	GND	Ground
5	GND	Ground
6	L_X	Switching Output
7	GND	Ground
8	FB	Output Voltage Sense
9	GND	Ground
10,13	L1	Inductor Electrodes
11,12	L2	Inductor Electrodes

■ FUNCTION CHART

PIN NAME	SIGNAL	STATUS
EN/SS	L	Stand-by
	H	Active
	OPEN	Undefined State[*1]

[*1] Please do not leave the EN/SS pin open. Each should have a certain voltage.

PIN NAME	CONDITION		SIGNAL
PG	EN/SS = H	$V_{FB} > V_{PGDET}$	H (High impedance)
		$V_{FB} \leqq V_{PGDET}$	L (Low impedance)
		Thermal Shutdown	L (Low impedance)
		UVLO ($V_{IN} < V_{UVLO1}$)	Undefined State
	EN/SS = L	Stand-by	L (Low impedance)

■ ABSOLUTE MAXIMUM RATINGS

PARAMETER		SYMBOL	RATINGS	UNITS
V_{IN} Pin Voltage		V_{IN}	-0.3 ~ 20	V
EN/SS Pin Voltage		$V_{EN/SS}$	-0.3 ~ 20	V
FB Pin Voltage		V_{FB}	-0.3 ~ 6.2	V
PG Pin Voltage		V_{PG}	-0.3 ~ 6.2	V
PG Pin Current		I_{PG}	8	mA
Lx Pin Voltage		V_{Lx}	-0.3 ~ V_{IN} + 0.3 or 20 [*1]	V
Lx Pin Current		I_{Lx}	1800	mA
Power Dissipation (Ta=25℃)	DFN3625-11B	Pd	2100 (JESD71-5 board) [*2]	mW
Operating Ambient Temperature		Topr	-40 ~ 105	℃
Storage Temperature		Tstg	-55 ~ 125	℃

All voltages are described based on the GND pin.

[*1] The maximum value should be either V_{IN}+0.3V or 20V in the lowest.

[*2] The power dissipation figure shown is PCB mounted and is for reference only.
   Please refer to PACKAGING INFORMATION for the mounting condition

■ELECTRICAL CHARACTERISTICS

XDL603/XDL604 series   Ta=25℃

PARAMETER	SYMBOL	CONDITIONS		MIN.	TYP.	MAX.	UNIT	CIRCUIT
FB Voltage	V_{FB}	V_{FB}=0.731V → 0.769V V_{FB} Voltage when Lx pin voltage changes from "H" level to "L" level		0.739	0.750	0.761	V	②
			-40℃≦Ta≦105℃	0.731	0.750	0.769		
Output Voltage Setting Range[*1]	V_{OUTSET}	-	-40℃≦Ta≦105℃	1.0	-	5.0	V	–
Operating Voltage Range	V_{IN}	-	-40℃≦Ta≦105℃	3.0	-	18.0	V	–
UVLO Detect Voltage	V_{UVLO1}	V_{IN}:2.87V→2.53V, V_{FB}=0.675V, V_{IN} Voltage when Lx pin voltage changes from "H" level to "L" level		2.60	2.70	2.80	V	②
			-40℃≦Ta≦105℃	2.53	-	2.87		
UVLO Release Voltage	V_{UVLO2}	V_{IN}:2.63V→2.97V, V_{FB}=0.675V V_{IN} Voltage when Lx pin voltage changes from "L" level to "H" level		2.70	2.80	2.90	V	②
			-40℃≦Ta≦105℃	2.63	-	2.97		
Quiescent Current	I_q	V_{FB}=0.825V	XDL603	-	145	238	µA	④
			-40℃≦Ta≦105℃	-	-	257		
			XDL604	-	13.5	18.5		
			-40℃≦Ta≦105℃	-	-	20.0		
Stand-by Current	I_{STB}	$V_{EN/SS}$=0V		-	1.65	2.50	µA	⑤
			-40℃≦Ta≦105℃	-	-	2.80		
Oscillation Frequency	f_{OSC}	Connected to external components I_{OUT}=100mA		2013	2200	2387	kHz	①
			-40℃≦Ta≦105℃	1813	-	2531		
Minimum Duty Cycle	D_{MIN}	V_{FB}=0.825V	-40℃≦Ta≦105℃	-	-	0	%	②
Maximum Duty Cycle	D_{MAX}	V_{FB}=0.675V	-40℃≦Ta≦105℃	100	-	-	%	②
Lx SW "H" On Resistance	R_{LxH}	V_{FB}=0.675V, I_{LX}=200mA		-	0.95	1.10	Ω	②
Lx SW "L" On Resistance	R_{LxL}	V_{FB}=0.825V, I_{LX}=200mA		-	0.69[*2]	-	Ω	②
PFM Switch Current	I_{PFM}	XDL604 only Connected to external components, I_{OUT}=1mA		-	370	-	mA	①
High side Current Limit[*3]	I_{LIMH}	V_{FB}=0.675V		920	1100	-	mA	②
Latch Time	t_{LAT}	Type A only Connected to external components, V_{FB}=0V		0.5	1.0	1.7	ms	⑥
Internal Soft-Start Time	t_{SS1}	$V_{EN/SS}$=0V→12V, V_{FB}=0.675V Time until Lx pin oscillates		0.5	1.0	1.7	ms	②
External Soft-Start Time	t_{SS2}	$V_{EN/SS}$=0V→12V、V_{FB}=0.675V R_{SS}=430kΩ, C_{SS}=0.47µF Time until Lx pin oscillates		17	26	35	ms	③

Test Condition: Unless otherwise stated: V_{IN}=12V, $V_{EN/SS}$=12V, V_{PG}:OPEN

Peripheral parts connection conditions：RFB1=220kΩ, RFB2=39kΩ, CFB=150pF, CL=10µF×2parallel, CIN=2.2µF

The ambient temperature range (-40℃≦Ta≦105℃) is a design Value.

[*1] Please use within the range of $V_{OUT}/V_{IN} \geqq 0.17$

[*2] Design reference value. This parameter is provided only for reference.

[*3] Current limit denotes the level of detection at peak of coil current.

■ELECTRICAL CHARACTERISTICS (Continued)

XDL603/XDL604 series   Ta=25℃

PARAMETER	SYMBOL	CONDITIONS		MIN.	TYP.	MAX.	UNIT	CIRCUIT
PG Detect Voltage	V_{PGDET}	V_{FB}=0.720V→0.630V, R_{PG}:100kΩ pull-up to 5V		0.638	0.675	0.712	V	②
		V_{FB} Voltage when PG pin voltage changes from "H" level to "L"level	-40℃≦Ta≦105℃	0.630	-	0.720		
PG Output Voltage	V_{PG}	V_{FB}=0.6V、I_{PG}=1mA	-40℃≦Ta≦105℃	-	-	0.3	V	②
Efficiency	EFFI[*2]	Connected to external components V_{OUTSET}=5V, I_{OUT}=300mA		-	81[*1]	-	%	①
FB "H" Current	I_{FBH}	V_{IN}=$V_{EN/SS}$=18V, V_{FB}=3.0V	-40℃≦Ta≦105℃	-0.1	0.0	0.1	µA	④
FB "L" Current	I_{FBL}	V_{IN}=$V_{EN/SS}$=18V, V_{FB}=0V	-40℃≦Ta≦105℃	-0.1	0.0	0.1	µA	④
EN/SS "H" Current	$I_{EN/SSH}$	V_{IN}=$V_{EN/SS}$=18V, V_{FB}=0.825V	-40℃≦Ta≦105℃	-	0.1	0.3	µA	④
EN/SS "L" Current	$I_{EN/SSL}$	V_{IN}=18V, $V_{EN/SS}$=0V, V_{FB}=0.825V	-40℃≦Ta≦105℃	-0.1	0.0	0.1	µA	④
EN/SS "H" Voltage	$V_{EN/SSH}$	$V_{EN/SS}$=0.3V→2.5V, V_{FB}=0.71V $V_{EN/SS}$ Voltage when Lx pin voltage changes from "L" level to "H"level	-40℃≦Ta≦105℃	2.5	-	18.0	V	②
EN/SS "L" Voltage	$V_{EN/SSL}$	$V_{EN/SS}$=2.5V→0.3V, V_{FB}=0.71V $V_{EN/SS}$ Voltage when Lx pin voltage changes from "H" level to "L"level	-40℃≦Ta≦105℃	GND	-	0.3	V	②
Thermal Shutdown Temperature	T_{TSD}	Junction Temperature		-	150	-	℃	－
Hysteresis Width	T_{HYS}	Junction Temperature		-	25	-	℃	－
Inductance	L	Test Freq.=1MHz		-	2.2	-	µH	－
Inductor Rated Current	I_{DC}	ΔT=+40deg		-	1.6	-	A	－

Test Condition: Unless otherwise stated: V_{IN}=12V, $V_{EN/SS}$=12V, V_{PG}:OPEN

Peripheral parts connection conditions： RFB1=220kΩ, RFB2=39kΩ, CFB=150pF, CL=10µF×2parallel, CIN=2.2µF

The ambient temperature range (-40℃≦Ta≦105℃) is a design Value.

[*1] Design reference value. This parameter is provided only for reference.

[*2] EFFI = {(output voltage) x (output current)} / {(input voltage) x (input current)} x 100
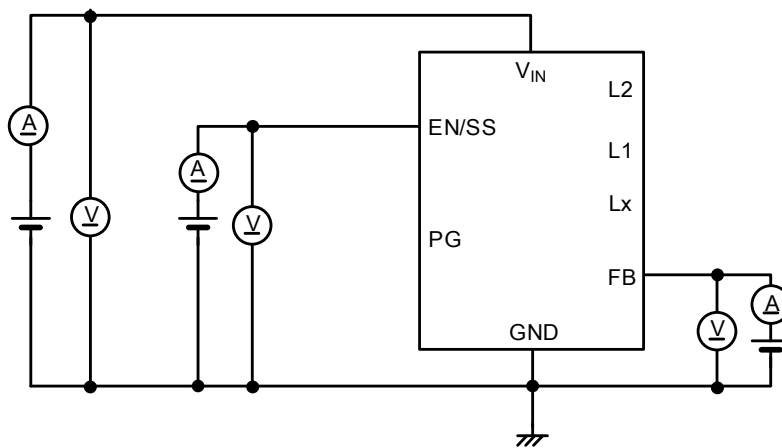
■TEST CIRCUITS

CIRCUIT①
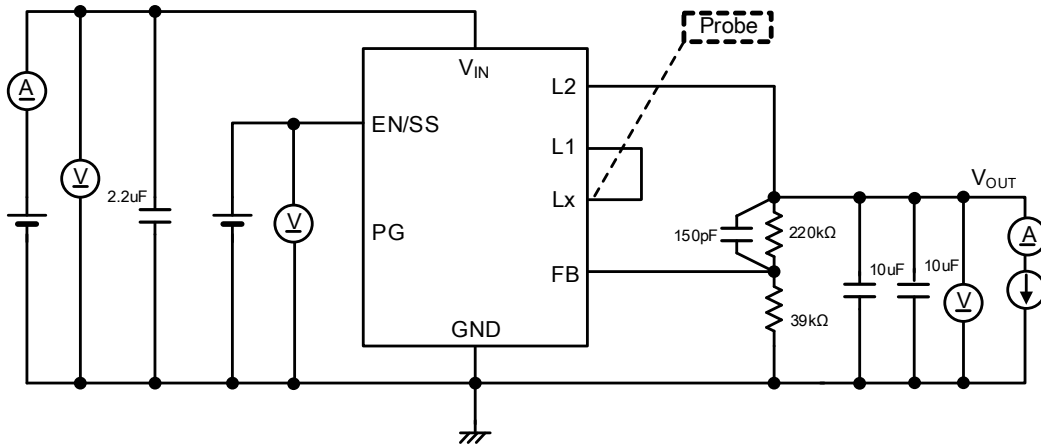
CIRCUIT②
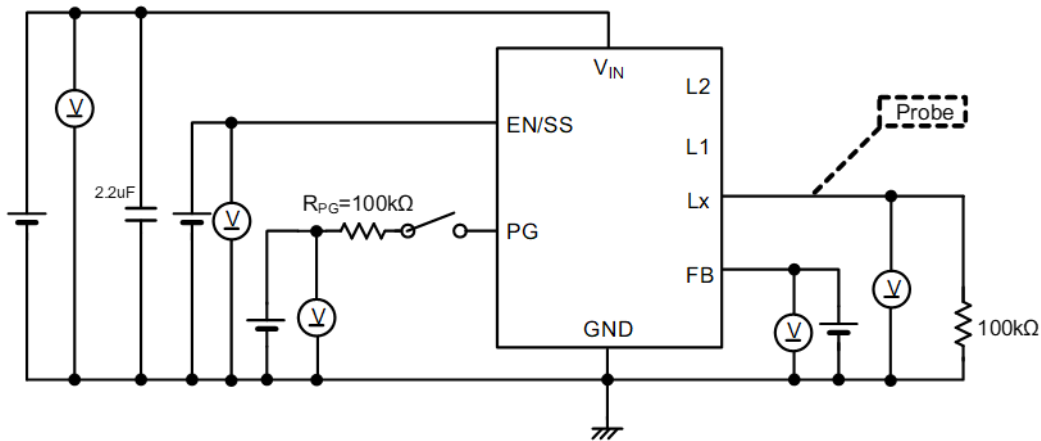
CIRCUIT③
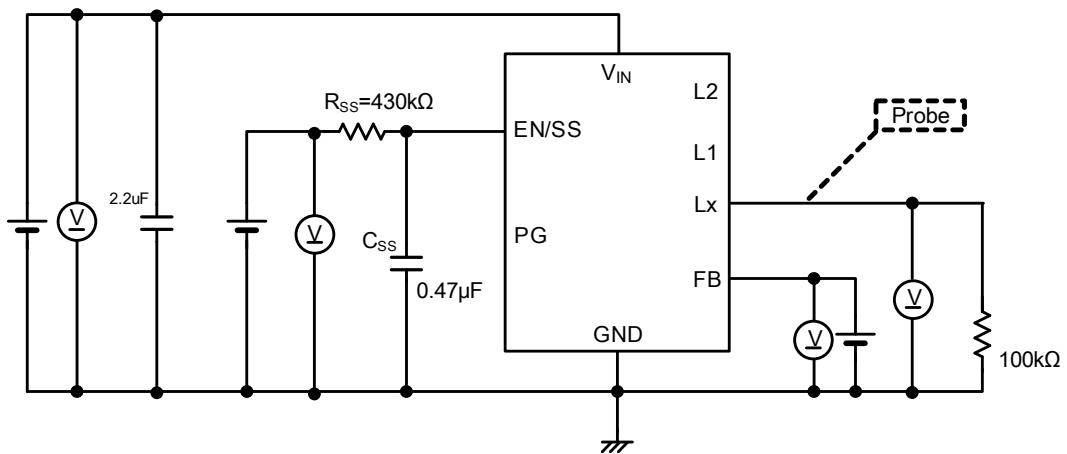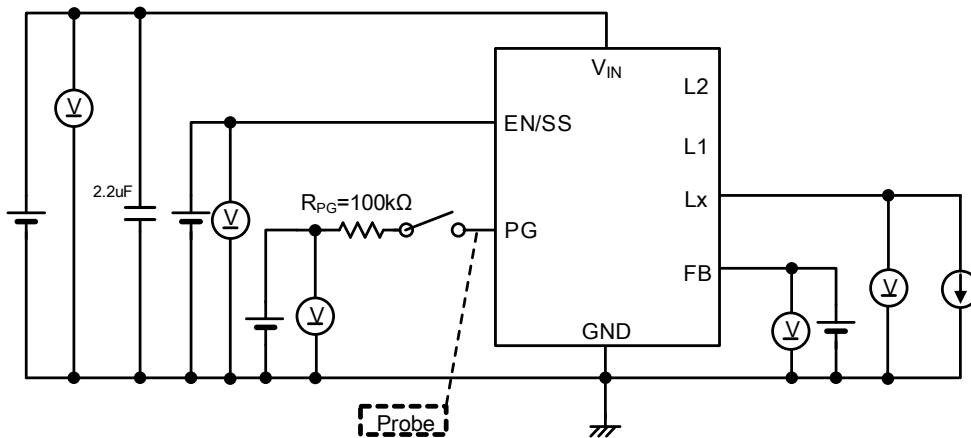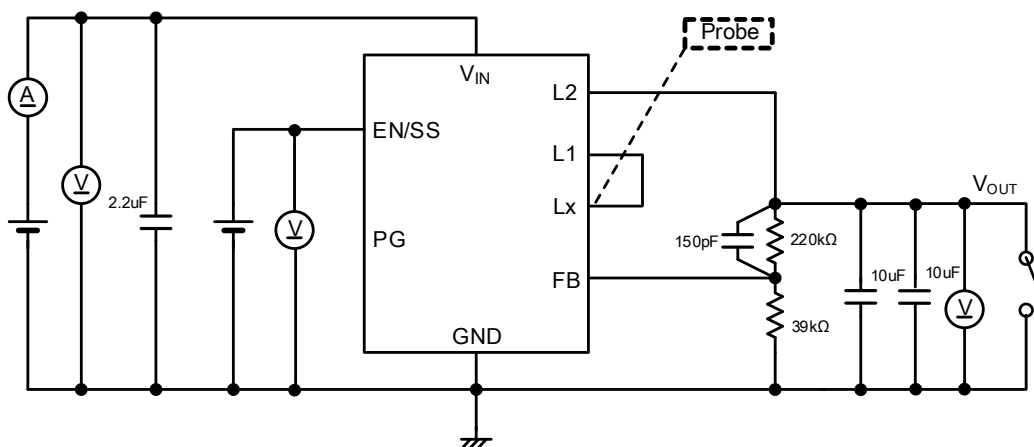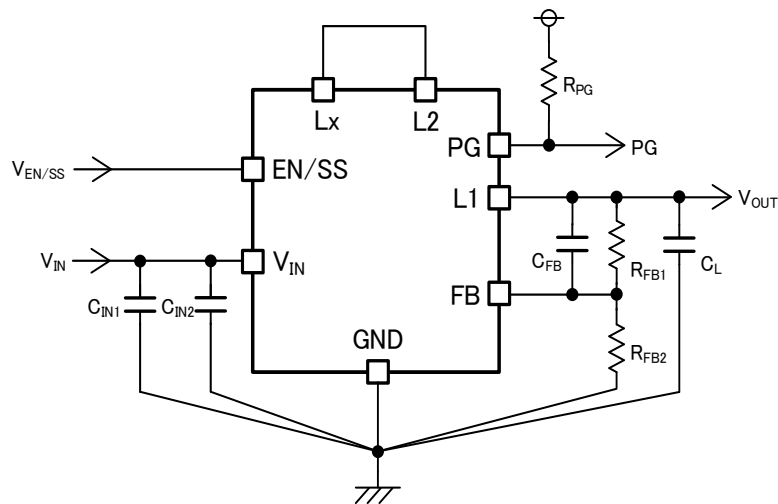
■TEST CIRCUITS(Continued)

CIRCUIT④

CIRCUIT⑤

CIRCUIT⑥

■TYPICAL APPLICATION CIRCUIT / Parts Selection Method

* The inductor is dedicated to this product. Please do not use it for purposes other than this product

【Typical Examples】

	MANUFACTURER	PRODUCT NUMBER	VALUE
C_{IN1}[*1]	TDK	CGA4J3X7R1E225K125AB	2.2µF/25V
C_{IN2}	TDK	CGA3E2X7R1H104K080AA	0.1µF/50V
C_L[*2]	Murata	GRT21BR71A106KE13	10µF/10V 2parallel
	Murata	GRT21BC71C106KE13	10µF/16V 2parallel
	TDK	CGA5L1X7R1C106K160AC	10µF/16V 2parallel

Select parts considering the DC bias characteristics and rated voltage of ceramic capacitors

[*1] For C_{IN1}, use a capacitor with the same or higher effective capacity value as the recommended components.
[*2] For C_L, use a capacitor with the same or higher effective capacity value as the recommended components.
　If a capacitor with a low effective capacity value is used, the output voltage may become unstable.
　However, if large capacity capacitors, such as electrolytic capacitors, are connected in parallel,
　the inrush current during startup could increase or the output could become unstable.

■TYPICAL APPLICATION CIRCUIT / Parts Selection Method (Continued)

< Output voltage setting >

The output voltage can be set by adding an external dividing resistor.　The output voltage is determined by the equation below based on the values of R_{FB1} and R_{FB2}.

$$V_{OUT} = V_{FB} \times (R_{FB1} + R_{FB2}) / R_{FB2}$$
$$\text{With } R_{FB2} \leqq 75k\Omega$$

<C_{FB} setting>

Adjust the value of the phase compensation speed-up capacitor C_{FB} using the equation below.

$$C_{FB} = \frac{1}{2\pi \times fzfb \times R_{FB1}}$$

* It is optimal to adjust so that fzfb = 5kHz.

【Setting Example】

XDL603 series

V_{OUTSET} [V]	R_{FB1} [kΩ]	R_{FB2} [kΩ]	C_{FB} [pF]
1.2	4.7	7.5	6800
3.3	16	4.7	2200
5.0	22	3.9	1500

XDL604 series

V_{OUTSET} [V]	R_{FB1} [kΩ]	R_{FB2} [kΩ]	C_{FB} [pF]
1.2	47	75	680
3.3	160	47	220
5.0	220	39	150

■OPERATIONAL EXPLANATION

The XDL603/XDL604 series consists internally of a reference voltage supply with soft-start function, a ramp wave circuit, an error amp, a PWM comparator, a High side driver FET, a Low side driver FET, a High side buffer circuit, a Low side buffer circuit, a current sense circuit, a phase compensation (Current feedback) circuit, a current limiting circuit, an under voltage lockout (UVLO) circuit, an internal power supply (Local Reg) circuit, a gate clamp (CLAMP) circuit and other elements.

The control method is the current mode control method for handling low ESR ceramic capacitors.
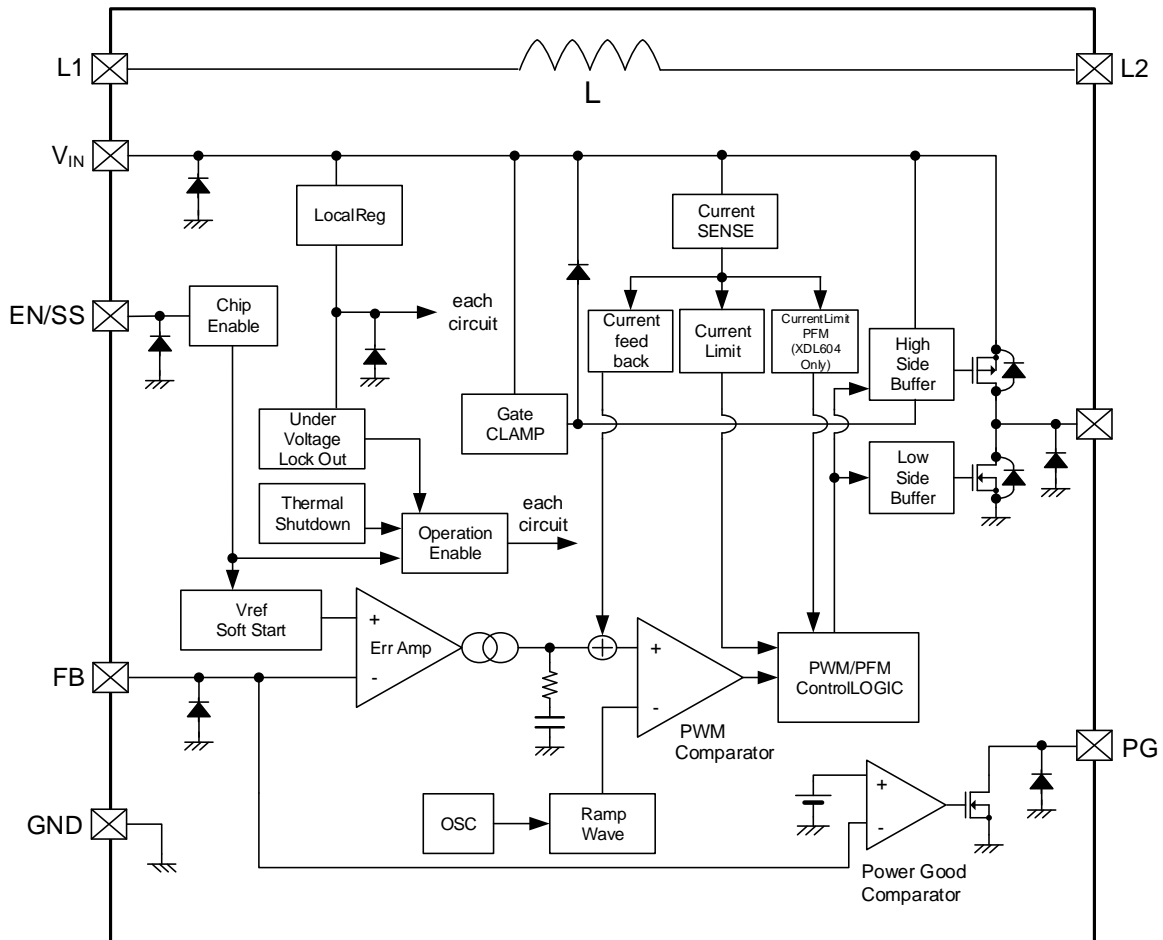
* Diodes inside the circuits are ESD protection diodes and parasitic diodes.

■OPERATIONAL EXPLANATION (Continued)

< Normal Operation >

The standard voltage Vref and FB pin voltage are compared using an error amplifier and then the control signal to which phase compensation has been added to the error amplifier output is input to the PWM comparator. The PWM comparator compares the above control signal and lamp wave to control the duty width during PWM control. Continuously conducting these controls stabilizes the output voltage.

In addition, the current detecting circuit monitors the driver FET current for each switching and modulates the error amplifier output signal into a multiple feedback signal (current feedback circuit). This achieves stable feedback control even when low ESR capacitors, such as ceramic capacitors, are used to stabilize the output voltage.

XDL603 Series

The XDL603 Series (PWM control) performs switching at a set switching frequency f_{OSC} regardless of the output current. At light loads the on time is short and the circuit operates in discontinuous mode, and as the output current increases, the on time becomes longer and the circuit operates in continuous mode.
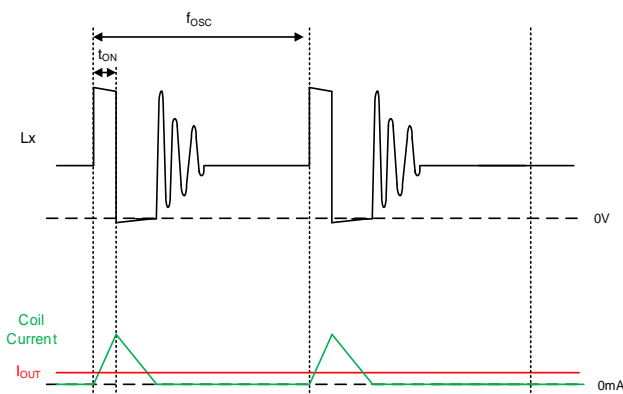
XDL603 series: Example of light load operation
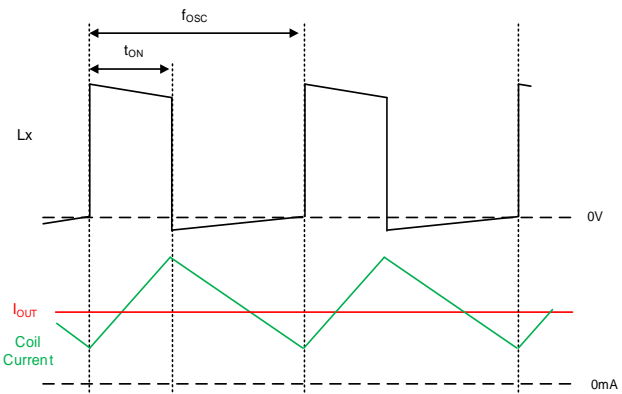
XDL603 series: Example of heavy load operation

XDL604 Series

The XDL604 Series (PWM control) performs switching at a set switching frequency f_{OSC} regardless of the output current. At light loads the on time is short and the circuit operates in discontinuous mode, and as the output current increases, the on time becomes longer and the circuit operates in continuous mode.
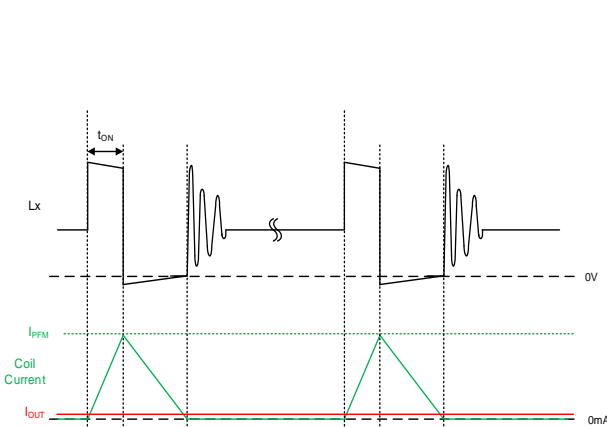
XDL604 series: Example of light load operation
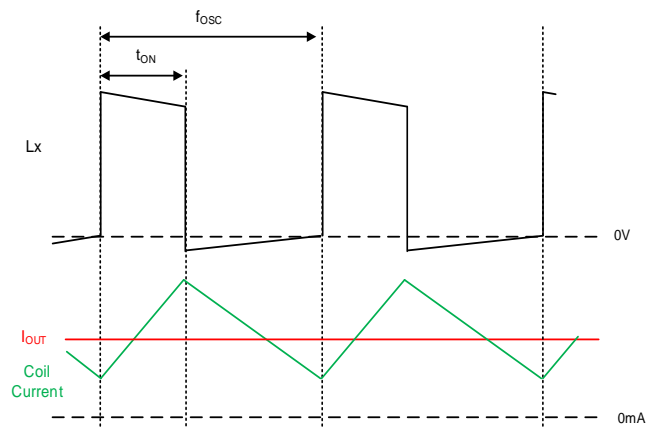
XDL604 series: Example of heavy load operation

< 100% Duty Cycle Mode >

When the dropout voltage is low or there is a transient response, the circuit might change to the 100% Duty cycle mode where the High side driver FET is continuously on.

The 100% Duty cycle mode operation makes it possible to maintain the output current even when the dropout voltage is low such as when the input voltage declines due to cranking, etc.

■OPERATIONAL EXPLANATION (Continued)

< CE Function >

When an "H" voltage ($V_{EN/SSH}$) is input to the EN/SS pin, normal operation is performed after the output voltage is started up by the soft start function, normal operation is performed. When the "L" voltage ($V_{EN/SSL}$) is input to the EN/SS pin, the circuit enters the standby state, the supply current is suppressed to the standby current I_{STB} (TYP. 1.65μA), and the High side driver FET and Low side driver FET are turned off.

< Soft Start Function >

This function gradually starts up the output voltage to suppress the inrush current.

The soft start time is the time until the output voltage from $V_{EN/SSH}$ reaches 90% of the output voltage set value, and when the output voltage increases further, the soft start function is cancelled to switch to normal operation.

Internal Soft Start Time

The internal soft start time (t_{SS1}) is configured so that after the "H" voltage ($V_{EN/SSH}$) is input to the EN/SS pin, the standard voltage connected to the error amplifier increases linearly during the soft-start period. This causes the output voltage to increase proportionally to the standard voltage increase. This operation suppresses the inrush current and smoothly increases the output voltage.
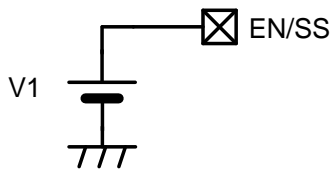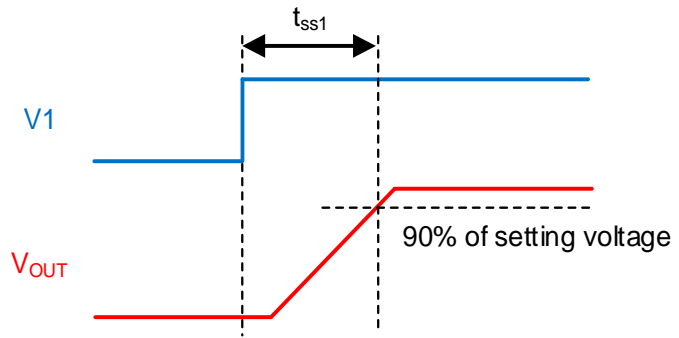
< Internal soft start EN/SS circuit >   < Overview of internal soft start >

External Setting Soft Start Time

The external setting soft start time (t_{SS2}) can adjust the increase speed of the standard voltage in the IC by adjusting the EN/SS pin voltage inclination during startup using externally connected component R_{SS} and C_{SS}. This makes it possible to externally adjust the soft start time.

Soft start time (t_{SS2}) is approximated by the equation below according to values of V1, R_{SS}, and C_{SS}
When t_{SS2} is shorter than t_{SS1}, the output voltage rises at the internal soft start time.

$$t_{SS2} = C_{SS} \times R_{SS} \times \ln (V1 / (V1-1.45V))$$

【Setting Example】
C_{SS} = 0.47μF, R_{SS} = 430kΩ, V1 = 12V
t_{SS2} = 0.47μF x 430kΩ x (ln (12V/(12V-1.45V)) = 26ms
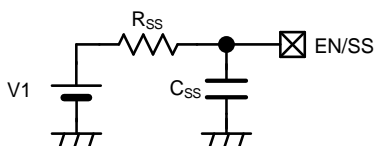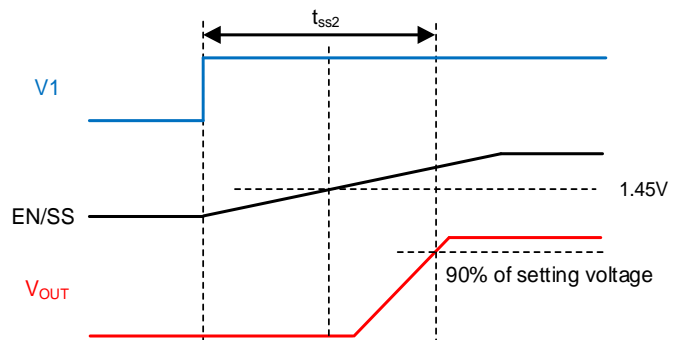
< External soft start EN/SS circuit >   < Overview of external soft start >

■OPERATIONAL EXPLANATION (Continued)

< Power Good >

The output state can be monitored using the power good function. The PG pin is an Nch open drain output, therefore a pull-up resistor (approx. 100kΩ) must be connected to the PG pin.

The pull-up voltage should be 5.5V or less. When not using the power good function, connect the PG terminal to GND or leave it open.

CONDITION		SIGNAL
EN/SS = H	$V_{FB} > V_{PGDET}$	H (High impedance)
	$V_{FB} \leqq V_{PGDET}$	L (Low impedance)
	Thermal Shutdown	L (Low impedance)
	UVLO ($V_{IN} < V_{UVLO1}$)	Undefined State
EN/SS = L	Stand-by	L (Low impedance)

< UVLO Function >

This is a function to monitor the internal power supply and to prevent the output of false pulses from the Lx pin when the output from the internal power supply is unstable at low voltages.As the V_{IN} pin voltage goes down, the internal power supply voltage falls. So the V_{IN} voltage drops, the UVLO function is activated.

When the V_{IN} pin voltage falls below V_{UVLO1} (TYP. 2.7V), the UVLO function is activated, the high side driver FET and low side driver FET are forcibly turned off to prevent false pulse output due to instable operation of the internal circuits. When the V_{IN} pin voltage rises above V_{UVLO2} (TYP. 2.8V), the UVLO function is released, the soft start function activates, and output start operation begins. Stopping by UVLO is not shutdown; only pulse output is stopped and the internal circuits continue to operate.

< Thermal Shutdown Function >

A thermal shutdown (TSD) function is built in for protection from overheating. When the junction temperature reaches the thermal shutdown detection temperature T_{TSD}, the High side driver FET and Low side driver FET are compulsorily turned off.

If the driver FET continues in the off state, the junction temperature declines, and when the junction temperature falls to the thermal shutdown cancel temperature, the thermal shutdown function is cancelled and the soft-start function operates to start up the output voltage.

.

■OPERATIONAL EXPLANATION (Continued)

< Current Limit Function >

The current limiting circuit of the XDL603/XDL604 series monitors the current that flows through the High side driver FET and Low side driver FET, and when over current is detected, the current limiting function activates.

① High side driver FET current limiting

The current in the High side driver FET is detected to equivalently monitor the peak value of the coil current. The High side driver FET current limiting function forcibly turns off the High side driver FET when the peak value of the coil current reaches the High side driver current limit value I_{LIMH}.

High side driver FET current limit value I_{LIMH}=1.1A (TYP.)

②Low side driver FET current limiting

The current in the Low side driver FET is detected to equivalently monitor the bottom value of the coil current. The Low side driver FET current limiting function operates when the High side driver FET current limiting value reaches I_{LIMH}. The Low side driver FET current limiting function prohibits the High side driver FET from turning on in an over current state where the bottom value of the coil current is higher than the Low side driver FET current limit value I_{LIML}.

Low side driver FET current limit value I_{LIM}=0.9A (TYP.)

When the output current increases and reaches the current limit value, the current foldback circuit operates and lowers the output voltage and FB voltage. The I_{LIMH} and I_{LIML} decline accompanying the FB voltage decrease to restrict the output current.

When the overcurrent state is removed, the foldback circuit operation increases the I_{LIMH} and I_{LIML} together with output voltage to return the output to the output voltage set value.

③ Over-current latch (Type A)

Type A turns off the High side and Low side driver FET when state ① or ② continues for t_{LAT} (TYP. 1.0ms). The L_x pin is latch stopped at the GND level (0V).

The latch stopped state only stops the pulse output from the Lx pin; the internal circuitry of the IC continues to operate. To restart after latch stopping, L level and then H level must be input into the EN/SS pin, or V_{IN} pin re-input must be performed (after lowering the voltage below the UVLO detection voltage) to resume operation by Soft-Start.

.

The over current latch function may occasionally be released from the current limit detection state by the effects of ambient noise, and it may also happen that the latch time becomes longer or latching does not take place due to board conditions. For this reason, place the input capacitor as close as possible to the IC.

Type B is an automatic recovery type that performs the operation of ① or ② until the over current state is released.
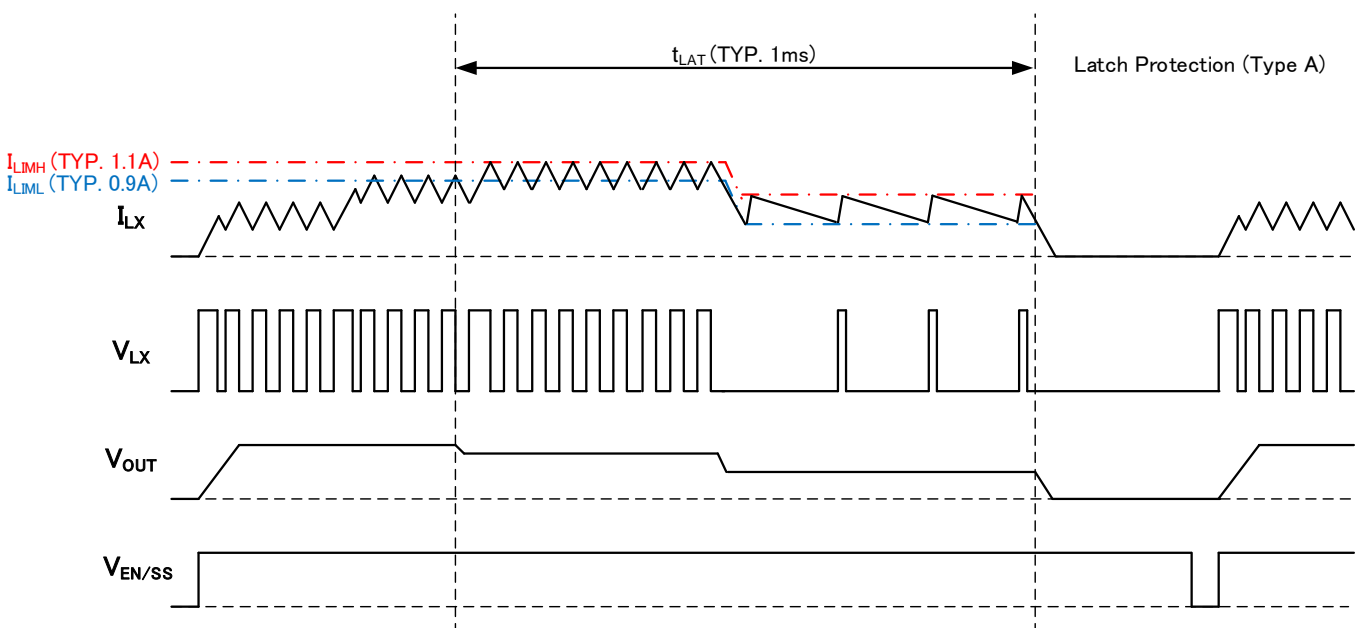
■NOTE ON USE

1) In the case of a temporary and transient voltage drop or voltage rise.
   If the absolute maximum ratings are exceeded, the IC may be deteriorated or destroyed.

   If a voltage exceeding the absolute maximum voltage is applied to the IC due to chattering caused by a mechanical switch or an external surge voltage, please use a protection element such as a TVS and a protection circuit as a countermeasure. Please see the countermeasures from (a) to (d) shown below.

   (a) When voltage exceeding the absolute maximum ratings comes into the VIN pin due to the transient change on the power line, there is a possibility that the IC breaks down in the end.
     To prevent such a failure, please add a TVS between V_{IN} and GND as a countermeasure

   (b) When the input voltage decreases below the output voltage, there is a possibility that an overcurrent will flow in the IC's Internal parasitic diode and exceed the absolute maximum rating of the Lx pin.
     If the current is pulled into the input side by the low impedance between V_{IN} -GND, then countermeasures, such as adding an SBD between V_{OUT}-V_{IN}, should be taken.

   (c) When a negative voltage is applied to the input voltage by a reverse connection or chattering, an overcurrent could flow in the IC's parasitic diode and damage the IC. Take countermeasures, such as adding a reverse touching protection diode

   (d) When a sudden surge of electrical current travels along the V_{OUT} pin and GND due to a short-circuit, electrical resonance of a circuit involving parasitic inductor of cable related to short circuit and an output capacitor (C_L) and impedance such as V_{OUT} line generates a negative voltage exceeding the breakdown voltage and may damage the device.
   Take countermeasures, such as connecting a schottky diode between the V_{OUT} and GND.
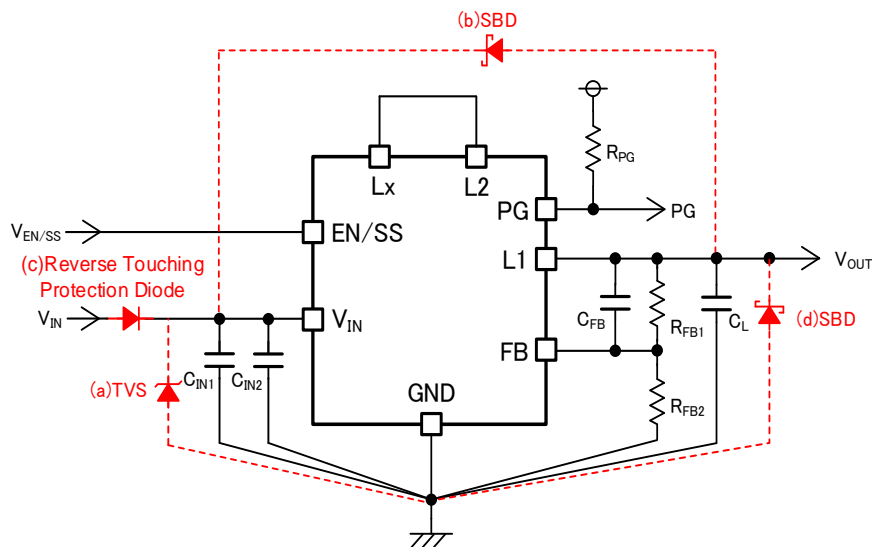
■NOTES ON USE(Continued)

2)   Make sure that the absolute maximum ratings of the external components and of this IC are not exceeded.

3)   The DC/DC converter characteristics depend greatly on the externally connected components as well as on the characteristics of this IC, so refer to the specifications and standard circuit examples of each component when carefully considering which components to select.
     Be especially careful of the capacitor characteristics and use X7R or X5R (EIA standard) ceramic capacitors.
     The capacitance decrease caused by the bias voltage may become large depending on the external size of the capacitor.

4)   The current limit value is the coil current peak value when switching is not conducted.
     The coil current peak value when the actual current limit function begins to operate may exceed the current limit of the electrical characteristics due to the effect of the propagation delay inside the circuit.

5)   When the On time is less than the Min On Time (t_{ONMIN}) and the dropout voltage is large or the load is low, the PWM control operates intermittently and the ripple voltage may become large or the output voltage may become unstable.

6)   The ripple voltage could be increased when switching from discontinuous conduction mode to continuous conduction mode and when switching to 100% Duty cycle.

7)   If the voltage at the EN/SS Pin does not start from 0V but is at the midpoint potential when the power is switched on, the soft start function may not work properly and it may cause larger inrush current and bigger ripple voltages.

8)   The effects of ambient noise and the state of the circuit board may cause release from the current limiting state, and the latch time may lengthen or latch operation may not take place. Please evaluate IC well on customer's PCB.

9)   Torex places an importance on improving our products and their reliability. We request that users incorporate fail safe designs and post aging protection treatment when using Torex products in their systems.

10)   In order to drive the IC normally, supply a stable input voltage to the V_{IN} pin after reducing the AC impedance due to the bypass capacitor. In particular, if the amplitude of the input voltage fluctuates by 2V or more and $\pm 0.1V/\mu$ s or more, there is a possibility that the UVLO function malfunctions due to fluctuations of the internal power supply of the IC.
     In that case, switching is stopped in a protected state that prevents false pulse output from the Lx pin. After that, the soft start function gets started, it shifts to normal operation.
     If the input voltage fluctuates momentarily, take measures such as increasing the input capacitance.

■NOTES ON USE(Continued)

10)   Instructions of pattern layouts

The operation may become unstable due to noise and/or phase lag from the output current when the wire impedance is high, please place the input capacitor(C_{IN1},C_{IN2}) and the output capacitor (C_L) as close to the IC as possible.

(1)   In order to stabilize V_{IN} voltage level, we recommend that a by-pass capacitor (C_{IN2}) be connected as close as possible to the V_{IN} and GND pins. If fluctuation of the V_{IN} potential is expected, please take measures such as increasing input capacitor(C_{IN}).
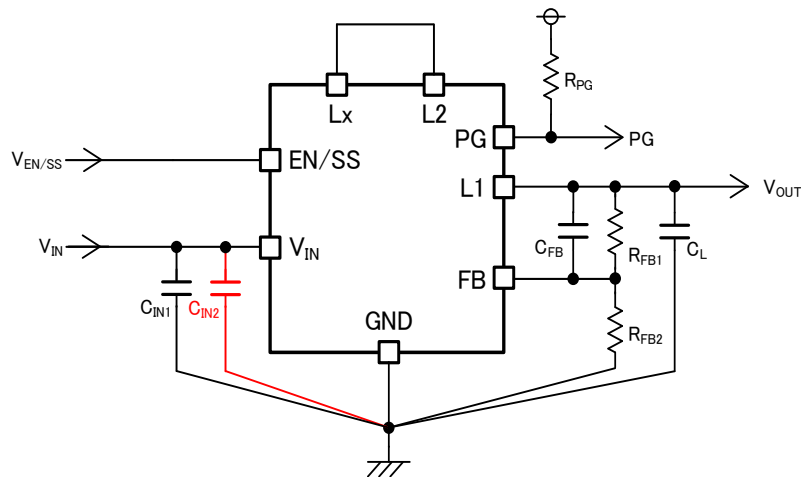
(2)   Please mount each external component as close to the IC as possible.
Please place the external parts on the same side of the PCB as the IC, not on the reverse side of the PCB and elsewhere.

(3)   Wire external components as close to the IC as possible and use thick, short connecting traces to reduce the circuit impedance.

(4)   Make sure that the GND traces are as thick as possible, as variations in ground potential caused by high ground currents at the time of switching may result in instability of the IC.

(5)   This product has a built-in driver FET and inductor, which causes heat generation from the on resistance, so take measures to dissipate the heat when necessary.

■NOTES ON USE(Continued)

＜Recommended Pattern Layout＞
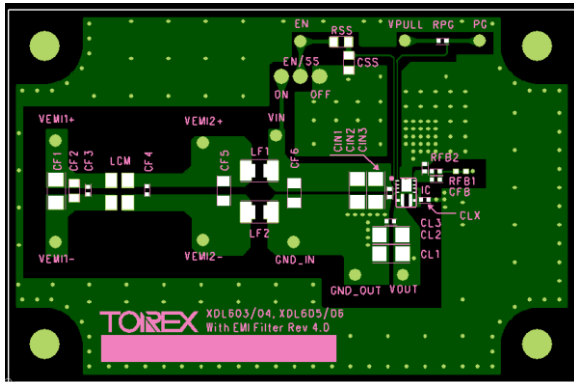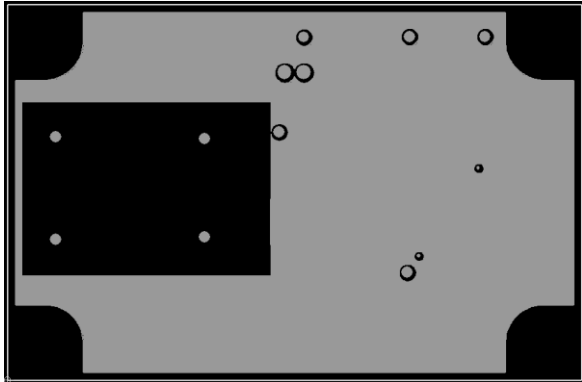
Layer 1

Layer 2

Layer 3
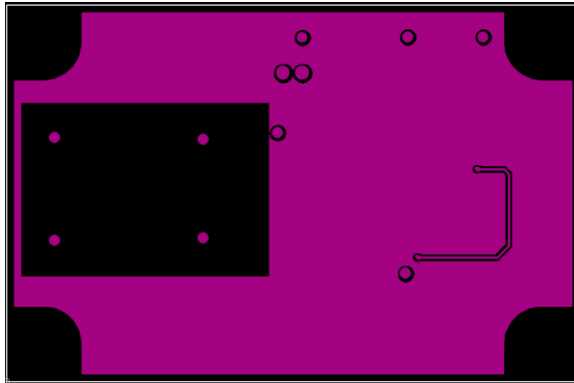
Layer 4
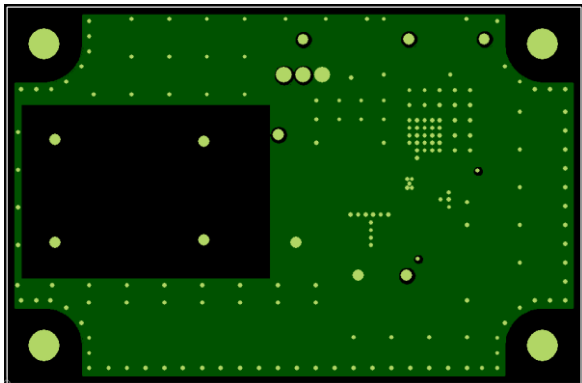
■TYPICAL PERFORMANCE CHARACTERISTICS

(1) Efficiency vs. Output current
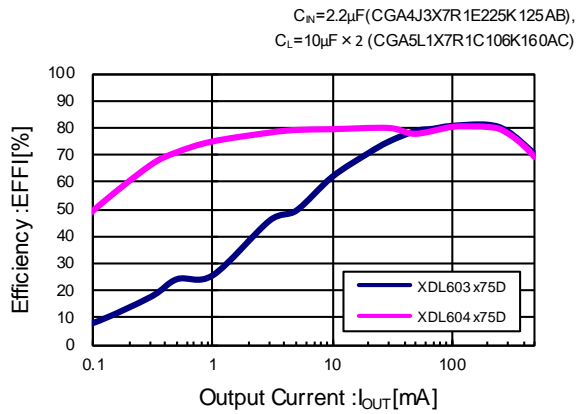
XDL603x75D/XDL604x75D
(V_{IN}=8V, V_{OUT}=3.3V)

C_{IN}=2.2μF(CGA4J3X7R1E225K125AB),
C_L=10μF × 2 (CGA5L1X7R1C106K160AC)

XDL603x75D/XDL604x75D
(V_{IN}=12V, V_{OUT}=3.3V)

C_{IN}=2.2μF(CGA4J3X7R1E225K125AB),
C_L=10μF × 2 (CGA5L1X7R1C106K160AC)

XDL603x75D/XDL604x75D
(V_{IN}=12V, V_{OUT}=5V)

C_{IN}=2.2μF(CGA4J3X7R1E225K125AB),
C_L=10μF×2 (CGA5L1X7R1C106K160AC)
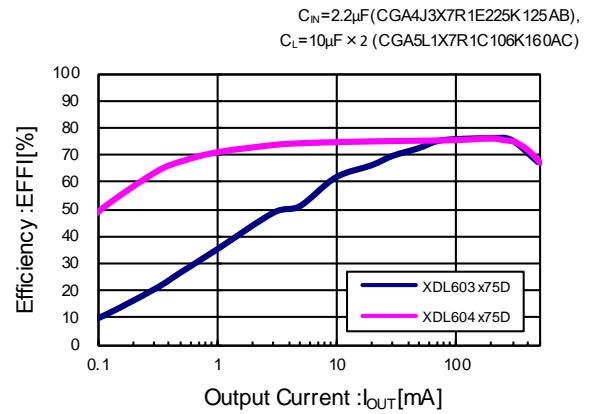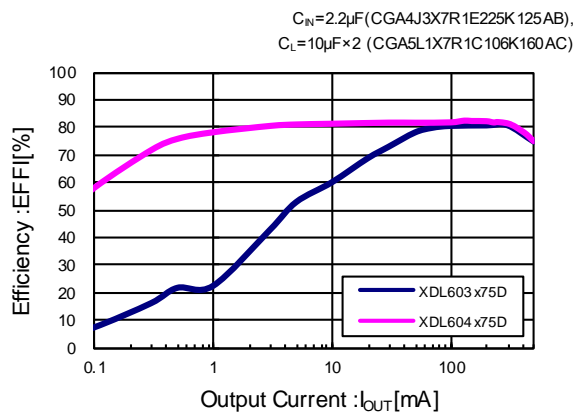
(2) Output Voltage vs. Output Current
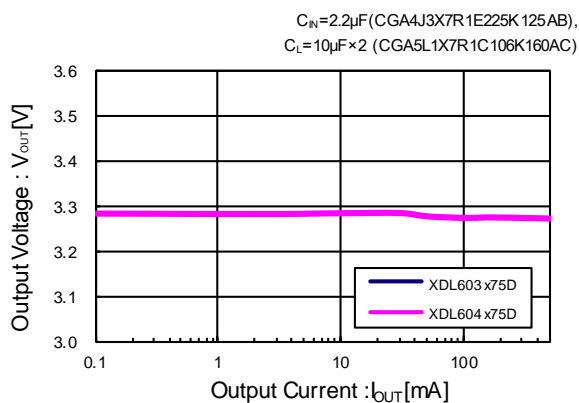
XDL603x75D/XDL604x75D
(V_{IN}=12V, V_{OUT}=3.3V)

C_{IN}=2.2μF(CGA4J3X7R1E225K125AB),
C_L=10μF×2 (CGA5L1X7R1C106K160AC)

XDL603x75D/XDL604x75D
(V_{IN}=12V, V_{OUT}=5V)

C_{IN}=2.2μF(CGA4J3X7R1E225K125AB),
C_L=10μF×2 (CGA5L1X7R1C106K160AC)
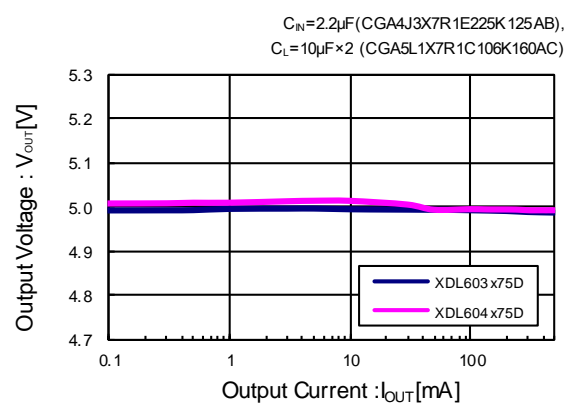
■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(3) Ripple Voltage vs. Output Current

XDL603x75D/XDL604x75D
(V_{IN}=12V, V_{OUT}=5V)
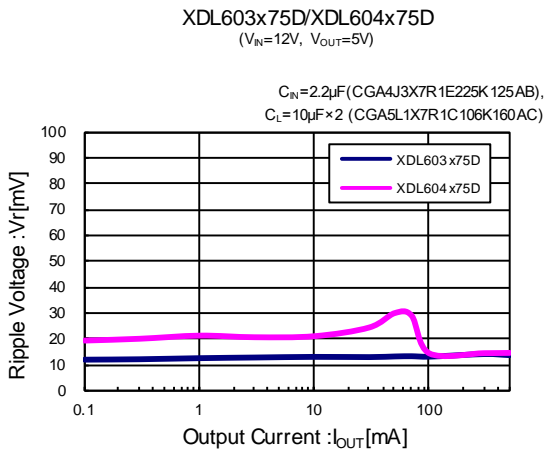
(4) FB Voltage vs. Ambient Temperature

XDL603x75D/XDL604x75D
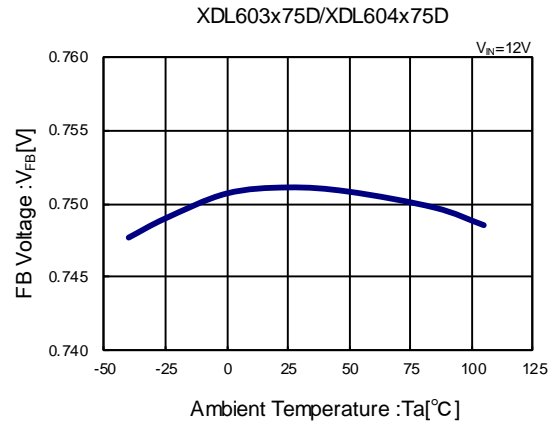
(5) UVLO Voltage vs. Ambient Temperature

XDL603x75D/XDL604x75D
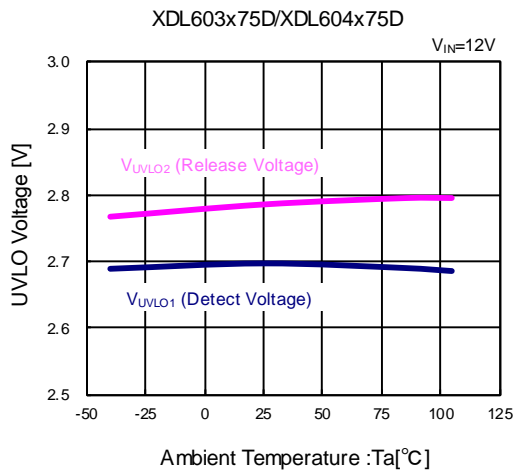
(6) Oscillation Frequency vs. Ambient Temperature

XDL603x75D/XDL604x75D
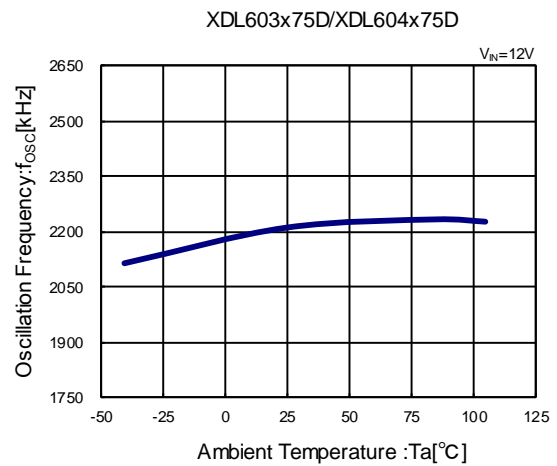
(7) Stand-by Current vs. Ambient Temperature

XDL603x75D/XDL604x75D
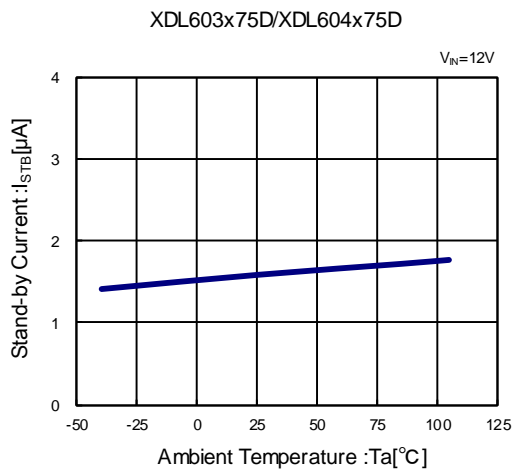
(8) Lx SW On Resistance vs. Ambient Temperature

XDL603x75D/XDL604x75D
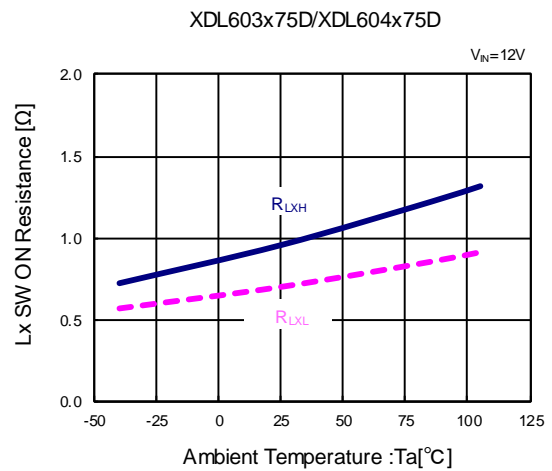
■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(9) Quiescent Current vs. Ambient Temperature

XDL603x75D/XDL604x75D

(10) Internal Soft-Start Time vs. Ambient Temperature

XDL603x75D/XDL604x75D
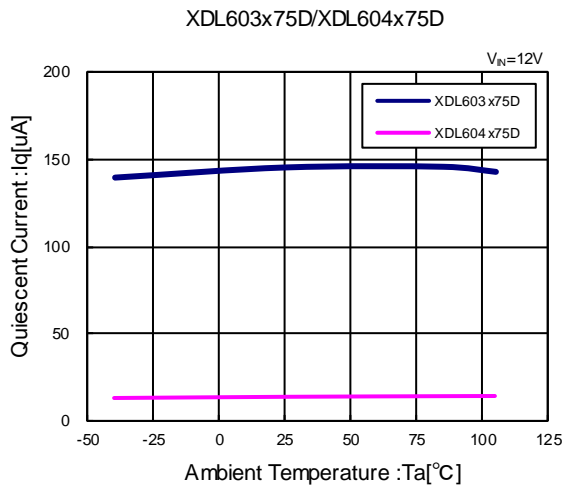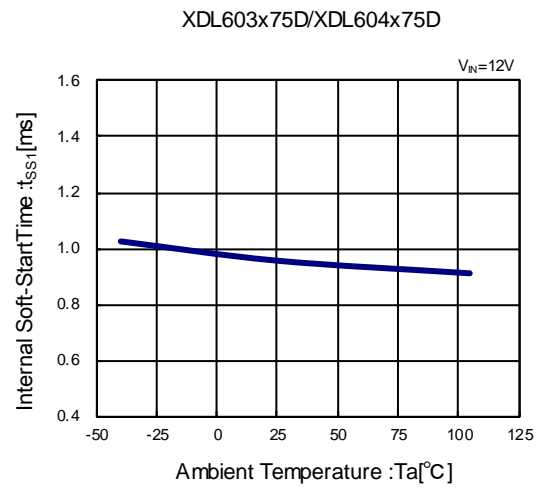
(11) External Soft-Start Time vs. Ambient Temperature

XDL603x75D/XDL604x75D
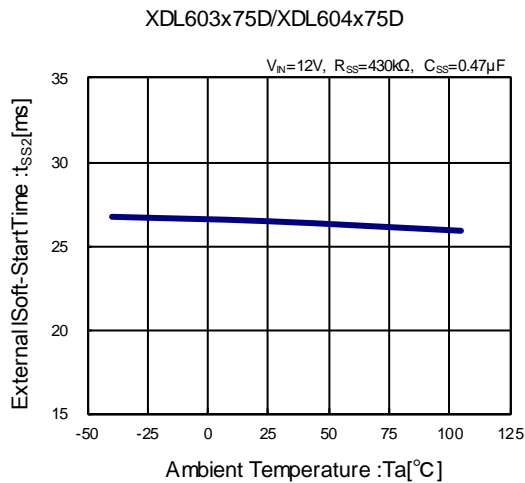
(12) PFM Switch Current vs. Ambient Temperature

XDL604x75D
(V_{IN}=12V, V_{OUT}=5V)
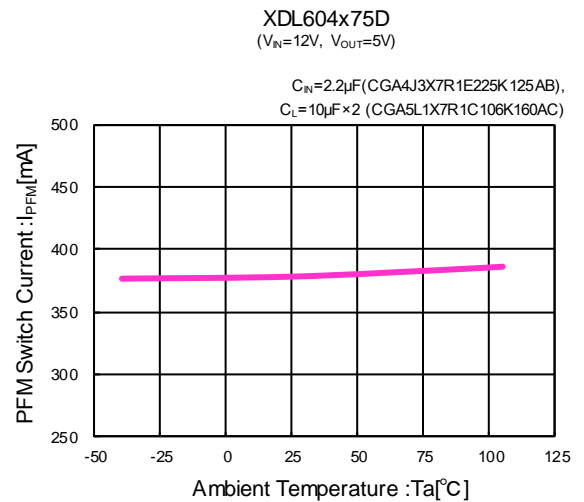
(13) PG Detect Voltage vs. Ambient Temperature

XDL603x75D/XDL604x75D
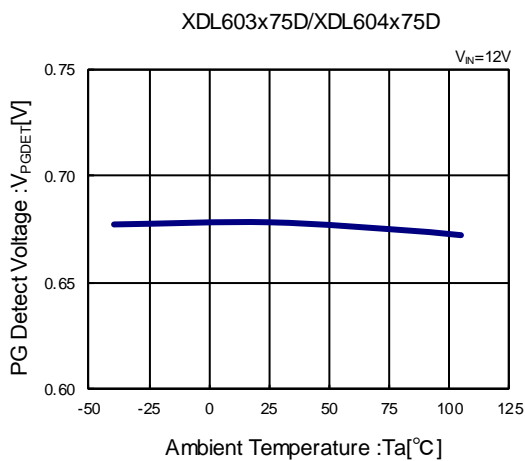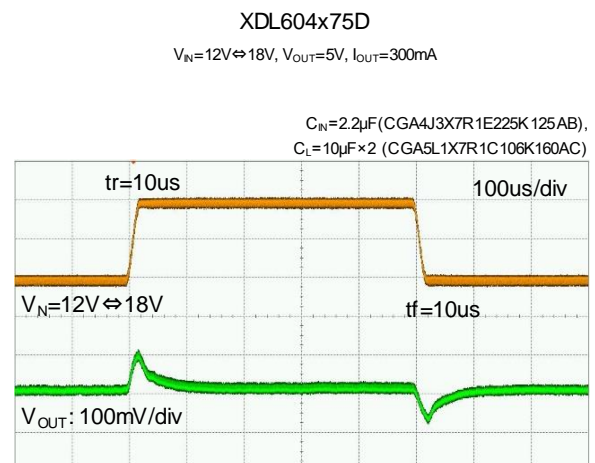

(14) PG Output Voltage vs. Ambient Temperature

XDL603x75D/XDL604x75D
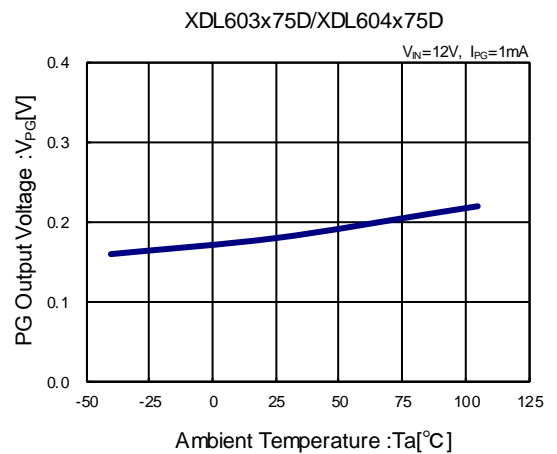
■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(15) EN/SS Voltage vs. Ambient Temperature

XDL603x75D/XDL604x75D

V_{IN}=12V

EN/SS "H"
EN/SS "L"

EN/SS Voltage [V]

Ambient Temperature :Ta[°C]
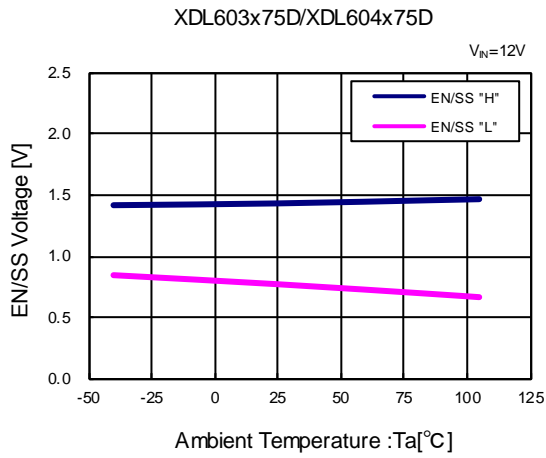
(16) V_{IN}-V_{OUT} Operation Area

XDL603/XDL604

Input Voltage:V_{IN}[V]

Operation Area

Output Voltage:V_{OUT}[V]
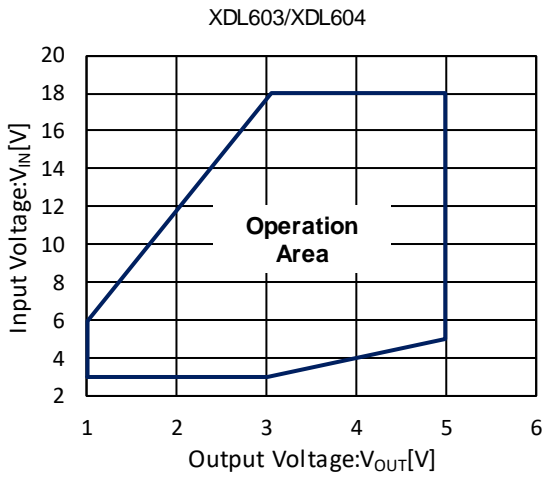
(17) Output Current Operation Area

XDL603/XDL604 (V_{OUT}=3.3V)

θja=46°C/W

V_{IN}=12V

Output Current:I_{OUT}[mA]

Operation Area

Ambient Temperature :Ta[°C]

XDL603/XDL604 (V_{OUT}=5V)

θja=46°C/W

V_{IN}=12V

Output Current:I_{OUT}[mA]

Operation Area

Ambient Temperature :Ta[°C]
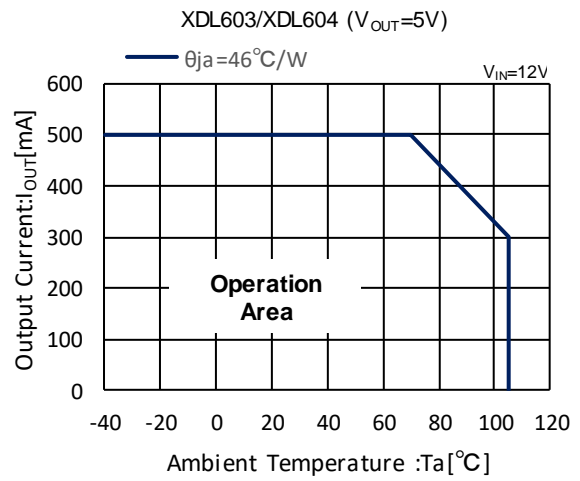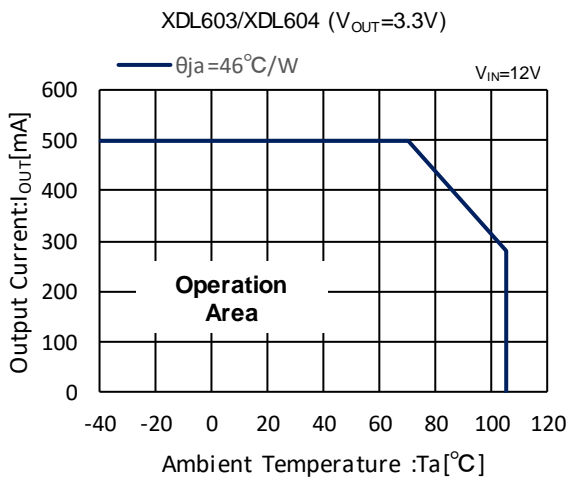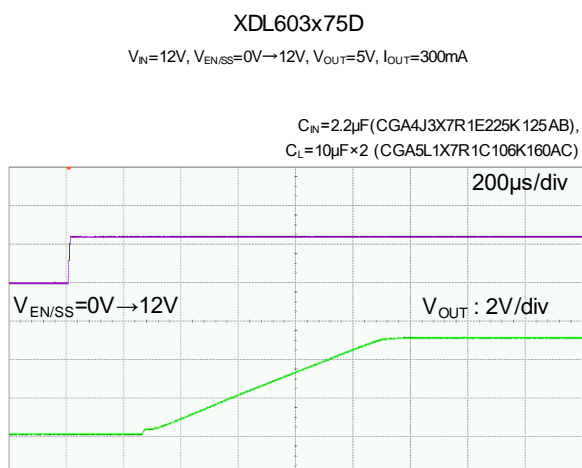
■TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

(18) Load Transient Response

XDL603x75D

V_{IN}=12V, V_{OUT}=5.0V, I_{OUT}=1mA⇔300mA

C_{IN}=2.2μF(CGA4J3X7R1E225K125AB),
C_L=10μF×2 (CGA5L1X7R1C106K160AC)

tr=10us   200us/div

I_{OUT}=1mA ⇔ 300mA   tf=10us

V_{OUT}: 100mV/div

XDL604x75D

V_{IN}=12V, V_{OUT}=5.0V, I_{OUT}=1mA⇔300mA

C_{IN}=2.2μF(CGA4J3X7R1E225K125AB),
C_L=10μF×2 (CGA5L1X7R1C106K160AC)

tr=10us   200us/div

I_{OUT}=1mA ⇔ 300mA   tf=10us

V_{OUT}: 100mV/div
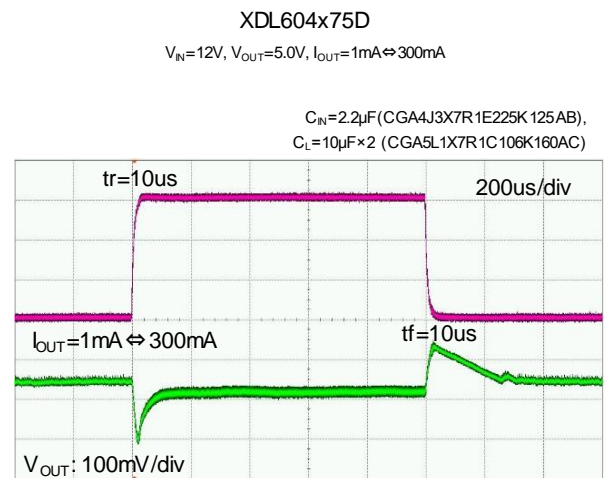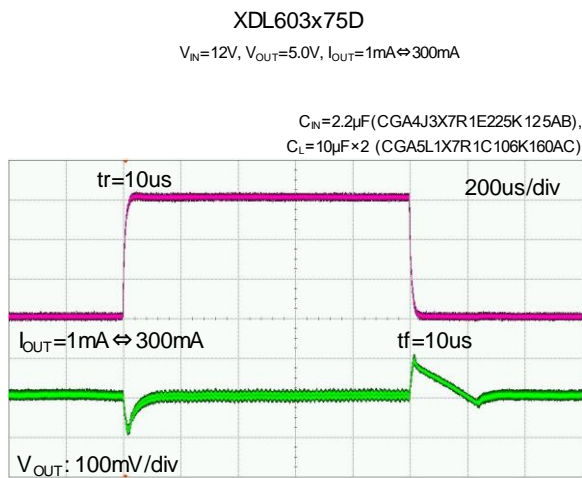
(19) Input Transient Response

XDL603x75D

V_{IN}=12V⇔18V, V_{OUT}=5V, I_{OUT}=300mA

C_{IN}=2.2μF(CGA4J3X7R1E225K125AB),
C_L=10μF×2 (CGA5L1X7R1C106K160AC)

tr=10us   100us/div

V_{IN}=12V ⇔ 18V   tf=10us

V_{OUT}: 100mV/div

XDL604x75D

V_{IN}=12V⇔18V, V_{OUT}=5V, I_{OUT}=300mA

C_{IN}=2.2μF(CGA4J3X7R1E225K125AB),
C_L=10μF×2 (CGA5L1X7R1C106K160AC)

tr=10us   100us/div

V_{IN}=12V ⇔ 18V   tf=10us

V_{OUT}: 100mV/div
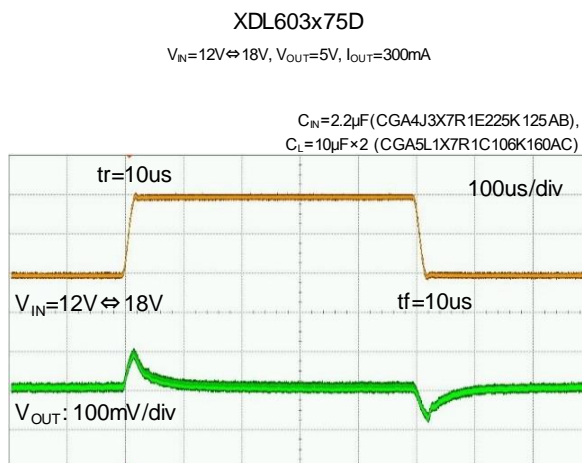
(20) EN/SS Rising Response

XDL603x75D

V_{IN}=12V, $V_{EN/SS}$=0V→12V, V_{OUT}=5V, I_{OUT}=300mA

C_{IN}=2.2μF(CGA4J3X7R1E225K125AB),
C_L=10μF×2 (CGA5L1X7R1C106K160AC)

200μs/div

$V_{EN/SS}$=0V→12V V_{OUT} : 2V/div

XDL604x75D

V_{IN}=12V, $V_{EN/SS}$=0V→12V, V_{OUT}=5V, I_{OUT}=300mA

C_{IN}=2.2μF(CGA4J3X7R1E225K125AB),
C_L=10μF×2 (CGA5L1X7R1C106K160AC)

200μs/div

$V_{EN/SS}$=0V→12V V_{OUT} : 2V/div

■PACKAGING INFORMATION

For the latest package information go to, www.torexsemi.com/technical-support/packages

PACKAGE	OUTLINE / LAND PATTERN	THERMAL CHARACTERISTICS
DFN3625-11B	DFN3625-11B PKG	DFN3625-11B Power Dissipation

■MARKING RULE

● DFN3625-11B

DFN3625-11B
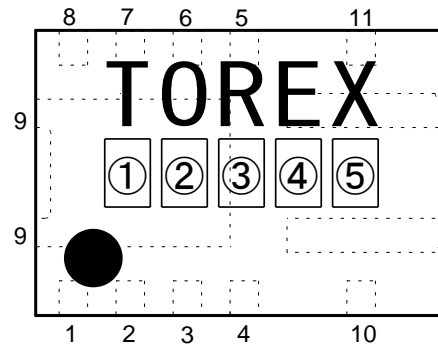
① represents product series

MARK	PRODUCT SERIES
H	XDL603****82-Q
K	XDL604****82-Q

② represents Type

MARK	Type	PRODUCT SERIES
A	A	XDL60*A**82-Q
B	B	XDL60*B**82-Q

③ represents FB Voltage

MARK	FB(V)	PRODUCT SERIES
0	0.75	XDL60**75*82-Q

④,⑤ represents production lot number 01～09、0A～0Z、11～9Z、A1～A9、AA～AZ、B1～ZZ in order
 (G, I, J, O, Q, W excluded*)No Character inversion used

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