**Features** 



## 12V MOSFET Drivers with Output Disable

# for Multiple Phase Synchronous-Rectified Buck Converter

## General Description

The uP1951 is a quad, high voltage MOSFET driver optimized for driving four N-Channel MOSFETs in a dual synchronous-rectified buck converter. Each driver is capable of driving a 5000pF load with 30ns transition time. This device combined with uPI multi-phase buck PWM controller forms a complete core voltage regulator for advanced micro-processors.

The uP1951 features adaptive anti-shoot-through protection that prevents cross-conduction of the external MOSFET while maintaining minimum deadtime for optimized efficiency.

Both gate drives are turned off by pulling low OD# pin or high-impedance at PWM pin, preventing rapid output capacitor discharge during system shutdowns.

Other feature is supply input under voltage lockout. The uP1951 is available in thermal enhanced WQFN3x3-16L package.

## Ordering Information

Order Number	Package Type	Top Marking
uP1951PQDD	WQFN3x3-16L	uP1951P

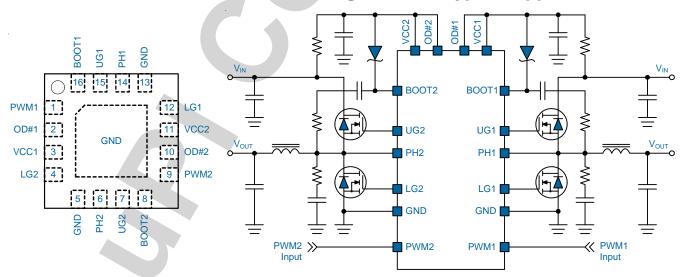
Note: uPI products are compatible with the current IPC/ JEDEC J-STD-020 requirement. They are halogen-free, RoHS compliant and 100% matte tin (Sn) plating that are suitable for use in SnPb or Pb-free soldering processes.

- All-In-One Synchronous Buck Drivers
- Bootstrapped High-Side Driver
- Adaptive Anti-Shoot-Through Protection Circuitry
- Tri-State Input for Bridge Shutdown
- Output Disable Control Turns Off both MOSFETs
- Under Voltage Lockout for Supply Input
- WQFN3x3-16L Package
- RoHS Compliant and Halogen Free

## Applications

- □ Core Voltage Supplies for Desktop, Motherboard CPUs
- ☐ High Frequency Low Profile DC/DC Converters
  - High Current Low Voltage DC/DC Converters

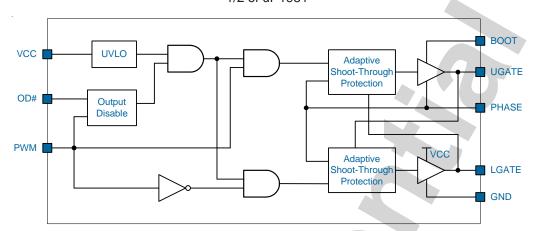
## Pin Configuration & Typical Application Circuit





# Functional Block Diagram

#### 1/2 of uP1951



# Functional Pin Description

Pin Name	Pin Function			
BOOTx	<b>Bootstrap Supply</b> for the floating upper gate driver. Connect the bootstrap capacitor $C_{\text{BOOT}}$ between BOOTx pin and the PHx pin to form a bootstrap circuit. The bootstrap capacitor provides the charge to turn on the upper MOSFET. Ensure that $C_{\text{BOOT}}$ is placed near the IC.			
PWMx	PWM Input. This pin receives logic level input and controls the driver outputs.			
OD#x	Output Disable. This pin disables normal operation and forces both UGx and LGx off when it is pulled low.			
VCCx	<b>Supply Voltage for the IC.</b> This pin provides bias voltage for the IC. Connect these pins to 12V voltage source and bypass it with an R/C filter. Note that VCC1 and VCC2 are two independent pins, and they are NOT internally connected together. Therefore user must connect both VCC1 and VCC2 to 12V voltage source to supply power if both of the two output channels are used.			
LGx	<b>Lower Gate Driver Output.</b> Connect this pin to the gate of lower MOSFET. This pin is monitored by the adaptive shoot-through protection circuitry to determine when the lower MOSFET has turned off.			
GND	Ground for the IC. All voltages levels are measured with respect to this pin.			
PHx	<b>PHASE Switch Node.</b> Connect this pin to the source of the upper MOSFET and the drain of the lower MOSFET. This pin is used as the sink for the UGx driver. This pin is also monitored by the adaptive shoot-through protection circuitry to determine when the upper MOSFET has turned off. A Schottky diode between this pin and ground is recommended to reduce negative transient voltage which is common in a power supply system.			
UGx	<b>Upper Gate Driver Output.</b> Connect this pin to the gate of upper MOSFET. This pin is monitored by the adaptive shoot-through protection circuitry to determine when the upper MOSFET has turned off.			
Exposed Pad	Ground for the IC. The exposed pad should be well soldered to PCB for effective heat conduction.			



### Functional Description

The uP1951 is a quad, high voltage MOSFET driver optimized for driving four N-Channel MOSFETs in a dual synchronous-rectified buck converter. Each driver is capable of driving a 5000pF load with 30ns transition time. This device combined with uPI multi-phase buck PWM controller forms a complete core voltage regulator for advanced micro-processors.

The uP1951 features adaptive anti-shoot-through protection that prevents cross-conduction of the external MOSFET while maintains minimum deadtime for optimized efficiency.

Both gate drives are turned off by pulling low OD# pin or high-impedance at PWM pin, preventing rapid output capacitor discharge during system shutdowns.

Other feature is supply input under voltage lockout. The uP1951 is available in thermal enhanced WQFN3x3-16L package.

#### **Output Disable**

Logic low of OD#x disables the gate drivers and keep both output low. Tie the OD#x pin to controller power directly if the output disable function is not used.

#### **PWM Input**

The PWMx pin is a tri-state input. Logic high turns on the high-side gate driver and turns off the low side gate driver once the POR of VCC is granted and OD#x is kept high. Logic low turns off the high side gate driver and turns on the low side gate driver.

High impedance input at PWMx pin will keep both highside and low-side gate drivers low and turns off both MOSFETs. The PWMx pin voltage is kept around 2.0V by internal bias circuit when floating.

#### **Low Side Driver**

The low-side driver is designed to drive a ground-referenced N-Channel MOSFET. The bias to the low-side driver is internally connected to VCCx supply and GND. The low-side driver output is out of phase with the PWM input when it is enabled. The low side driver is held low if the OD#x pin is pulled low or high-impedance at PWM pin.

#### **High-Side Driver**

The high-side driver is designed to drive a floating N-Channel MOSFET. The bias voltage to the high-side driver internally connected to BOOTx and PHx pins. An external bootstrap supply circuit that is connected between BOOTx and PHx pins provides the bias current for the high-side gate driver.

The bootstrap capacitor  $C_{\text{BOOT}}$  is charged to  $V_{\text{CC}}$  when PHx pin is grounded by turning on the low-side MOSFET. The PHx rises to  $V_{\text{IN}}$  when the high-side MOSFET is turned on, forcing the BOOT pin voltage to  $V_{\text{IN}} + V_{\text{CC}}$  that provides voltage to hold the high-side MOSFET on.

The high-side gate driver output is in phase with the PWMx input when it is enabled. The high-side driver is held low if the OD#x pin is pulled low or high-impedance at PWM pin.

#### **Adaptive Shoot Through Protection**

The adaptive shoot-through circuit prevents the high-side and low-side MOSFETs from being ON simultaneously and conducting destructive large current. It is done by turning on one MOSFET only after the other MOSFET is off already with adequately delay time.

At the high-side off edge, UGx and PHx voltages are monitored for anti-shoot-through protection. The uP1951 will not begin to output low-side driver high until both ( $V_{\text{UGx}}$  -  $V_{\text{PHx}}$ ) and  $V_{\text{PHx}}$  are lower than 1.2V, making sure the high-side MOSFET is turned off completely.

At the low-side off edge, LGx voltage is monitored for antishoot-through protection. The uP1951 will not begin to output high-side driver high until  $V_{LGx}$  is lower than 1.2V, making sure the low-side MOSFET is turned off completely.



(Note 1) Supply Input Voltage, VCC12		Absolute Maximum Rating
BOOTx to PHx	(Note 1)	
PHx to GND	Supply Input Voltage, VCC12	
DC		
Comparison   Co		0.7\/+45\/
BOOTx to GND		
DC		6V to 30V
< 200ns	ROOTX to GND	0.21/40//0012 + 15//
UGx to PHx		
$\begin{array}{c} DC \\ < 200 \text{ns} \\ > 5 \text{V to (BOOTx - PHx + 0.3 \text{V})} \\ \\ ZOONS \\ > COONS \\ >$		
Comparison		0.2\/ to \/POOTy \ PHy +0.2\/\
LGx to GND  DC	<200ns	
DC		
<200ns		
PWMx ————————————————————————————————————		
$ \begin{array}{c} \text{OD#x} & -0.3 \text{V to (VCCx} + 0.3) \text{V} \\ \text{Storage Temperature Range} & -65^{\circ}\text{C to} + 150^{\circ}\text{C} \\ \text{Junction Temperature} & -150^{\circ}\text{C} \\ \text{Lead Temperature (Soldering, 10 sec)} & -260^{\circ}\text{C} \\ \text{ESD Rating (Note 2)} & -2k \text{V} \\ \text{MM (Machine Mode)} & -2k \text{V} \\ \text{MM (Machine Mode)} & -200 \text{V} \\ \hline                                  $		
Storage Temperature Range		
Junction Temperature       150°C         Lead Temperature (Soldering, 10 sec)       260°C         ESD Rating (Note 2)       2kV         MM (Machine Mode)       2kV         MM (Machine Mode)       200V         Thermal Information         Package Thermal Resistance (Note 3)         WQFN3x3 - 16L q <sub>Ja</sub> 68°C/W         WQFN3x3 - 16L q <sub>Ja</sub> 6°C/W         Power Dissipation, P <sub>D</sub> @ T <sub>A</sub> = 25°C       1.47W         Recommended Operation Conditions         (Note 4)       70°C to +125°C         Operating Junction Temperature Range       40°C to +85°C         Supply Input Voltage, V <sub>CC</sub> 10.8V to 13.2V         Note 1. Stresses listed as the above Absolute Maximum Ratings may cause permanent damage to the device.		
Lead Temperature (Soldering, 10 sec)       260°C         ESD Rating (Note 2)       2kV         MM (Human Body Mode)       200V         Thermal Information         Package Thermal Resistance (Note 3)         WQFN3x3 - 16L q <sub>JA</sub> 68°C/W         WQFN3x3 - 16L q <sub>JC</sub> 6°C/W         Power Dissipation, P <sub>D</sub> @ T <sub>A</sub> = 25°C         WQFN3x3 - 16L       1.47W         Recommended Operation Conditions         (Note 4)       -40°C to +125°C         Operating Ambient Temperature Range       -40°C to +85°C         Supply Input Voltage, V <sub>CC</sub> 10.8V to 13.2V         Note 1. Stresses listed as the above Absolute Maximum Ratings may cause permanent damage to the device.		
ESD Rating (Note 2)  HBM (Human Body Mode) ————————————————————————————————————		
HBM (Human Body Mode)	The state of the s	260°C
Package Thermal Resistance (Note 3) $WQFN3x3 - 16L \ q_{JA}$	ESD Rating (Note 2)	
Package Thermal Resistance (Note 3) $WQFN3x3 - 16L \ q_{JA}$	HBM (Human Body Mode)	2kV
Package Thermal Resistance (Note 3)	MIM (Machine Mode)	200V
Package Thermal Resistance (Note 3)		Thermal Information
$\begin{array}{cccccccccccccccccccccccccccccccccccc$		Thermal information
Power Dissipation, $P_D @ T_A = 25^{\circ}C$ WQFN3x3 - 16L	Package Thermal Resistance (Note 3)	
Power Dissipation, $P_D @ T_A = 25^{\circ}C$ WQFN3x3 - 16L	WQFN3x3 - 16L q <sub>JA</sub>	68°C/W
WQFN3x3 - 16L	WQFN3x3 - 16L q <sub>JC</sub>	6°C/W
(Note 4) Operating Junction Temperature Range	Power Dissipation, $P_D @ T_A = 25^{\circ}C$	
(Note 4) Operating Junction Temperature Range	WQFN3x3 - 16L	1.47W
(Note 4) Operating Junction Temperature Range	Recomm	ended Operation Conditions
Operating Junction Temperature Range		criaca operation containing
Operating Ambient Temperature Range		
Supply Input Voltage, V <sub>cc</sub> 10.8V to 13.2V  Note 1. Stresses listed as the above <i>Absolute Maximum Ratings</i> may cause permanent damage to the device.		
Note 1. Stresses listed as the above Absolute Maximum Ratings may cause permanent damage to the device.	Supply Input Voltage, V	10.8V to 13.2V
These are for stress ratings. Functional operation of the device at these of any other conditions beyond		
those indicated in the operational sections of the specifications is not implied. Exposure to absolute		
maximum rating conditions for extended periods may remain possibility to affect device reliability.		

- **Note 2.** Devices are ESD sensitive. Handling precaution recommended.
- Note 3.  $\theta_{JA}$  is measured in the natural convection at  $T_A = 25^{\circ}\text{C}$  on a low effective thermal conductivity test board of JEDEC 51-3 thermal measurement standard.
- Note 4. The device is not guaranteed to function outside its operating conditions.

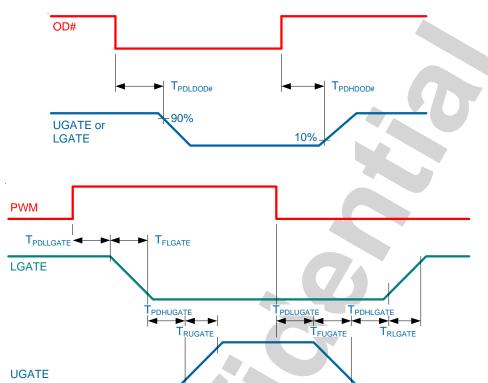


## Electrical Characteristics

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Supply Input						
Supply Input Voltage	V <sub>cc</sub>	_	10.8	7-	13.2	V
Supply Input Current	I <sub>cc</sub>	PWMx = OD#x = 0V, each channel		2	5	mA
VCCx POR Rising Threshold	V <sub>CCXRTH</sub>	V <sub>ccx</sub> rising	4.0	4.2	4.4	V
VCCx POR Hysteresis	V <sub>CCXHYS</sub>		4-7	0.25		V
PWM Input						
Input High Threshold	PWM <sub>RTH</sub>		3.15	3.45	3.75	V
Input Low Threshold	PWM <sub>FTH</sub>		0.6	0.9	1.2	V
PWM Floating Voltage	PWM <sub>FLT</sub>		1.75	1.95	2.15	V
	121	PWM = 0V	-420	-280	-140	uA
PWM Input Current	PWM	PWM = 5V	1.0	1.6	1.9	mA
Output Disable Input OD#x	1	. (9)				
Input High	OD# <sub>H</sub>		2.6			V
Input Low	OD# <sub>L</sub>				0.8	V
OD# Input Current	I <sub>OD#</sub>	OD# = 0V to 5V	-1		1	uA
Propagation Delay Time	T <sub>PDHOD#</sub>			20	45	ns
	T <sub>PDLOD#</sub>			20	45	ns
High Side Driver	1					
Output Resistance, Sourcing	R <sub>H_SRC</sub>	$V_{BOOT} - V_{PHASE} = 12V, I_{UGATE} = -80mA$		1.6	3.2	Ω
Output Resistance, Sinking	R <sub>H_SNK</sub>	$V_{BOOT} - V_{PHASE} = 12V, I_{UGATE} = 80mA$		1.5	3.0	Ω
Output Rising Time	T <sub>RUGATE</sub>	$V_{BOOT} - V_{PHASE} = 12V, C_{LOAD} = 3nF$		35	45	ns
Output Falling Time	T <sub>FUGATE</sub>	$V_{BOOT} - V_{PHASE} = 12V, C_{LOAD} = 3nF$		20	30	ns
Propagation Delay Time	T <sub>PDHUGATE</sub>	$V_{BOOT} - V_{PHASE} = 12V$		40	65	ns
	T <sub>PDLUGATE</sub>	V <sub>BOOT</sub> - V <sub>PHASE</sub> = 12V		20	35	ns
Low Side Driver						
Output Resistance, Sourcing	R <sub>L_SRC</sub>	$V_{CC} = 12V$ , $I_{LGATE} = -80$ mA		1.2	2.4	Ω
Output Resistance, Sinking	R <sub>L_SNK</sub>	$V_{CC} = 12V$ , $I_{LGATE} = 80$ mA		0.8	1.6	Ω
Output Rising Time	T <sub>RLGATE</sub>	$V_{CC} = 12V$ , $C_{LOAD} = 3nF$		35	45	ns
Output Falling Time	T <sub>FLGATE</sub>	$V_{CC} = 12V, C_{LOAD} = 3nF$		20	30	ns
Propagation Delay Time	T <sub>PDHLGATE</sub>	$V_{CC} = 12V$		40	65	ns
	T	V <sub>CC</sub> = 12V		20	35	ns

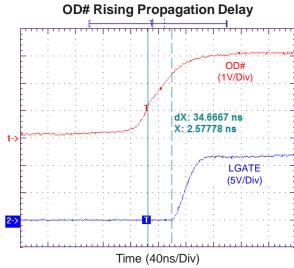


## Electrical Characteristics

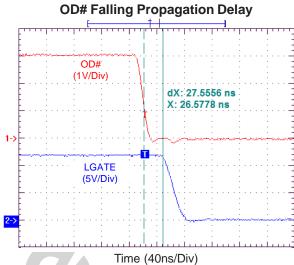




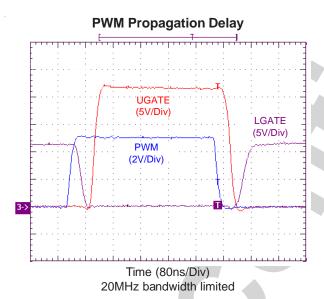
## **Typical Operation Characteristics**

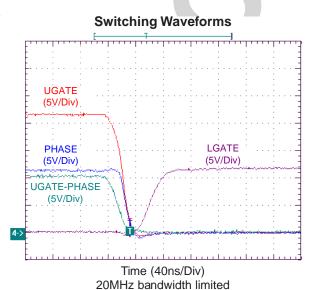


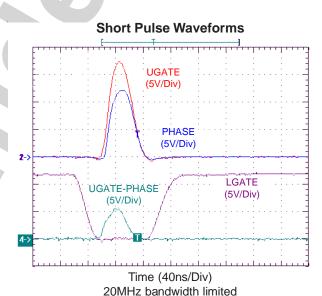
PWM = 0V, 20MHz bandwidth limited

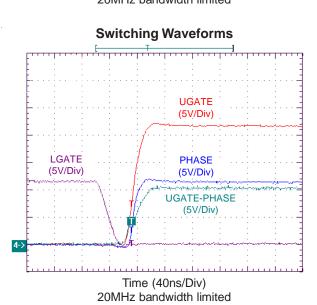


PWM = 0V, 20MHz bandwidth limited











## **Application Information**

#### **Thermal Consideration**

The power dissipation in uP1951 is dependent of the supply voltage, the PWM frequency and the input capacitance of the MOSFET:

$$= V_{CC} \{I_{CC} + [V_{CC}(C_{ISS\_U} + C_{ISS\_L}) + V_{IN} \times C_{OSS\_U}]f_{PWM}\}$$

where  $V_{CC}$  is the supply voltage,  $I_{CC}$  is the operation current of the control circuit,  $C_{ISS\_U}$  and  $C_{ISS\_L}$  are the total input capacitance of the upper and lower MOSFET respectively,  $V_{IN}$  is the supply voltage of the buck converter,  $C_{OSS\_U}$  is the reverse transfer capacitance regarding the Miller effect and  $f_{PWM}$  is the PWM input frequency. Take a typical case for example,  $V_{CC} = 12V$ ,  $I_{CC} = 1mA$ ,  $C_{ISS\_U} = 2x1.5nF$ ,  $C_{OSS\_U} = 2x0.1nF$ ,  $V_{IN} = 12V$ ,  $CISS\_L = 2x3nF$ ,  $f_{OSC} = 300kHz$  for each phase, the total power dissipation is calculated as:

#### $P_{loss}$

- $= 2 \times 12V\{1mA + [12V(3nF + 6nF) + 12V \times 0.2nF]300kHz\}$
- = 0.82W

The junction temperature raise is calculated as:

$$\Delta T = T_J - T_A = \theta_{JA} \times P_{LOSS}$$

- $= 68^{\circ} C/W \times 0.82W$
- $= 55.8^{\circ}C$

#### **Layout Consideration**

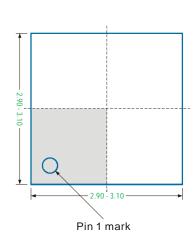
Follow the layout guidelines for optimum performance of uP1951.

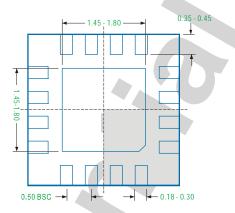
- 1.) Place the uP1951 physically near the power stages of buck converter.
- 2.) The exposed pad should be well soldered to the PCB with multiple vias to inner ground plane for optimum thermal performance.
- 3.) Place the bootstrap diode and capacitor physically near the IC.
- 4.) Connect the PHx, UGx, LGx to the power MOSFET with short and wide traces. Minimize the number of vias along the traces.
- 5.) The PWMx and OD#x are high impedence node and sensitive to noise. They should be kept away from noise source, especilly when the connecting traces between controller and driver are long.



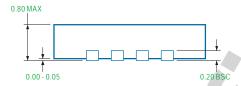
## **Package Information**

#### WQFN3x3 - 16L





Bottom View - Exposed Pad



#### Note

1. Package Outline Unit Description:

BSC: Basic. Represents theoretical exact dimension or dimension target

MIN: Minimum dimension specified. MAX: Maximum dimension specified.

REF: Reference. Represents dimension for reference use only. This value is not a device specification.

TYP. Typical. Provided as a general value. This value is not a device specification.

- 2. Dimensions in Millimeters.
- 3. Drawing not to scale.
- 4. These dimensions do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm.



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Headquarter 9F.,No.5, Taiyuan 1st St. Zhubei City, Hsinchu Taiwan, R.O.C.

TEL: 886.3.560.1666 FAX: 886.3.560.1888

Sales Branch Office 12F-5, No. 408, Ruiguang Rd. Neihu District, Taipei Taiwan, R.O.C.

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