

1 Features

- Ensured Linearity 0.01% Maximum
- Improved Performance in Existing Voltage-to-Frequency Conversion Applications
- Split or Single-Supply Operation
- Operates on Single 5-V Supply
- Pulse Output Compatible With All Logic Forms
- Excellent Temperature Stability: ±50 ppm/°C Maximum
- Low Power Consumption: 15 mW Typical at 5 V
- Wide Dynamic Range, 100 dB Minimum at 10-kHz Full Scale Frequency
- Wide Range of Full Scale Frequency:
 1 Hz to 100 kHz
- Low-Cost

2 Applications

- Voltage to Frequency Conversions
- · Frequency to Voltage Conversions
- Remote-Sensor Monitoring
- Tachometers

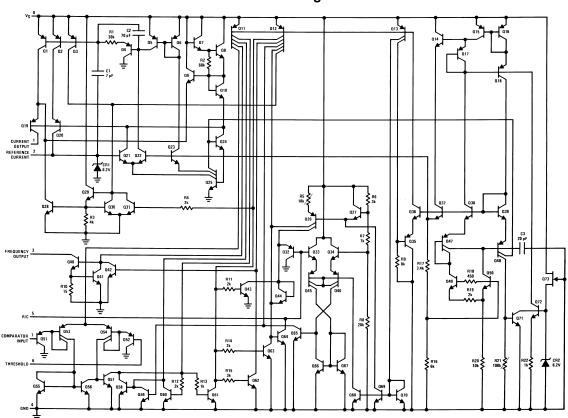
3 Description

The XDx31 family of voltage-to-frequency converters are ideally suited for use in simple low-cost circuits for analog-to-digital conversion, precision frequency-to-voltage conversion, long-term integration, linear frequency modulation or demodulation, and many other functions. The output when used as a voltage-to-frequency converter is a pulse train at a frequency precisely proportional to the applied input voltage. Thus, it provides all the inherent advantages of the voltage-to-frequency conversion techniques, and is easy to apply in all standard voltage-to-frequency converter applications.

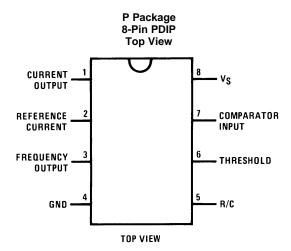
4 Device Information⁽¹⁾

PART NUMBER	PACKAGE
XD331-231	DIP8
XL331	SOP8

Schematic Diagram



5 Pin Configuration and Functions



Pin Functions

Р	IN	1/0	DESCRIPTION	
NAME	NO.	1/0		
IOUT	1	0	Current Output	
IREF	2	1	Reference Current	
FOUT	3	0	Frequency Output. This output is an open-collector output and requires a pullup resistor.	
GND	4	G	Ground	
RC	5	1	R-C filter input	
THRESH	6	I	Threshold input	
COMPIN	7	I	Comparator Input	
VS	8	Р	Supply Voltage	

6 Specifications

6.1 Absolute Maximum Ratings (1)(2)(3)

	MIN	MAX	UNIT
Supply Voltage, V _S		40	V
Output Short Circuit to Ground	Cont	inuous	
Output Short Circuit to V _{CC}	Cont	inuous	
Input Voltage	-0.2	+V _S	V
Lead Temperature (Soldering, 10 sec.) PDIP		260	°C

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.2 ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)(2)	±500	V

⁽¹⁾ JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

⁽²⁾ All voltages are measured with respect to GND = 0 V, unless otherwise noted.

⁽²⁾ Human body model, 100 pF discharged through a 1.5-kΩ resistor.

6.3 Recommended Operating Conditions

		MIN	MAX	UNIT
Operating Ambient Temperature	XD231	-25	85	°C
	XD331 XL331	0	70	ů
Supply Voltage, V _S ⁽¹		4	40	V

⁽¹⁾ All voltages are measured with respect to GND = 0 V, unless otherwise noted.

6.4 Thermal Information

		XD331 XL331	
	THERMAL METRIC ⁽¹⁾	P (PDIP)	UNIT
		8 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	100	°C/W

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report, SPRA953.

6.5 Electrical Characteristics

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
VFC Non-Linearity ⁽¹⁾		4.5 V ≤ V _S ≤ 20 V		±0.003	±0.01	% Full- Scale
VFC Non-Linearity		$T_{MIN} \le T_A \le T_{MAX}$		±0.006	±0.02	% Full- Scale
VFC Non-Linearity in Cir	cuit of Figure 14	$V_S = 15 \text{ V}, f = 10 \text{ Hz to } 11 \text{ kHz}$		±0.024	±0.14	%Full- Scale
Conversion Accuracy	XD231	$V_{IN} = -10 \text{ V}, \text{ R}_{S} = 14 \text{ k}\Omega$	0.95	1	1.05	kHz/V
Scale Factor (Gain)	XD331, XL331		0.9	1	1.1	kHz/V
Temperature Stability	XDx31	$T_{MIN} \le T_A \le T_{MAX}$ $4.5 \text{ V} \le V_S \le 20 \text{ V}$		±30	±150	ppm/°C
of Gain	XDx31			±20	±50	ppm/°C
Change of Coin with \/		4.5 V ≤ V _S ≤ 10 V		0.01	0.1	%/V
Change of Gain with V _S		10 V ≤ V _S ≤ 40 V		0.006	0.06	%/V
Rated Full-Scale Freque	ncy	V _{IN} = −10 V	10.0			kHz
Gain Stability vs. Time (1000 Hours)		$T_{MIN} \le T_A \le T_{MAX}$		±0.02		% Full- Scale
Over Range (Beyond Fu	II-Scale) Frequency	V _{IN} = −11 V	10%			
INPUT COMPARATO	R					
Offset Voltage				±3	±10	mV
XD231		$T_{MIN} \le T_A \le T_{MAX}$		±4	±14	mV
XD331/XL331		$T_{MIN} \le T_A \le T_{MAX}$		±3	±10	mV
Bias Current				-80	-300	nA
Offset Current				±8	±100	nA
Common-Mode Range		$T_{MIN} \le T_A \le T_{MAX}$	-0.2		V _{CC} - 2	V
TIMER						
Timer Threshold Voltage, Pin 5			0.63 × V _S	$0.667 \times V_S$	$0.7 \times V_S$	
Input Bias Current, Pin 5		V _S = 15 V				
All Devices		$0V \le V_{PIN 5} \le 9.9 V$		±10	±100	nA
XD231		V _{PIN 5} = 10 V		200	1000	nA
XD331/XL331		V _{PIN 5} = 10 V		200	500	nA
V _{SAT PIN 5} (Reset)		I = 5 mA		0.22	0.5	V

⁽¹⁾ Non-linearity is defined as the deviation of f_{OUT} from $V_{IN} \times (10 \text{ kHz/-}10 \text{ V}_{DC})$ when the circuit has been trimmed for zero error at 10 Hz and at 10 kHz, over the frequency range 1 Hz to 11 kHz. For the timing capacitor, C_T , use NPO ceramic, Teflon[®], or polystyrene.

Electrical Characteristics (continued)

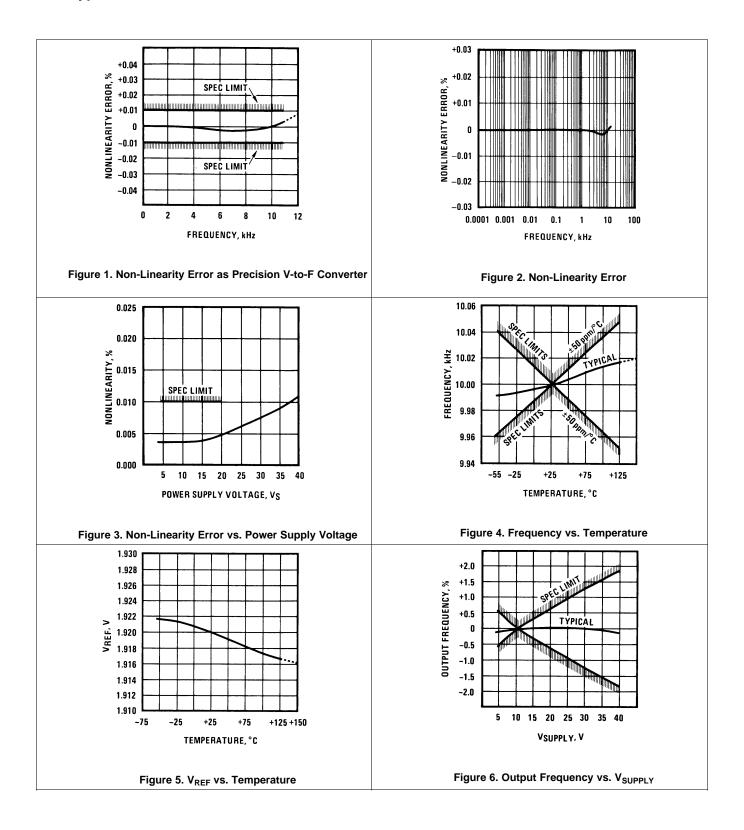
PAR	AMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
CURRENT SOURCE	E (PIN 1)					
0.1.10	XD231	$R_S = 14 \text{ k}\Omega, V_{PIN 1} = 0$	126	135	144	μΑ
Output Current	XD331, XL331		116	136	156	μΑ
Change with Voltage		0V ≤ V _{PIN 1} ≤ 10 V		0.2	1	μA
Current Source OFF	XD231 XD331 XL331			0.02	10	nA
Leakage	All Devices	$T_A = T_{MAX}$		2	50	nA
Operating Range of Cu	urrent (Typical)		(10 to 500)		μA
REFERENCE VOLT	AGE (PIN 2)				•	
XD231			1.76	1.89	2.02	V_{DC}
XD331, XL331			1.7	1.89	2.08	V_{DC}
Stability vs. Temperatu	ire			±60		ppm/°C
Stability vs. Time, 1000) Hours		±0.1%			
LOGIC OUTPUT (PI	N 3)		•		Ÿ	
		I = 5 mA		0.15	0.5	V
V_{SAT}		I = 3.2 mA (2 TTL Loads), $T_{MIN} \le T_A \le T_{MAX}$		0.1	0.4	V
OFF Leakage				±0.05	1	μΑ
SUPPLY CURRENT						
VD004		V _S = 5 V	2	3	4	mA
XD231		V _S = 40 V	2.5	4	6	mA
XD331, XL331		V _S = 5 V	1.5	3	6	mA
		V _S = 40 V	2	4	8	mA

6.6 Dissipation Ratings

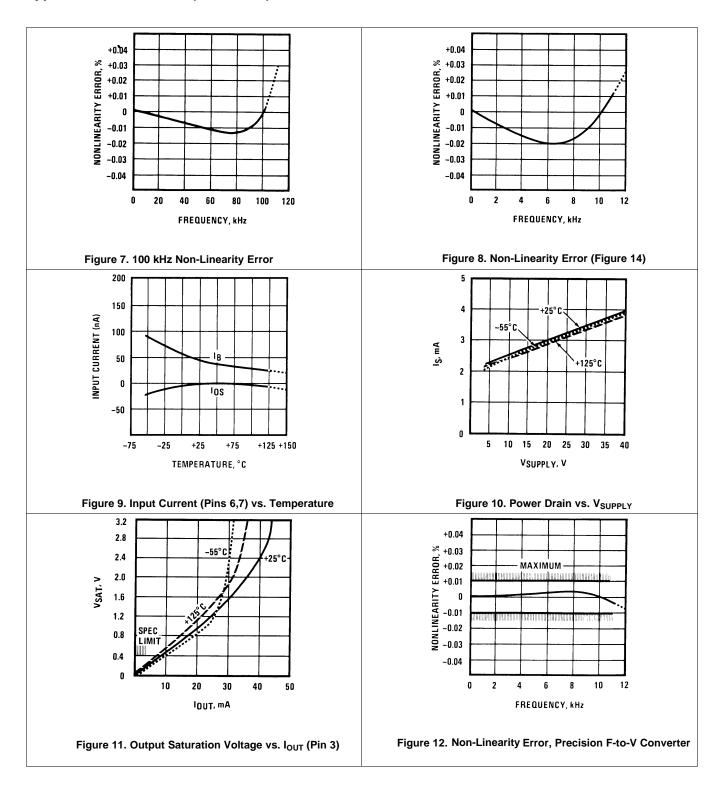
	VALUE	UNIT
Package Dissipation at 25°C ⁽¹⁾	1.25	W

⁽¹⁾ The absolute maximum junction temperature $(T_J max)$ for this device is 150°C. The maximum allowable power dissipation is dictated by $T_J max$, the junction-to-ambient thermal resistance (θ_{JA}) , and the ambient temperature T_A , and can be calculated using the formula $P_D max = (T_J max - T_A) / \theta_{JA}$. The values for maximum power dissipation will be reached only when the device is operated in a severe fault condition (e.g., when input or output pins are driven beyond the power supply voltages, or the power supply polarity is reversed). Obviously, such conditions should always be avoided.

6.7 Typical Characteristics



Typical Characteristics (continued)



7 Detailed Description

7.1 Overview

7.1.1 Detail of Operation, Functional Block Diagram

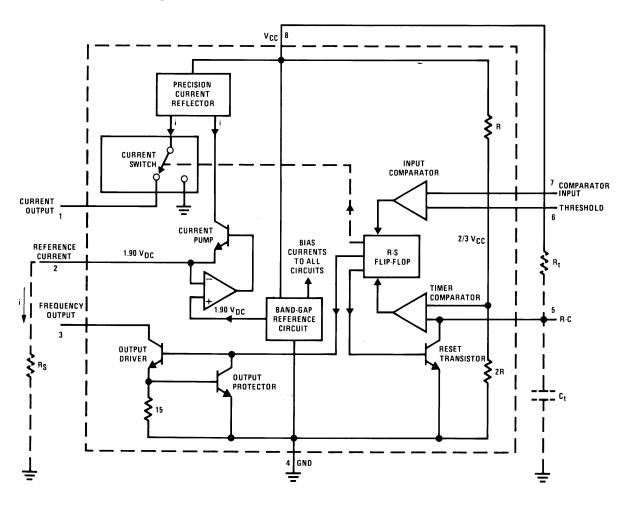
The Functional Block Diagram shows a band gap reference which provides a stable 1.9- V_{DC} output. This 1.9 V_{DC} is well regulated over a V_{S} range of 3.9 V to 40 V. It also has a flat, low temperature coefficient, and typically changes less than ½% over a 100°C temperature change.

The current pump circuit forces the voltage at pin 2 to be at 1.9 V, and causes a current $i = 1.90 \text{ V/R}_S$ to flow. For R_S =14 k, i=135 μ A. The precision current reflector provides a current equal to i to the current switch. The current switch switches the current to pin 1 or to ground, depending upon the state of the R-S flip-flop.

The timing function consists of an R-S flip-flop and a timer comparator connected to the external R_tC_t network. When the input comparator detects a voltage at pin 7 higher than pin 6, it sets the R-S flip-flop which turns ON the current switch and the output driver transistor. When the voltage at pin 5 rises to $\frac{2}{3}$ V_{CC} , the timer comparator causes the R-S flip-flop to reset. The reset transistor is then turned ON and the current switch is turned OFF.

However, if the input comparator still detects the voltage on pin 7 as higher than pin 6 when pin 5 crosses $\frac{2}{3}$ V_{CC}, the flip-flop will not be reset, and the current at pin 1 will continue to flow, trying to make the voltage at pin 6 higher than pin 7. This condition will usually apply under start-up conditions or in the case of an overload voltage at signal input. During this sort of overload the output frequency will be 0. As soon as the signal is restored to the working range, the output frequency will be resumed.

7.2 Functional Block Diagram



8 Application and Implementation

8.1 Application Information

8.1.1 Simplified Voltage-to-Frequency Converter

The operation of these blocks is best understood by going through the operating cycle of the basic V-to-F converter, Figure 13, which consists of the simplified block diagram of the XDx31 and the various resistors and capacitors connected to it.

The voltage comparator compares a positive input voltage, V1, at pin 7 to the voltage, V_x , at pin 6. If V1 is greater, the comparator will trigger the 1-shot timer. The output of the timer will turn ON both the frequency output transistor and the switched current source for a period $t = 1.1 R_t C_t$. During this period, the current i will flow out of the switched current source and provide a fixed amount of charge, $Q = i \times t$, into the capacitor, C_L . This will normally charge V_x up to a higher level than V1. At the end of the timing period, the current i will turn OFF, and the timer will reset itself.

Now there is no current flowing from pin 1, and the capacitor C_L will be gradually discharged by R_L until V_x falls to the level of V1. Then the comparator will trigger the timer and start another cycle.

The current flowing into C_L is exactly $I_{AVE} = i \times (1.1 \times R_t C_t) \times f$, and the current flowing out of C_L is exactly $V_x/R_L \simeq V_{IN}/R_L$. If V_{IN} is doubled, the frequency will double to maintain this balance. Even a simple V-to-F converter can provide a frequency precisely proportional to its input voltage over a wide range of frequencies.

9.1.2 Principles of Operation

The XDx31 are monolithic circuits designed for accuracy and versatile operation when applied as voltage-to-frequency (V-to-F) converters or as frequency-to-voltage (F-to-V) converters. A simplified block diagram of the XDx31 is shown in Figure 13 and consists of a switched current source, input comparator, and 1-shot timer.

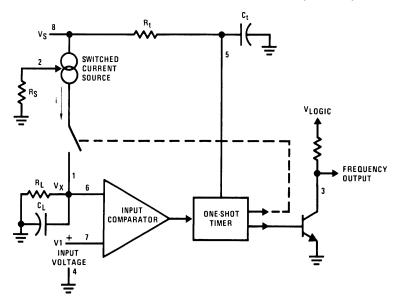
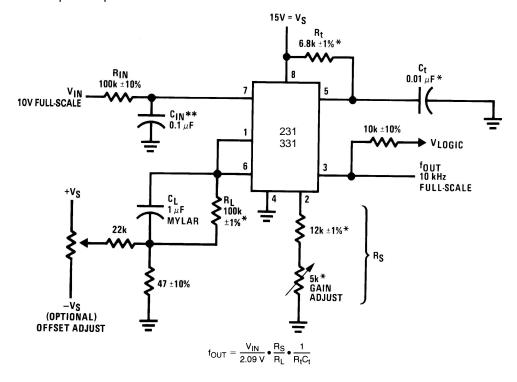


Figure 13. Simplified Block Diagram of Stand-Alone Voltage-to-Frequency Converter and External Components

8.2 Typical Applications

8.2.1 Basic Voltage-to-Frequency Converter

The simple stand-alone V-to-F converter shown in Figure 14 includes all the basic circuitry of Figure 13 plus a few components for improved performance.



^{*}Use stable components with low temperature coefficients. See Application Information.

Figure 14. Simple Stand-Alone V-to-F Converter with ±0.03% Typical Linearity (f = 10 Hz to 11 kHz)

8.2.1.1 Design Requirements

For this example, the system requirements are 0.05% linearity over an output frequency range of 10 Hz to 4 kHz with an input voltage range of 25 mV to 12.5 V. The available supply voltage is 15.0 V.

8.2.1.2 Detailed Design Procedure

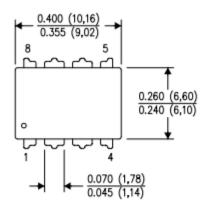
A capacitor C_{IN} is added from pin 7 to ground to act as a filter for V_{IN} , use of a 0.1 μF is appropriate for this application. A value of 0.01 μF to 0.1 μF will be adequate in most cases; however, in cases where better filtering is required, a 1- μF capacitor can be used. When the RC time constants are matched at pin 6 and pin 7, a voltage step at V_{IN} will cause a step change in f_{OUT} . If C_{IN} is much less than C_L , a step at V_{IN} may cause f_{OUT} to stop momentarily.

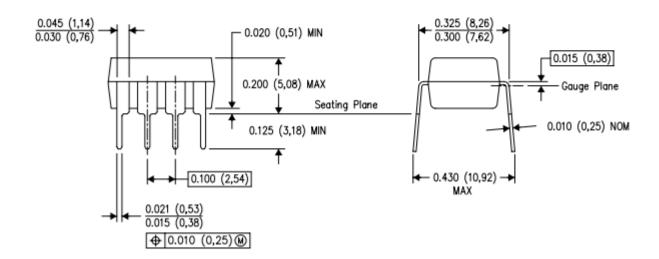
Next, we cancel the comparator bias current by setting R_{IN} to 100 k Ω to match R_L . This will help to minimize any frequency offset.

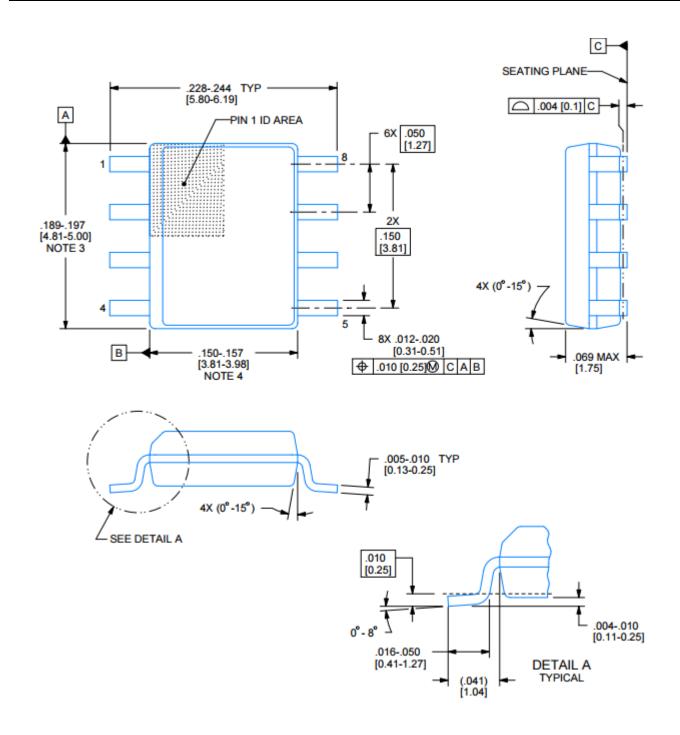
For best results, all the components should be stable low-temperature-coefficient components, such as metal-film resistors. The capacitor should have low dielectric absorption; depending on the temperature characteristics desired, NPO ceramic, polystyrene, Teflon or polypropylene are best suited.

The resistance R_S at pin 2 is made up of a 12-k Ω fixed resistor plus a 5-k Ω (cermet, preferably) gain adjust rheostat. The function of this adjustment is to trim out the gain tolerance of the XDx31, and the tolerance of R_t , and C_t .

^{**0.1} µF or 1 µF, See Typical Applications.







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