

## 1A Synchronous Step-Down DC/DC Converter

### **GENERAL DESCRIPTION**

The XM5062S-ADJ is a current mode synchronous step-down converter with fixed operating frequency. The supply current is only 20uA at operating and less than 1uA at shutdown. This device is the ideally solution for Li-lon battery powered system applications and small size board applications.

This device has internal switch, no external diode. It has internal fixed 1.5MHz frequency and makes application circuit smaller. It can reach very ripple with small capacity & inductor.

The XM5062S-ADJ is available in an adjustable output voltage version. The adjustable version has wide output range from 0.6V to VIN. The XM5062S-ADJ series products are available in a low profile (1mm) tiny SOT23-5 package.

### FEATURES

- 1A Output Current
- High Efficiency up to 96%
- 2.5V to 6.0 V Input Range
- Adjustable Output Voltage from 0.6V to VIN
- Low Quiescent Current 20uA
- No Schottky Diode Required
- 1.5MHz Constant Frequency Operation
- Low Dropout Operation: 100% Duty Cycle
- Current Mode Operation for Excellent Line and Load Transient Response
- Operating Temperature: -40°C to +85°C
- Available in tiny SOT23-5 Package

### **Applications**

Mobile Phone/Smartphone DVB GPS FPGA Core Power Supply Communications

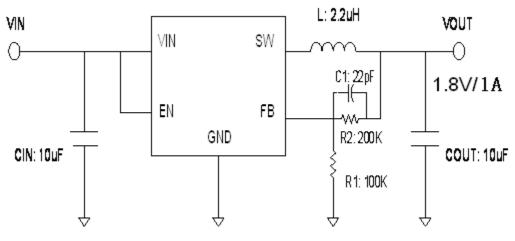


Figure 1. Typical Application Circuit



## **ORDERING INFORMATION**

PART NUMBER	TEMP RANGE	SWICHING FREQUENCY	OUTPUT VOLTAGE (V)	OUTPUT CURRENT (A)	PACKAGE	PINS	MARK
XM5062S-ADJ	-40°C to 85°C	1.5MHz	Adjustable	1	SOT23-5	5	6S-AYW

Note: "YW" is manufacture date code, "Y" means the year, "W" means the week

## **PIN CONFIGURATION**

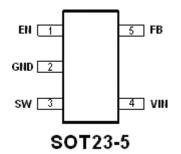


Figure 2. PIN Configuration

### **PIN DESCRIPTION**

PIN NUMBER	PIN NAME	PIN DESCRIPTION			
1 Enable Input. When higher than 1.5V, this pin turns the IC on. When lower than 0.3 this pin turns the IC off.					
2	GND	round.			
3	SW	Switch Output. Connect this pin to the inductor.			
4	4 VIN Power Supply.				
		Feedback Input. The voltage at this pin is regulated to 0.6V. Connect to the resistor divider between output and ground to set output voltage.			

### **ABSOLUTE MAXIMUM RATINGS**

(Note: Do not exceed these limits to prevent damage to the device. Exposure to absolute maximum rating conditions for long periods may affect device reliability.)

PARAMETER	VALUE	UNIT
Supply Voltage VIN	-0.3 to 6.5	V
FB, EN Voltage	-0.3 to VIN+0.3	V
SW Voltage	-0.3 to VIN+0.3	V
Operating Ambient Temperature	-40 to 85	°C



Maximum Junction Temperature	125	°C
Storage Temperature	-55 to 150	°C
Lead Temperature (Soldering, 10 sec)	300	°C

## **ELECTRICAL CHARACTERISTICS**

 $(V_{IN} = 3.6V, T_A = 25^{\circ}C$  unless otherwise specified)

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
Input Voltage Range	V <sub>IN</sub>		2.5		6.0	V	
UVLO Threshold	V <sub>UVLO</sub>	V <sub>HYSTERESIS</sub> =100mV	2.35	2.45	2.5	V	
Operating Supply Current		$V_{FB}$ =0.5V or $V_{OUT}$ =90%, $I_{Load}$ =0		130	170		
Standby Mode Supply Current	I <sub>SUPPLY</sub>	VFB =0.62V or VOUT =103%, ILoad =0 20		20	35	μA	
Shutdown Supply Current		V <sub>EN</sub> =0V, V <sub>IN</sub> =4.2V		0.1	1	1	
		T <sub>a</sub> =25°C	0.588	0.6	0.612		
Regulated Feedback Voltage	$V_{FB}$	0< T <sub>a</sub> <85°C	0.5865	0.6	0.6135	V	
C		-40°C < T <sub>a</sub> <85°C	0.585	0.6	0.615		
Reference Voltage Line Regulation		V <sub>IN</sub> =2.7V to 5.5V		0.04	0.4	%	
Regulated Output Voltage	V <sub>OUT</sub>	V <sub>OUT</sub> =1.8V; I <sub>OUT</sub> =100mA	1.746	1.8	1.854	V	
Output Voltage Load Regulation				0.5		%	
Peak Inductor Current	I <sub>PEAK</sub>	V <sub>IN</sub> =3V, V <sub>FB</sub> =0.5V or V <sub>OUT</sub> =90%,Duty Cycle<35%		1.8		А	
Oscillator Frequency	F <sub>osc</sub>	$V_{FB}$ =0.6V or $V_{OUT}$ =100%	1.2	1.5	1.8	MHz	
Uscillator r requency	I OSC	V <sub>FB</sub> =0 or V <sub>OUT</sub> =0		220		KHz	
Rds(ON) of P-channel FET		I <sub>SW</sub> =100mA		0.2	0.3	Ohm	
Rds(ON) of N-channel FET		I <sub>SW</sub> =-100mA		0.15	0.2	Ohm	
Enable Threshold		V <sub>IN</sub> = 2.5V to 5.5V	0.3	1	1.5	V	
Enable Leakage Current			-0.1		0.1	μA	
SW Leakage Current		$V_{EN} = 0V$ , $V_{SW} = 0V$ or 5V, $V_{IN} = 5V$	-1		1	uA	



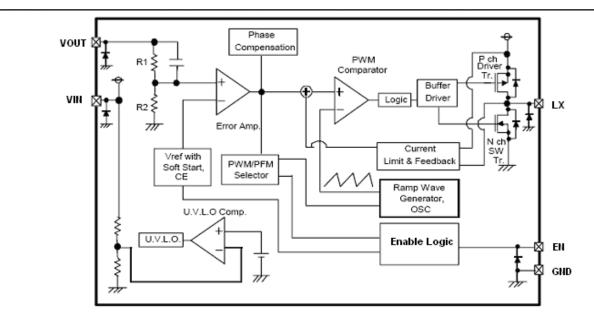


Figure 3. Functional Block Diagram

#### FUNCTIONAL DESCRIPTION

#### NORMAL OPERATION

In normal operation the high-side MOSFET turns on each cycle and remains on until the current comparator turns it off. At this point the low-side MOSFET turns on and remains on until either the end of the switching cycle or until the inductor current approaches zero. The error amplifier adjusts the current comparator's threshold as necessary in order to ensure that the output remains in regulation.

#### **OVER CURRENT OPERATION**

The part has internal current limit function, which is detected cycle by cycle. When its maximum inductor current limit is reached the charging cycle is terminated, and the low-side MOSFET is turned on to allow the inductor current to decrease. Under extreme overloads, such as short-circuit conditions, it reduces the oscillator frequency to 220KHz to allow further inductor current reduction and to minimize power dissipation.

### APPLICATION INFORMATION

#### INDUCTOR SELECTION

In normal operation, the inductor maintains continuous current to the output. The inductor current has a ripple that is dependent on the inductance value. The high inductance reduces the ripple current. In general, select the inductance by the following equation:

$$L = \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \cdot f \cdot VI}$$

Where  $V_{OUT}$  is the output voltage,  $V_{IN}$  is the input voltage, f is the switch frequency, and VI is the peak-to-peak inductor ripple current. Typically, choose VI as the 30% of the maximum output current.



Manuf acture r	Part Number	Induct ance(u H)	DRC max (Ohms )	Dimensions L*W*H(mm3)
Murata	LQH32PN	1	0.06	3.2*2.5*1.7
		2.2	0.09	
		4.7	0.15	
Sumid	CDRH3D 16	1.5	0.04	4*4*1.8
а		2.2	0.07	
		3.3	0.11	
		4.7	0.16	

 Table 1. Recommend Surface Mount Inductors

#### INPUT CAPACITOR SELECTION

The input capacitor reduces input voltage ripple to the converter, low ESR ceramic capacitor is highly recommended. For most applications, a 10uF capacitor is used. The input capacitor should be

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placed as close as possible to VIN and GND.

#### OUTPUT CAPACITOR SELECTION

A low ESR output capacitor is required in order to maintain low output voltage ripple. In the case of ceramic output capacitors, capacitor ESR is very small and does not contribute to the ripple, so a lower capacitance value is acceptable when ceramic capacitors are used. A 10uF ceramic output capacitor is suitable for most applications.

#### **OUTPUT VOLTAGE PROGRAMMING**

In the adjustable version, the output voltage is set by a resistive divider according to the following equation:

$$\boldsymbol{R}_2 = \boldsymbol{R}_1 \times \left( \frac{\boldsymbol{V}_{OUT}}{0.6} - 1 \right)$$

Typically choose R1=100K and determine R2 from the following equation:

Connect a small capacitor across R1 feed forward capacitance at the FB pin for better performance.

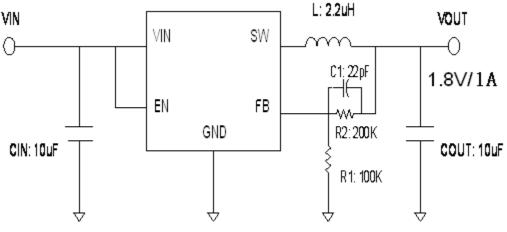
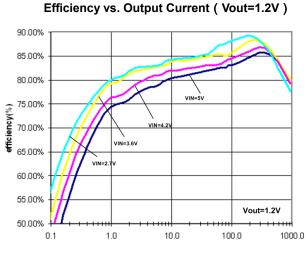


Figure 4. Typical Application Circuit for Adjustable Version

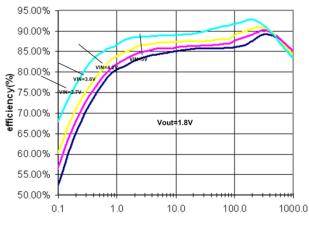


### **TYPICAL PERFORMANCE CHARACTERISTICS**

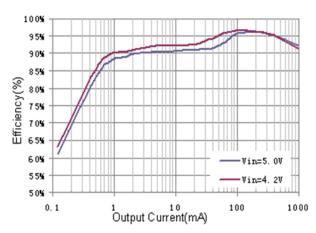
(VIN=VEN=5V, L=2.2uH, CIN=10uF, COUT=10uF, if not mentioned)

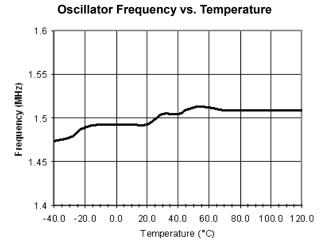


Efficiency vs. Output Current (Vout=1.8V)

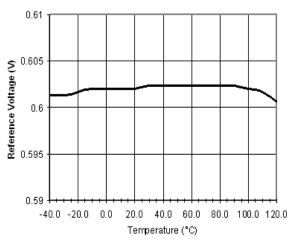


Efficiency vs. Output Current (Vout=3.3V)

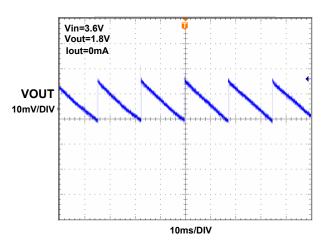




**Reference Voltage vs. Temperature** 



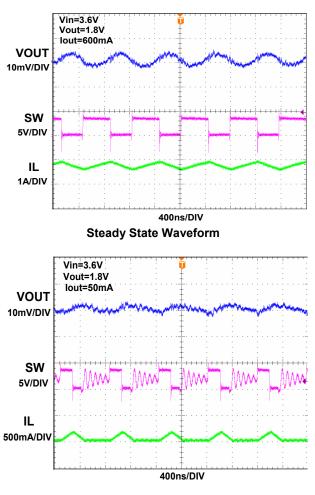
**Steady State Waveform** 



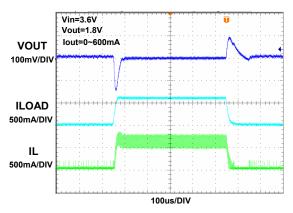
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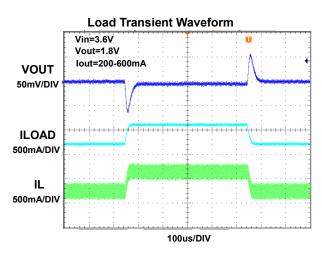


Steady State Waveform

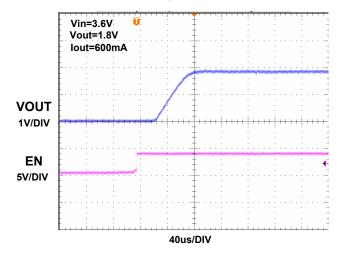








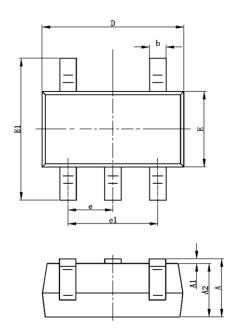
Startup through Enable Waveform

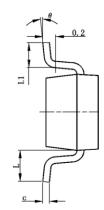




## **PACKAGE OUTLINE**

#### SOT23-5 PACKAGE OUTLINE AND DIMENSIONS





SYMB OL	I	NSION N ETERS	DIMENSION IN INCHES		
	MIN	MAX	MIN	MAX	
А	0.950	1.250	0.038	0.049	
A1	0.000	0.100	0.000	0.004	
A2	0.900	1.150	0.036	0.045	
b	0.300	0.400	0.012	0.016	
С	0.100	0.200	0.004	0.008	
D	2.820	3.020	0.111	0.119	
E	1.500	1.700	0.059	0.067	
E1	2.650	2.950	0.104	0.116	
е	0.950 TYP		0.037 TYP		
e1	1.800	2.000	0.071	0.079	
L	0.700 REF		0.028 REF		
L1	0.300	0.600	0.012	0.024	
θ	0° 8°		0°	8°	

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